HOLTEK

Sub-1GHz Transceiver Flash MCU

BC66F3663

Revision: V1.00 Date: November 01, 2023

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Features

CPU Features

- Operating Voltage
  - f_{SYS}=8MHz: 1.8V~3.6V
  - f_{SYS}=12MHz: 2.7V~3.6V
  - f_{SYS}=16MHz: 3.3V~3.6V
- Up to 0.25μs instruction cycle with 16MHz system clock at V_{DD}=3.3V
- Power down and wake-up functions to reduce power consumption
- Oscillator types
  - External High Speed Crystal – HXT
  - Internal High Speed 8/12/16MHz RC – HIRC
  - External Low Speed 32.768kHz Crystal – LXT
  - Internal Low Speed 32kHz RC – LIRC
- Multi-mode operation: FAST, SLOW, IDLE and SLEEP
- Fully integrated internal oscillators require no external components
- All instructions executed in 1~3 instruction cycles
- Table read instructions
- 115 powerful instructions
- 12-level subroutine nesting
- Bit manipulation instruction

Peripheral Features

- Flash Program Memory: 16K×16
- RAM Data Memory: 1K×8
- True EEPROM Memory: 1K×8
- Watchdog Timer function
- In Application Programming – IAP
- 24 bidirectional I/O lines
- Three external interrupt lines shared with I/O pins
- Programmable I/O port source current for LED applications
- Multiple Timer Modules for time measure, compare match output, PWM output function or single pulse output function
- Serial Interface Module – SIM for SPI or I^2C communication
- Two Fully-duplex / Half-duplex Asynchronous Receiver and Transmitter Interfaces – UARTs
- Software controlled 24-SCOM/SSEG lines LCD driver with 1/3 bias
- 11 external channel 12-bit resolution A/D converter with Programmable Internal Reference Voltage V_{VR}
- Single comparator function
- Dual Time Base functions for generation of fixed time interrupt signals
- Low voltage reset function
- Low voltage detect function
- Package type: 46-pin QFN

RF Features

- Frequency band: 315/433/470/868/915MHz
- OOK/GFSK modulation
- Supports 3-wire or 4-wire SPI interface
- Programmable data rate:
  - OOK: 0.5kbps~20kbps
  - GFSK: 2kbps~250kbps
- Programmable TX output power: up to 20dBm
- RF Low current consumption
  - 0.4μA deep sleep mode current with data retention
  - RX current consumption (AGC on & low data rate) @ 433.92MHz: 5.8mA
  - RX current consumption (AGC on & low data rate) @ 868.3MHz: 6.8mA
  - TX current consumption @ 433.92MHz: 33mA @ 10dBm P_{OUT}
  - TX current consumption @ 868.3MHz: 35mA @ 10dBm P_{OUT}
- High RX sensitivity (433.92MHz)
  - -117dBm at 2kbps on-air data rate
  - -100dBm at 250kbps on-air data rate
- High RX sensitivity (868.3MHz)
  - -116dBm at 2kbps on-air data rate
  - -100dBm at 250kbps on-air data rate
- On-chip VCO and Fractional-N synthesizer with internal loop filter
- Supports low cost crystal: 16MHz with integrated load capacitor
- AGC (Auto Gain Control) to achieve wide input range, up to +10dBm
- AFC (Auto Frequency Compensation) for frequency drift due to X'tal aging
- On-chip low power RC oscillator for WOR (Wake-on-RX) and WOT (Wake-on-TX) functions
- On-chip 8-bit RSSI (Received Signal Strength Indicator)
- Physical TX/RX FIFO buffers: TX 64 bytes, RX 64 bytes
- Simple FIFO/Block FIFO/Extend FIFO (up to 255 bytes) / Infinite FIFO modes
- Programmable threshold for carrier detection
- Frame synchronization recognition for both FIFO mode and Direct mode
- Packet handling
  - FEC (Forward Error Correction)
  - Data whitening
  - Manchester encoding
  - CRC-16 checking
- ATR (Auto-Transmit-Receive)
  - Auto-resend
  - Auto-acknowledgment
  - WOT+Auto-resend
  - WOR+Auto-acknowledgment

- Packet filtering
  - CRC filtering
  - Address filtering

General Description

The device is a Flash Memory type 8-bit high performance RISC architecture microcontroller with a sub-1GHz RF transceiver.

For memory features, the Flash Memory offers users the convenience of Flash Memory multi-programming features. Other memory includes an area of RAM Data Memory as well as an area of true EEPROM memory for storage of non-volatile data such as serial numbers, calibration data etc. By using the In Application Programming technology, user have a convenient means to directly store their measured data in the Flash Program Memory as well as having the ability to easily update their application programs.

Analog features include a multi-channel 12-bit A/D converter and a comparator function. Multiple and extremely flexible Timer Modules provide timing, pulse generation and PWM output functions. Communication with the outside world is catered for by including fully integrated SPI, I^2C and UART interface functions, three popular interfaces which provide designers with a means of easy communication with external peripheral hardware. Protective features such as an internal Watchdog Timer, Low Voltage Reset and Low Voltage Detector coupled with excellent noise immunity and ESD protection ensure that reliable operation is maintained in hostile electrical environments.

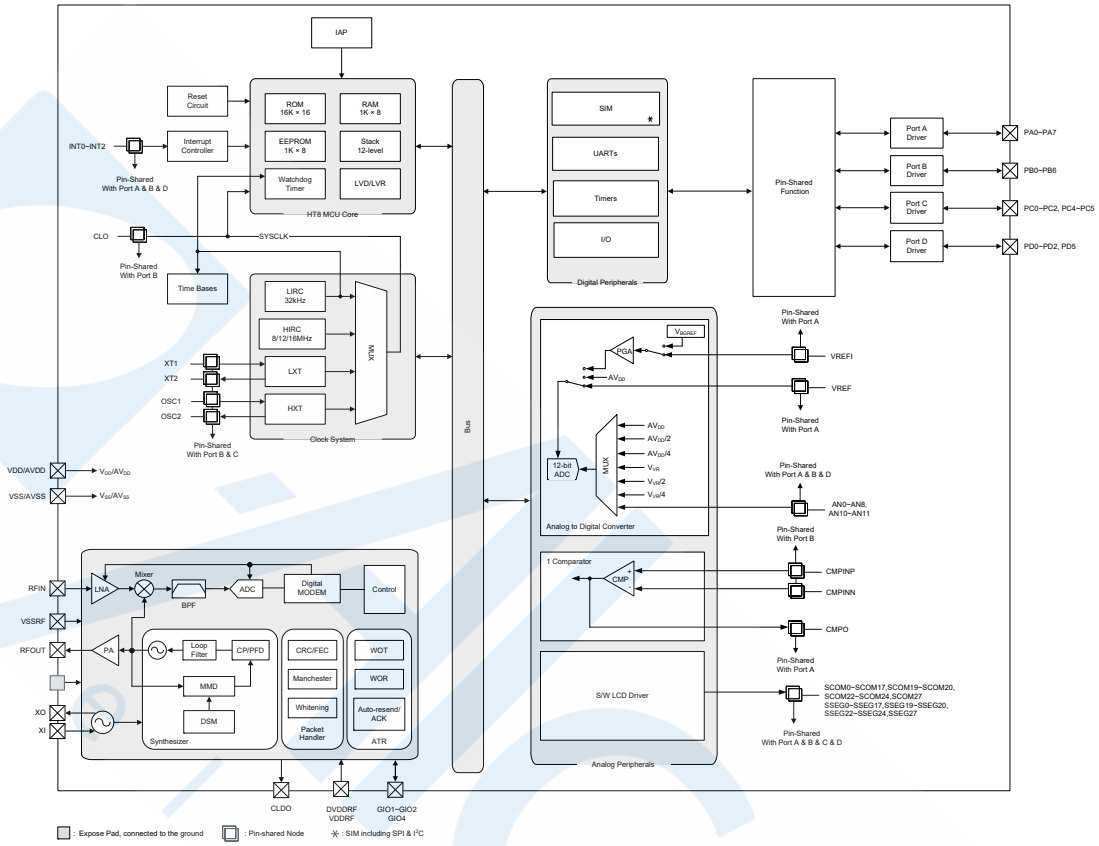
A full choice of external and internal low and high oscillator functions are provided, the fully integrated system oscillators require no external components for their implementation. The ability to operate and switch dynamically between a range of operating modes using different clock sources gives users the ability to optimise microcontroller operation and minimise power consumption.

The integrated RF transceiver is a high performance and low cost OOK/GFSK transceiver for wireless applications in the 315MHz, 433MHz, 470MHz, 868MHz and 915MHz frequency bands. It incorporates a highly integrated sub-1GHz transceiver and a baseband modem with programmable GFSK data rates from 2kbps to 250kbps and OOK data rates from 0.5kbps to 20kbps. Data handling features include 64-byte TX/RX FIFO and packet handling such as CRC generation, Forward Error Correction and data whitening, Manchester encoding.

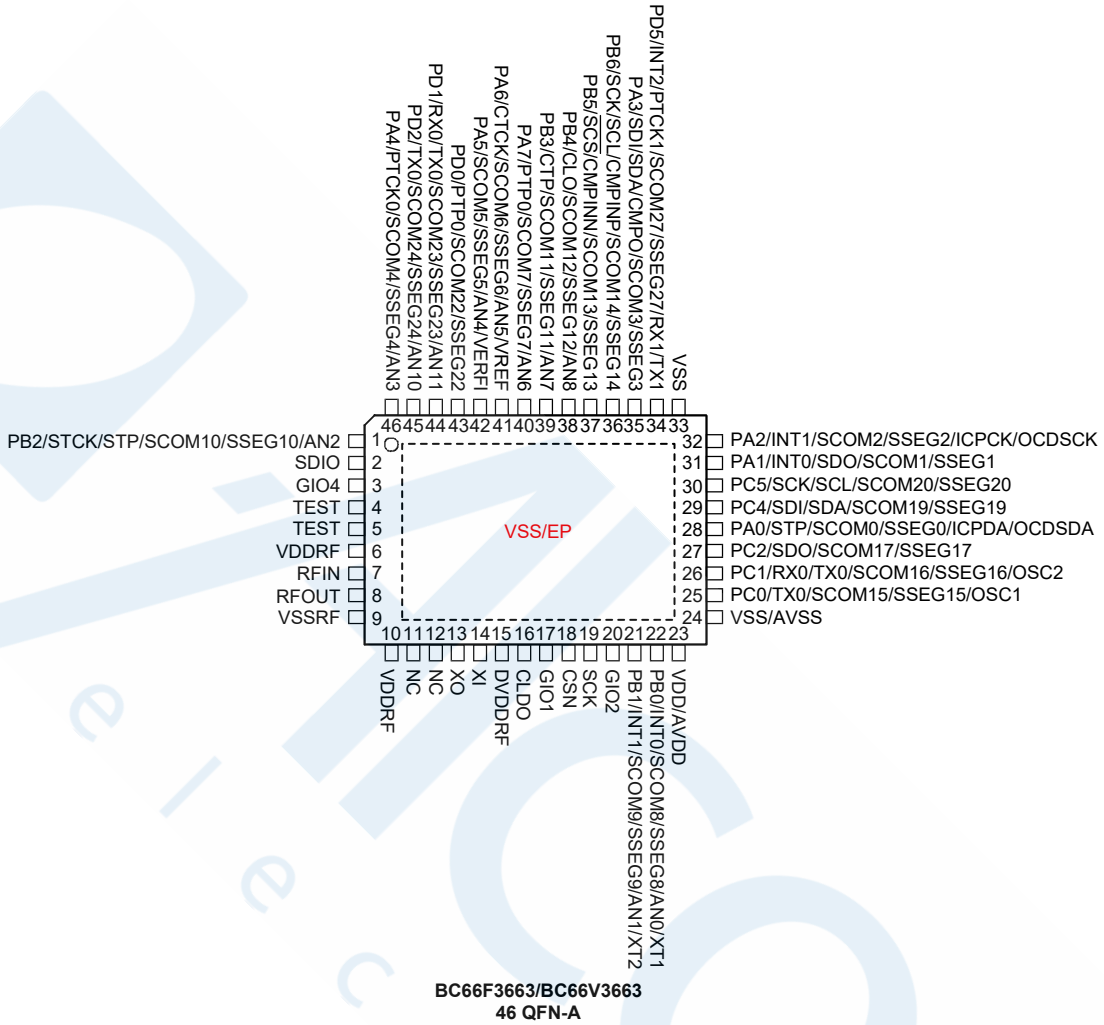
The integrated RF transceiver is optimized for the very low power consumption applications. At 433MHz band, its RX mode is operated at 5.8mA and it delivers +19dBm TX output power at 71mA current consumption. A low-noise low-IF receiver can achieve -117dBm sensitivity of 2kbps data rate at 433MHz bands. A Class-E Power Amplifier can deliver up to +20dBm output power at 433/868MHz bands. A fully integrated Fractional-N synthesizer can support a wide frequency range with a fine resolution. Both loop filter and XO load capacitors are integrated to on-chip to minimize the requirement for external components.

The inclusion of flexible I/O programming features, Time Base functions along with many other features ensure that the device will find excellent use in applications such as smart home, security, industrial/agricultural automation, data acquisition and recording in addition to many others.

Block Diagram



Pin Assignment



BC66F3663/BC66V3663
46 QFN-A

Note: 1. If the pin-shared pin functions have multiple outputs, the desired pin-shared function is determined by the corresponding software control bits.

2. The OCDSDA and OCDSCK pins are used as the OCDS dedicated pins and only available for the BC66V3663 device which is the OCDS EV chip of the BC66F3663.

3. For the unbonded lines, PC3, PC6, PD3~PD4 and PD6~PD7, the line status should be properly configured to avoid unwanted current consumption resulting from floating input conditions. Refer to the "Standby Current Considerations" and "Input/Output Ports" sections.

Pin Description

The function of each pin is listed in the following table, however the details behind how each pin is configured is contained in other sections of the datasheet.

Pin Name	Function	OPT	I/T	O/T	Description
PA0/STP/SCOM0/SSEG0/ ICPDA/OCDSDA	PA0	PAS0 PAWU PAPU	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	STP	PAS0	—	CMOS	STM output
	SCOM0	PAS0	—	SCOM	Software controlled LCD common output
	SSEG0	PAS0	—	SSEG	Software controlled LCD segment output
	ICPDA	—	ST	CMOS	ICP data/address pin
	OCDSDA	—	ST	CMOS	OCDS data/address pin, for EV chip only
PA1/INT0/SDO/SCOM1/ SSEG1	PA1	PAS0 PAWU PAPU	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	INT0	PAS0 INTEG INTC0 IFS0	ST	—	External interrupt 0
	SDO	PAS0	—	CMOS	SPI serial data output
	SCOM1	PAS0	—	SCOM	Software controlled LCD common output
	SSEG1	PAS0	—	SSEG	Software controlled LCD segment output
PA2/INT1/SCOM2/SSEG2/ ICPCK/OCDSCK	PA2	PAS0 PAWU PAPU	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	INT1	PAS0 INTEG INTC2 IFS0	ST	—	External interrupt 1
	SCOM2	PAS0	—	SCOM	Software controlled LCD common output
	SSEG2	PAS0	—	SSEG	Software controlled LCD segment output
	ICPCK	—	ST	—	ICP clock pin
	OCDSCK	—	ST	—	OCDS clock pin, for EV chip only
PA3/SDI/SDA/CMPO/ SCOM3/SSEG3	PA3	PAS0 PAWU PAPU	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	SDI	PAS0 IFS0	ST	—	SPI serial data input
	SDA	PAS0 IFS0	ST	NMOS	I^2C data line
	CMPO	PAS0	—	CMOS	Comparator output
	SCOM3	PAS0	—	SCOM	Software controlled LCD common output
	SSEG3	PAS0	—	SSEG	Software controlled LCD segment output
PA4/PTCK0/SCOM4/ SSEG4/AN3	PA4	PAS1 PAWU PAPU	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	PTCK0	PAS1	ST	—	PTM0 clock input
	SCOM4	PAS1	—	SCOM	Software controlled LCD common output
	SSEG4	PAS1	—	SSEG	Software controlled LCD segment output
	AN3	PAS1	AN	—	A/D Converter analog input channel 3

Pin Name	Function	OPT	I/T	O/T	Description
PA5/SCOM5/SSEG5/AN4/VREFI	PA5	PAS1 PAWU PAPU	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	SCOM5	PAS1	—	SCOM	Software controlled LCD common output
	SSEG5	PAS1	—	SSEG	Software controlled LCD segment output
	AN4	PAS1	AN	—	A/D Converter analog input channel 4
	VREFI	PAS1	AN	—	A/D Converter PGA input
PA6/CTCK/SCOM6/SSEG6/AN5/VREF	PA6	PAS1 PAWU PAPU	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	CTCK	PAS1	ST	—	CTM clock input
	SCOM6	PAS1	—	SCOM	Software controlled LCD common output
	SSEG6	PAS1	—	SSEG	Software controlled LCD segment output
	AN5	PAS1	AN	—	A/D Converter analog input channel 5
	VREF	PAS1	AN	—	A/D Converter external reference voltage input
PA7/PTP0/SCOM7/SSEG7/AN6	PA7	PAS1 PAWU PAPU	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up
	PTP0	PAS1	—	CMOS	PTM0 output
	SCOM7	PAS1	—	SCOM	Software controlled LCD common output
	SSEG7	PAS1	—	SSEG	Software controlled LCD segment output
	AN6	PAS1	AN	—	A/D Converter analog input channel 6
PB0/INT0/SCOM8/SSEG8/AN0/XT1	PB0	PBS0 PBPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	INT0	PBS0 INTEG INTC0 IFS0	ST	—	External interrupt 0
	SCOM8	PBS0	—	SCOM	Software controlled LCD common output
	SSEG8	PBS0	—	SSEG	Software controlled LCD segment output
	AN0	PBS0	AN	—	A/D Converter analog input channel 0
	XT1	PBS0	AN	—	LXT oscillator pin
PB1/INT1/SCOM9/SSEG9/AN1/XT2	PB1	PBS0 PBPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	INT1	PBS0 INTEG INTC2 IFS0	ST	—	External Interrupt 1
	SCOM9	PBS0	—	SCOM	Software controlled LCD common output
	SSEG9	PBS0	—	SSEG	Software controlled LCD segment output
	AN1	PBS0	AN	—	A/D Converter analog input channel 1
	XT2	PBS0	—	AN	LXT oscillator pin
PB2/STCK/STP/SCOM10/SSEG10/AN2	PB2	PBS0 PBPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	STCK	PBS0	ST	—	STM clock input
	STP	PBS0	—	CMOS	STM output
	SCOM10	PBS0	—	SCOM	Software controlled LCD common output
	SSEG10	PBS0	—	SSEG	Software controlled LCD segment output
	AN2	PBS0	AN	—	A/D Converter analog input channel 2

Pin Name	Function	OPT	I/T	O/T	Description
PB3/CTP/SCOM11/ SSEG11/AN7	PB3	PBS0 PBPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	CTP	PBS0	—	CMOS	CTM output
	SCOM11	PBS0	—	SCOM	Software controlled LCD common output
	SSEG11	PBS0	—	SSEG	Software controlled LCD segment output
	AN7	PBS0	AN	—	A/D Converter analog input channel 7
PB4/CLO/SCOM12/ SSEG12/AN8	PB4	PBS1 PBPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	CLO	PBS1	—	CMOS	System clock output
	SCOM12	PBS1	—	SCOM	Software controlled LCD common output
	SSEG12	PBS1	—	SSEG	Software controlled LCD segment output
	AN8	PBS1	AN	—	A/D Converter analog input channel 8
PB5/$\overline{SCS}$/CMPINN/ SCOM13/SSEG13	PB5	PBS1 PBPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	$\overline{SCS}$	PBS1 IFS0	ST	CMOS	SPI slave select pin
	CMPINN	PBS1	AN	—	Comparator negative input
	SCOM13	PBS1	—	SCOM	Software controlled LCD common output
	SSEG13	PBS1	—	SSEG	Software controlled LCD segment output
PB6/SCK/SCL/CMPINP/ SCOM14/SSEG14	PB6	PBS1 PBPU	ST	CMOS	General purpose I/O. Register enabled pull- high
	SCK	PBS1 IFS0	ST	CMOS	SPI serial clock
	SCL	PBS1 IFS0	ST	NMOS	I^2C clock line
	CMPINP	PBS1	AN	—	Comparator positive input
	SCOM14	PBS1	—	SCOM	Software controlled LCD common output
	SSEG14	PBS1	—	SSEG	Software controlled LCD segment output
PC0/TX0/SCOM15/ SSEG15/OSC1	PC0	PCS0 PCPU	ST	CMOS	General purpose I/O. Register enabled pull- high
	TX0	PCS0	—	CMOS	UART0 serial data output
	SCOM15	PCS0	—	SCOM	Software controlled LCD common output
	SSEG15	PCS0	—	SSEG	Software controlled LCD segment output
	OSC1	PCS0	AN	—	HXT oscillator pin
PC1/RX0/TX0/SCOM16/ SSEG16/OSC2	PC1	PCS0 PCPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	RX0/TX0	PCS0 IFS0	ST	CMOS	UART0 serial data input in full-duplex communication or UART0 serial data input/output in single wire mode communication
	SCOM16	PCS0	—	SCOM	Software controlled LCD common output
	SSEG16	PCS0	—	SSEG	Software controlled LCD segment output
	OSC2	PCS0	—	AN	HXT oscillator pin
PC2/SDO/SCOM17/ SSEG17	PC2	PCS0 PCPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	SDO	PCS0	—	CMOS	SPI serial data output
	SCOM17	PCS0	—	SCOM	Software controlled LCD common output
	SSEG17	PCS0	—	SSEG	Software controlled LCD segment output

Pin Name	Function	OPT	I/T	O/T	Description
PC4/SDI/SDA/SCOM19/ SSEG19	PC4	PCS1 PCPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	SDI	PCS1 IFS0	ST	—	SPI serial data input
	SDA	PCS1 IFS0	ST	NMOS	I²C data line
	SCOM19	PCS1	—	SCOM	Software controlled LCD common output
	SSEG19	PCS1	—	SSEG	Software controlled LCD segment output
PC5/SCK/SCL/SCOM20/ SSEG20	PC5	PCS1 PCPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	SCK	PCS1 IFS0	ST	CMOS	SPI serial clock
	SCL	PCS1 IFS0	ST	NMOS	I²C clock line
	SCOM20	PCS1	—	SCOM	Software controlled LCD common output
	SSEG20	PCS1	—	SSEG	Software controlled LCD segment output
PD0/PTP0/SCOM22/ SSEG22	PD0	PDS0 PDPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	PTP0	PDS0	—	CMOS	PTM0 output
	SCOM22	PDS0	—	SCOM	Software controlled LCD common output
	SSEG22	PDS0	—	SSEG	Software controlled LCD segment output
PD1/RX0/TX0/SCOM23/ SSEG23/AN11	PD1	PDS0 PDPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	RX0/TX0	PDS0 IFS0	ST	CMOS	UART0 serial data input in full-duplex communication or UART0 serial data input/output in single wire mode communication
	SCOM23	PDS0	—	SCOM	Software controlled LCD common output
	SSEG23	PDS0	—	SSEG	Software controlled LCD segment output
	AN11	PDS0	AN	—	A/D Converter analog input channel 11
PD2/TX0/SCOM24/ SSEG24/AN10	PD2	PDS0 PDPU	ST	CMOS	General purpose I/O. Register enabled pull-high
	TX0	PDS0	—	CMOS	UART0 serial data output
	SCOM24	PDS0	—	SCOM	Software controlled LCD common output
	SSEG24	PDS0	—	SSEG	Software controlled LCD segment output
	AN10	PDS0	AN	—	A/D Converter analog input channel 10
PD5/INT2/PTCK1/SCOM27/ SSEG27/RX1/TX1	PD5	PDS1	ST	CMOS	General purpose I/O. Register enabled pull-high
	INT2	PDS1 INTEG INTC3 IFS0	ST	—	External interrupt 2
	PTCK1	PDS1 IFS1	ST	—	PTM1 clock input
	SCOM27	PDS1	—	SCOM	Software controlled LCD common output
	SSEG27	PDS1	—	SSEG	Software controlled LCD segment output
	RX1/TX1	PDS1 IFS0	ST	CMOS	UART1 serial data input in full-duplex communication or UART1 serial data input/output in single wire mode communication
GIO1	GIO1	—	ST	CMOS	RF Multi-function I/O 1
GIO2	GIO2	—	ST	CMOS	RF Multi-function I/O 2
GIO4	GIO4	—	ST	CMOS	RF Multi-function I/O 4
CSN	CSN	—	ST	—	RF SPI select input
SCK	SCK	—	ST	—	RF SPI clock input

Pin Name	Function	OPT	I/T	O/T	Description
SDIO	SDIO	—	ST	—	RF SPI data input/output
TEST	TEST	—	—	—	Not connected, leave floating
VDDRF	VDDRF	—	PWR	—	RF Analog positive power supply
DVDDRF	DVDDRF	—	PWR	—	RF Digital positive power supply
RFIN	RFIN	—	AN	—	RF LNA input
RFOUT	RFOUT	—	—	AN	RF power amplifier output
VSSRF	VSSRF	—	PWR		RF ground
XO	XO	—	—	AO	RF Crystal oscillator output
XI	XI	—	—	AI	RF Crystal oscillator input
CLDO	CLDO	—	—	PWR	RF LDO output, connected to a bypass capacitor
NC	NC	—	—	—	Not connected
—	VSS/EP	—	PWR	—	Exposed pad, must be connected to ground
VDD/AVDD	VDD	—	PWR	—	Digital positive power supply
	AVDD	—	PWR	—	Analog positive power supply
VSS/AVSS	VSS	—	PWR	—	Digital negative power supply, ground
	AVSS	—	PWR	—	Analog negative power supply, ground

Legend: I/T: Input type;
O/T: Output type;
OPT: Optional by register option;
AN: Analog signal;
PWR: Power;
CMOS: CMOS output;
ST: Schmitt Trigger input;
NMOS: NMOS output;
SSEG: Software controlled LCD SEG;
SCOM: Software controlled LCD COM.

1. The VSS/EP pin is located at the exposed pad.
2. The backside plate of EP shall be well soldered to ground on PCB, otherwise it will downgrade RF performance.

Absolute Maximum Ratings

Supply Voltage ..... V_{SS}-0.3V to 3.6V
Input Voltage ..... V_{SS}-0.3V to V_{DD}+0.3V
Storage Temperature ..... -60°C to 150°C
Operating Temperature ..... -40°C to 85°C
I_{OL} Total 80mA
I_{OH} Total ..... -80mA
Total Power Dissipation 500mW
ESD HBM ..... ±2kV

* The device is ESD sensitive. HBM (Human Body Mode) is based on MIL-STD-883.

Note: These are stress ratings only. Stresses exceeding the range specified under "Absolute Maximum Ratings" may cause substantial damage to the device. Functional operation of this device at other conditions beyond those listed in the specification is not implied and prolonged exposure to extreme conditions may affect device reliability.

D.C. Characteristics

For data in the following tables, note that factors such as oscillator type, operating voltage, operating frequency, pin load conditions, temperature and program instruction type, etc., can all exert an influence on the measured values.

Operating Voltage Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
V_{DD}	Operating Voltage – HXT	—	f_{SYS}=8MHz	1.8	—	3.6	V
			f_{SYS}=12MHz	2.7	—	3.6	
			f_{SYS}=16MHz	3.3	—	3.6	
	Operating Voltage – HIRC	—	f_{SYS}=8MHz	1.8	—	3.6	V
			f_{SYS}=12MHz	2.7	—	3.6	
			f_{SYS}=16MHz	3.3	—	3.6	
	Operating Voltage – LXT	—	f_{SYS}=32.768kHz	1.8	—	3.6	V
	Operating Voltage – LIRC	—	f_{SYS}=32kHz	1.8	—	3.6	V

Operating Current Characteristics

Ta=-40°C~85°C

Symbol	Operation Mode	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
I_{DD}	SLOW Mode – LIRC	1.8V	f_{SYS}=32kHz	—	8	16	μA
		3V		—	10	20	
	SLOW Mode – LXT	1.8V	f_{SYS}=32768Hz	—	8	16	μA
		3V		—	10	20	
I_{DD}	FAST Mode – HIRC	1.8V	f_{SYS}=8MHz	—	0.8	1.2	mA
		3V		—	1.0	1.5	
		2.7V	f_{SYS}=12MHz	—	1.2	2.2	mA
		3V		—	1.50	2.75	
		3.3V	f_{SYS}=16MHz	—	3.2	4.8	mA
	FAST Mode – HXT	1.8V	f_{SYS}=8MHz	—	0.8	1.2	mA
		3V		—	1.0	1.5	
		2.7V	f_{SYS}=12MHz	—	1.2	2.2	mA
		3V		—	1.50	2.75	
		3.3V	f_{SYS}=16MHz	—	3.2	4.8	mA

Note: When using the characteristic table data, the following notes should be taken into consideration:

1. Any digital inputs are setup in a non-floating condition.
2. All measurements are taken under conditions of no load and with all peripherals in an off state.
3. There are no DC current paths.
4. All Operating Current values are measured using a continuous NOP instruction program loop.

Standby Current Characteristics

Ta=25°C, unless otherwise specified

Symbol	Operation Mode	Test Conditions		Min.	Typ.	Max.	Max. @85°C	Unit
		V_{DD}	Conditions					
I_{STB}	SLEEP Mode	1.8V	WDT off	—	0.45	1.50	6.00	μA
		3V		—	0.45	1.80	7.00	
		1.8V	WDT on	—	1.5	3.0	7.0	μA
		3V		—	1.8	3.6	8.0	
	IDLE0 Mode – LIRC	1.8V	f_{SUB} on	—	2.4	4.0	8.0	μA
		3V		—	3.0	5.0	9.0	
	IDLE0 Mode – LXT	1.8V	f_{SUB} on	—	2.4	4.0	8.0	μA
		3V		—	3	5	9	
	IDLE1 Mode – HIRC	1.8V	f_{SUB} on, f_{SYS}=8MHz	—	288	400	480	μA
		3V		—	360	500	600	
		2.7V	f_{SUB} on, f_{SYS}=12MHz	—	550	700	800	μA
		3V		—	650	800	900	
		3.3V	f_{SUB} on, f_{SYS}=16MHz	—	1.8	3.6	4.4	mA
I_{STB}	IDLE1 Mode – HXT	1.8V	f_{SUB} on, f_{SYS}=8MHz	—	288	400	480	μA
		3V		—	360	500	600	
		2.7V	f_{SUB} on, f_{SYS}=12MHz	—	432	600	720	μA
		3V		—	540	750	900	
		3.3V	f_{SUB} on, f_{SYS}=16MHz	—	1.8	3.6	4.4	mA

Note: When using the characteristic table data, the following notes should be taken into consideration:

1. Any digital inputs are setup in a non-floating condition.
2. All measurements are taken under conditions of no load and with all peripherals in an off state.
3. There are no DC current paths.
4. All Standby Current values are taken after a HALT instruction execution thus stopping all instruction execution.

A.C. Characteristics

For data in the following tables, note that factors such as oscillator type, operating voltage, operating frequency and temperature etc., can all exert an influence on the measured values.

High Speed Internal Oscillator – HIRC – Frequency Accuracy

During the program writing operation the writer will trim the HIRC oscillator at a user selected HIRC frequency and a fixed voltage of 3V.

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Temp.				
f_{HIRC}	8MHz Writer Trimmed HIRC Frequency	3V	25°C	-1%	8	+1%	MHz
			-40°C~85°C	-2%	8	+2%	
		2.7V~3.6V	25°C	-2.5%	8	+2.5%	
			-40°C~85°C	-3%	8	+3%	
		2.2V~3.6V	25°C	-3.5%	8	+3.5%	
			-40°C~85°C	-5%	8	+5%	
		1.8V~3.6V	25°C	-5%	8	+5%	
			-40°C~85°C	-10%	8	+10%	

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Temp.				
f_{HIRC}	12MHz Writer Trimmed HIRC Frequency	3V	25°C	-1%	12	+1%	MHz
			-40°C~85°C	-2%	12	+2%	
		2.7V~3.6V	25°C	-2.5%	12	+2.5%	
			-40°C~85°C	-3%	12	+3%	
	16MHz Writer Trimmed HIRC Frequency	3.3V~3.6V	25°C	-2.5%	16	+2.5%	MHz
			-40°C~85°C	-3%	16	+3%	

Note: 1. The 3V value for V_{DD} is provided as this is the fixed voltage at which the HIRC frequency is trimmed by the writer.
2. The row below the 3V trim voltage row is provided to show the values for the full V_{DD} range operating voltage.
3. The minimum and maximum tolerance values provided in the table are for the frequency at which the writer trims the HIRC oscillator. After trimming at this chosen specific frequency any change in HIRC oscillator frequency using the oscillator register control bits by the application program will give a frequency tolerance to within ±20%.

External High Speed Crystal/Ceramic Oscillator Characteristics – HXT

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Temp.				
f_{HXT}	System Clock – HXT	1.8V~3.6V	-40°C~85°C	—	8	—	MHz
		2.7V~3.6V	-40°C~85°C	—	12	—	
		3.3V~3.6V	-40°C~85°C	—	16	—	

Low Speed Internal Oscillator Characteristics – LIRC

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Temp.				
f_{LIRC}	LIRC Frequency	3V	25°C	-2%	32	+2%	kHz
		1.8V~3.6V	-40°C~85°C	-10%	32	+10%	
t_{START}	LIRC Start-up Time	—	-40°C~85°C	—	—	100	μs

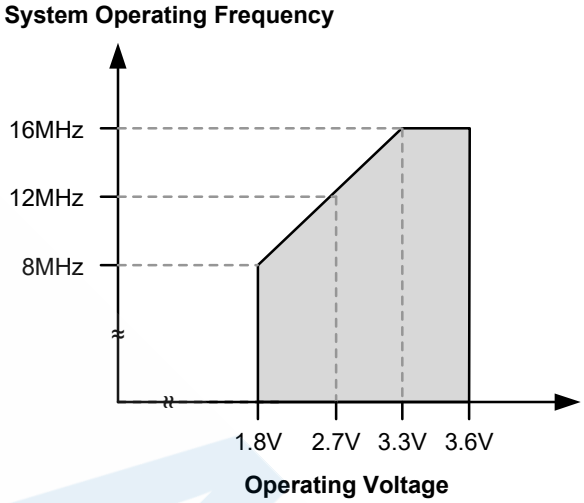
External Low Speed Crystal Oscillator Characteristics – LXT

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
f_{LXT}	System Clock – LXT	1.8V~3.6V	—	—	32768	—	Hz
t_{START}	LXT Start Up Time	3V	—	—	—	1000	ms
Duty Cycle	Duty Cycle	—	—	40	—	60	%
R_{NEG}	Negative Resistance	1.8V	—	3×ESR	—	—	Ω

Note: C1, C2 and R_P are external components. C1=C2=10pF. R_P=10MΩ. C_L=7pF, ESR=30kΩ.

Operating Frequency Characteristic Curves



System Start Up Time Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
t_{SST}	System Start-up Time (Wake-up from Condition where f_{SYS} is off)	—	$f_{SYS}=f_H \sim f_H/64$, $f_H=f_{HXT}$	—	128	—	t_{HXT}
		—	$f_{SYS}=f_H \sim f_H/64$, $f_H=f_{HIRC}$	—	16	—	t_{HIRC}
		—	$f_{SYS}=f_{SUB}=f_{LXT}$	—	1024	—	t_{LXT}
		—	$f_{SYS}=f_{SUB}=f_{LIRC}$	—	2	—	t_{LIRC}
	System Start-up Time (Wake-up from Condition where f_{SYS} is on)	—	$f_{SYS}=f_H \sim f_H/64$, $f_H=f_{HXT}$ or f_{HIRC}	—	2	—	t_H
		—	$f_{SYS}=f_{SUB}=f_{LXT}$ or f_{LIRC}	—	2	—	t_{SUB}
	System Speed Switch Time (FAST to SLOW Mode or SLOW to FAST Mode)	—	f_{HXT} switches from off → on	—	1024	—	t_{HXT}
		—	f_{HIRC} switches from off → on	—	16	—	t_{HIRC}
		—	f_{LXT} switches from off → on	—	1024	—	t_{LXT}
t_{RSTD}	System Reset Delay Time (Reset source from Power-on Reset or LVR Hardware Reset)	—	RR_{POR}=5V/ms	14	16	18	ms
	System Reset Delay Time (LVRC/WDTC/RSTC Register Software Reset)	—	—				
	System Reset Delay Time (Reset Source from WDT Overflow Reset)	—	—				
t_{SRESET}	Minimum Software Reset Width to Reset	—	—	45	90	120	μs

Note: 1. For the System Start-up time values, whether f_{SYS} is on or off depends upon the mode type and the chosen f_{SYS} system oscillator. Details are provided in the System Operating Modes section.

2. The time units, shown by the symbols t_{HIRC} etc. are the inverse of the corresponding frequency values as provided in the frequency tables. For example $t_{HIRC}=1/f_{HIRC}$, $t_{SYS}=1/f_{SYS}$ etc.

3. If the LIRC is used as the system clock and if it is off when in the SLEEP Mode, then an additional LIRC start up time, t_{START}, as provided in the LIRC frequency table, must be added to the t_{SST} time in the table above.

4. The System Speed Switch Time is effectively the time taken for the newly activated oscillator to start up.

Input/Output Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
V_{IL}	Input Low Voltage for I/O Ports	—	—	0	—	$0.2V_{DD}$	V
V_{IH}	Input High Voltage for I/O Ports	—	—	$0.8V_{DD}$	—	V_{DD}	V
I_{OL}	Sink Current for I/O Ports	3V	$V_{OL}=0.1V_{DD}$	16	32	—	mA
I_{OH}	Source Current for I/O Ports	3V	$V_{OH}=0.9V_{DD}$, SLEDCn[m+1:m]=00B (n=0, 1; m=0, 2, 4, 6)	-0.7	-1.5	—	mA
		3V	$V_{OH}=0.9V_{DD}$, SLEDCn[m+1:m]=01B (n=0, 1; m=0, 2, 4, 6)	-1.3	-2.5	—	
		3V	$V_{OH}=0.9V_{DD}$, SLEDCn[m+1:m]=10B (n=0, 1; m=0, 2, 4, 6)	-1.8	-3.6	—	
		3V	$V_{OH}=0.9V_{DD}$, SLEDCn[m+1:m]=11B (n=0, 1; m=0, 2, 4, 6)	-4	-8	—	
R_{PH}	Pull-high Resistance for I/O Ports (Note)	3V	LVPU=0, PxPU=FFH (Px: PA, PB, PC, PD)	20	60	100	kΩ
		3V	LVPU=1, PxPU=FFH (Px: PA, PB, PC, PD)	6.67	15.00	23.00	kΩ
I_{LEAK}	Input Leakage Current	3V	$V_{IN}=V_{DD}$ or $V_{IN}=V_{SS}$	—	—	±1	μA
t_{TCK}	TM Clock Input Pin Minimum Pulse Width	—	—	0.3	—	—	μs
t_{INT}	External Interrupt Minimum Pulse Width	—	—	10	—	—	μs

Note: The R_{PH} internal pull-high resistance value is calculated by connecting to ground and enabling the input pin with a pull-high resistor and then measuring the pin current at the specified supply voltage level. Dividing the voltage by this measured current provides the R_{PH} value.

Memory Characteristics

Ta=-40°C~85°C, unless otherwise specified

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
Flash Program Memory							
t_{FER}	ROM Erase Time	—	—	2.273	2.500	2.778	ms
	IAP Erase Time	—	FWERTS=0	—	3.2	3.9	ms
		—	FWERTS=1	—	3.7	4.5	
t_{FWR}	ROM Write Time	—	—	1.364	1.500	1.667	ms
	IAP Write Time	—	FWERTS=0	—	2.2	2.7	ms
		—	FWERTS=1	—	3.0	3.6	
E_P	Cell Endurance	—	—	100K	—	—	E/W
t_{RETD}	ROM Data Retention Time	—	Ta=25°C	—	40	—	Year
t_{ACTV}	ROM Activation Time – Wake-up from IDLE/SLEEP Mode	—	—	32	—	64	μs

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
Data EEPROM Memory							
t_{EERD}	EEPROM Read Time	—	—	—	—	4	t_{SYS}
t_{EEWR}	EEPROM Write Time (Byte Mode)	—	EWERTS=0	—	5.4	6.6	ms
		—	EWERTS=1	—	6.7	8.1	ms
	EEPROM Write Time (Page Mode)	—	EWERTS=0	—	2.2	2.7	ms
		—	EWERTS=1	—	3.0	3.6	ms
t_{EEER}	EEPROM Erase Time	—	EWERTS=0	—	3.2	3.9	ms
		—	EWERTS=1	—	3.7	4.5	ms
E_P	Cell Endurance	—	—	100K	—	—	E/W
t_{RETD}	Data Retention Time	—	Ta=25°C	—	40	—	Year
RAM Data Memory							
V_{DR}	RAM Data Retention Voltage	—	—	1	—	—	V

Note: 1. "E/W" means Erase/Write times.
2. The ROM activation time t_{ACTV} should be added when calculating the total system start-up time of a wake-up from the IDLE/SLEEP mode.

A/D Converter Electrical Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
AV_{DD}	Operating Voltage	—	—	1.8	—	3.6	V
V_{ADI}	Input Voltage	—	—	0	—	V_{REF}	V
V_{REF}	Reference Voltage	—	—	1.2	—	AV_{DD}	V
N_R	Resolution	—	—	—	—	12	Bit
DNL	Differential Non-linearity	1.8V	SAINS[3:0]=0000B, SAVRS[1:0]=01B, V_{REF}=AV_{DD}, t_{ADCK}=2μs	-4	—	+4	LSB
		2V	SAINS[3:0]=0000B, SAVRS[1:0]=01B, V_{REF}=AV_{DD}, t_{ADCK}=1μs				
		1.8V	SAINS[3:0]=0000B, SAVRS[1:0]=01B, V_{REF}=AV_{DD}, t_{ADCK}=10μs	-3	—	+3	LSB
		3V	SAINS[3:0]=0000B, SAVRS[1:0]=01B, V_{REF}=AV_{DD}, t_{ADCK}=0.5μs				
			SAINS[3:0]=0000B, SAVRS[1:0]=01B, V_{REF}=AV_{DD}, t_{ADCK}=10μs				

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
INL	Integral Non-linearity	1.8V	SAINS[3:0]=0000B, SAVRS[1:0]=01B, V_{REF}=AV_{DD}, t_{ADCK}=2.0μs	-4	—	+4	LSB
			SAINS[3:0]=0000B, SAVRS[1:0]=01B, V_{REF}=AV_{DD}, t_{ADCK}=10μs				
		2V	SAINS[3:0]=0000B, SAVRS[1:0]=01B, V_{REF}=AV_{DD}, t_{ADCK}=1μs				
		3V	SAINS[3:0]=0000B, SAVRS[1:0]=01B, V_{REF}=AV_{DD}, t_{ADCK}=0.5μs				
			SAINS[3:0]=0000B, SAVRS[1:0]=01B, V_{REF}=AV_{DD}, t_{ADCK}=10μs				
I_{ADC}	Additional Current for A/D Converter Enable	1.8V	No load, t_{ADCK}=2.0μs	—	280	400	μA
		3V	No load, t_{ADCK}=0.5μs	—	450	600	
t_{ADCK}	Clock Period	—	1.8V≤AV_{DD}<2.0V	2.0	—	10.0	μs
			2.0V≤AV_{DD}≤3.6V	0.5	—	10.0	
t_{ADS}	Sampling Time	—	—	—	4	—	t_{ADCK}
t_{ADC}	Conversion Time (Includes A/D Sample and Hold Time)	—	—	—	16	—	t_{ADCK}
t_{ON2ST}	A/D Converter On-to-Start Time	—	—	4	—	—	μs
I_{PGA}	Additional Current for PGA Enable	2.2V	No load, PGAIS=1, PGAGS[1:0]=01B	—	250	500	μA
		3V	No load, PGAIS=1, PGAGS[1:0]=01B	—	300	600	μA
V_{OR}	PGA Maximum Output Voltage Range	2.2V	—	AV_{SS} +0.1	—	AV_{DD} -0.1	V
		3V	—	AV_{SS} +0.1	—	AV_{DD} -0.1	V
V_{VR}	Fix Voltage Output of PGA	2.2V~ 3.6V	V_{RI}=V_{BGREF} (PGAIS=1)	-1%	2	+1%	V
		3.2V~ 3.6V	V_{RI}=V_{BGREF} (PGAIS=1)	-1%	3	+1%	V
V_{IR}	PGA Input Voltage Range	3V	Gain=1, PGAIS=0, Relative gain, Gain error <±5%	AV_{SS} +0.1	—	AV_{DD} -1.4	V
V_{OS_PGA}	PGA Input Offset Voltage	3V	—	-15	—	+15	mV

LVD/LVR Electrical Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
V_{LVR}	Low Voltage Reset Voltage	—	LVR enable, voltage select 1.7V	-5%	1.7	+5%	V
			LVR enable, voltage select 1.9V	-5%	1.9	+5%	
			LVR enable, voltage select 2.55V	-3%	2.55	+3%	
			LVR enable, voltage select 3.15V	-3%	3.15	+3%	

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
V_{LVD}	Low Voltage Detector Voltage	—	LVD enable, voltage select 1.8V	-5%	1.8	+5%	V
			LVD enable, voltage select 2.0V		2.0		
			LVD enable, voltage select 2.4V		2.4		
			LVD enable, voltage select 2.7V		2.7		
			LVD enable, voltage select 3.0V		3.0		
			LVD enable, voltage select 3.3V		3.3		
I_{LVRLVD}	Operating Current	3V	LVD enable, LVR enable, V_{LVR}=1.9V, V_{LVD}=2V	—	—	10	μA
t_{LVDS}	LVDO Stable Time	—	For LVR enable, VBGEN=0, LVD off → on	—	—	18	μs
			For LVR disable, VBGEN=0, LVD off → on	—	—	150	
t_{LVR}	Minimum Low Voltage Width to Reset	—	—	120	240	480	μs
t_{LVD}	Minimum Low Voltage Width to Interrupt	—	—	60	120	240	μs
I_{LVR}	Additional Current for LVR Enable	3V	LVD disable	—	—	14	μA
I_{LVD}	Additional Current for LVD Enable	3V	LVR disable	—	—	14	μA

Reference Voltage Electrical Characteristics

Ta=-40°C~85°C, unless otherwise specified

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
V_{DD}	Operating Voltage	—	—	1.8	—	3.6	V
V_{BGREF}	Bandgap Reference Voltage	1.8V~<2.2V	—	-10%	1.2	+10%	V
		2.2V~3.6V	—	-1%	1.2	+1%	
I_{BGREF}	Operating Current	3.6V	—	—	25	35	μA
PSRR	Power Supply Rejection Ratio	—	Ta=25°C, V_{RIPPLE}=1V_{P-P}, f_{RIPPLE}=100Hz	75	—	—	dB
En	Output Noise	—	Ta=25°C, No load current, f=0.1Hz~10Hz	—	300	—	μV_{RMS}
I_{SD}	Shutdown Current	—	VBGREN=0	—	—	0.1	μA
t_{START}	Startup Time	1.8V~3.6V	Ta=25°C	—	—	400	μs

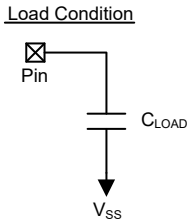
Note: The V_{BGREF} voltage is used as the A/D converter PGA input signal.

Comparator Electrical Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
I_{CMP}	Additional Current for Comparator Enable	3V	—	—	70	90	μA
I_{CSTB}	Power Down Current	3V	Comparator disable	—	—	0.1	μA
V_{CM}	Common Mode Voltage Range	3V	—	V_{SS}	—	V_{DD}-1	V
A_{OL}	Open Loop Gain	3V	—	60	80	—	dB
V_{HYS}	Hysteresis	3V	CMPHYEN=0	0	0	5	mV
		3V	CMPHYEN=1	5	24	30	mV
t_{RP}	Response Time	3V	With 100mV overdrive (1) (2)	—	200	400	ns

Note: 1. All measurement is under the condition where the comparator positive input voltage is equal to (V_{DD}-1)/2 and remains constant.
2. Measured with comparator one input pin at V_{CM}=(V_{DD}-1.0)/2 while the other pin input transition from V_{SS} to (V_{CM}+100mV) or from (V_{DD}-1V) to (V_{CM}-100mV).
3. Load Condition: C_{LOAD}=50pF.

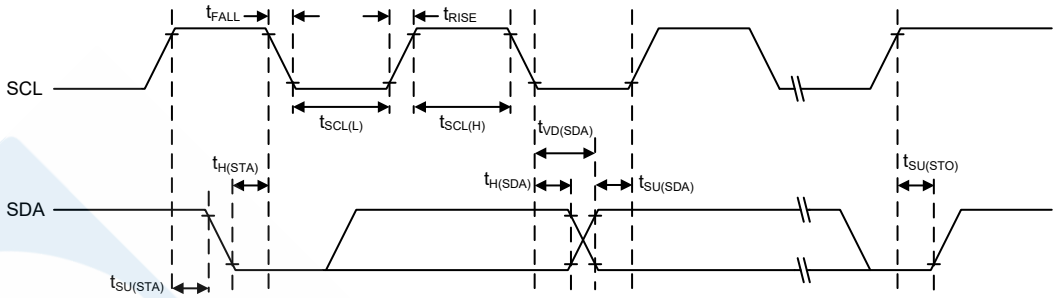


I²C Electrical Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
f_{I2C}	I²C Standard Mode (100kHz) f_{SYS} Frequency (Note)	—	No clock debounce	2	—	—	MHz
			2 system clock debounce	4	—	—	
			4 system clock debounce	4	—	—	
	I²C Fast Mode (400kHz) f_{SYS} Frequency (Note)	—	No clock debounce	4	—	—	MHz
			2 system clock debounce	8	—	—	
			4 system clock debounce	8	—	—	
f_{SCL}	SCL Clock Frequency	3V	Standard mode	—	—	100	kHz
			Fast mode	—	—	400	
$t_{SCL(H)}$	SCL Clock High Time	3V	Standard mode	3.5	—	—	μs
			Fast mode	0.9	—	—	
$t_{SCL(L)}$	SCL Clock Low Time	3V	Standard mode	3.5	—	—	μs
			Fast mode	0.9	—	—	
t_{FALL}	SCL and SDA Fall Time	3V	Standard mode	—	—	1.3	μs
			Fast mode	—	—	0.34	
t_{RISE}	SCL and SDA Rise Time	3V	Standard mode	—	—	1.3	μs
			Fast mode	—	—	0.34	
$t_{SU(SDA)}$	SDA Data Setup Time	3V	Standard mode	0.25	—	—	μs
			Fast mode	0.1	—	—	
$t_{H(SDA)}$	SDA Data Hold Time	3V	—	0.1	—	—	μs
$t_{VD(SDA)}$	SDA Data Valid Time	3V	—	—	—	0.6	μs
$t_{SU(STA)}$	Start Condition Setup Time	3V	Standard mode	3.5	—	—	μs
			Fast mode	0.6	—	—	
$t_{H(STA)}$	Start Condition Hold Time	3V	—	0.6	—	—	μs
$t_{SU(STO)}$	Stop Condition Setup Time	3V	Standard mode	3.5	—	—	μs
			Fast mode	0.6	—	—	

Note: Using the debounce function can make the transmission more stable and reduce the probability of communication failure due to interference.



I²C Timing Diagram

Software Controlled LCD Driver Electrical Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
I_{BIAS}	Bias Current	3V	ISEL[1:0]=00B	5.81	8.30	10.79	μA
			ISEL[1:0]=01B	11.62	16.60	21.58	
			ISEL[1:0]=10B	35	50	65	
			ISEL[1:0]=11B	70	100	130	
V_{SCOM}	V_{DD}×2/3 Voltage for LCD SCOM/SSEG Output	2.2V~3.6V	No load	0.305V_{DD}	0.330V_{DD}	0.355V_{DD}	V
	V_{DD}×1/3 Voltage for LCD SCOM/SSEG Output	2.2V~3.6V	No load	0.31V_{DD}	0.33V_{DD}	0.35V_{DD}	V

RF Electrical Characteristics

Ta=25°C, V_{DD}=3.3V, f_{XTAL}=16MHz, GFSK modulation with matching circuit and low/high pass filter, RF output is powered by V_{DD} (3.3V), unless otherwise specified

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
T_{OP}	Operating Temperature	—	-40	—	85	°C
V_{DD}	Supply Voltage	—	1.8	3.3	3.6	V
Digital I/Os						
V_{IH}	High Level Input Voltage	—	0.7×V_{DD}	—	V_{DD}	V
V_{IL}	Low Level Input Voltage	—	0	—	0.3×V_{DD}	V
V_{OH}	High Level Output Voltage	I_{OH}=-5mA	0.8×V_{DD}	—	V_{DD}	V
V_{OL}	Low Level Output Voltage	I_{OL}=5mA	0	—	0.2×V_{DD}	V
Current Consumption						
I_{Sleep}	Deep Sleep Mode Current Consumption	—	—	0.4	1.0	μA
I_{IL}	Idle Mode Current Consumption	LIRC on, X'tal off	—	1.6	—	μA
$I_{LightSleep}$	Light Sleep Mode Current Consumption	X'tal on	—	0.6	—	mA
$I_{Standby}$	Standby Mode Current Consumption @ 315/433MHz	X'tal on, Synthesizer on	—	3.9	—	mA
	Standby Mode Current Consumption @ 868/915MHz		—	3.9	—	

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
I_{RX} or I_{TX}	433MHz Band Current Consumption	RX mode @ 2kbps	—	5.8	—	mA
		RX mode @ 250kbps	—	6.4	—	
		TX mode @ 0dBm P_{OUT}	—	19	—	
		TX mode @ 10dBm P_{OUT}	—	33	—	
		TX mode @ 13dBm P_{OUT}	—	43		
		TX mode @ 19dBm P_{OUT}	—	71	—	
	868MHz Band Current Consumption	RX mode @ 2kbps	—	6.8	—	mA
		RX mode @ 250kbps	—	7.5	—	
		TX mode @ 0dBm P_{OUT}	—	19	—	
		TX mode @ 10dBm P_{OUT}	—	35	—	
		TX mode @ 13dBm P_{OUT}	—	47		
		TX mode @ 19dBm P_{OUT}	—	88	—	
R_{PH}	Pull-high Resistance for I/O Ports	3.3V	—	33	—	kΩ
RF Characteristics						
f_{RF}	RF Frequency Band	315MHz band	—	315	—	MHz
		433MHz band	—	433.92	—	
		470~510MHz band	—	490	—	
		868MHz band	—	868.3	—	
		915MHz band	—	915	—	
DR	Data Rate	OOK modulation	0.5	—	20	kbps
		GFSK modulation	2	—	250	
f_{LIRC}	RF Internal Low Frequency RC Oscillator	3.3V	-10%	32.768	+10%	kHz
Transmitter						
P_{OUT}	TX Output Power	433MHz band	0	—	20	dBm
		868MHz band	0	—	20	
S.E.$_{TX}$	TX Spurious Emission (P_{OUT}=10dBm)	f<1GHz	—	—	-36	dBm
		47MHz<f<74MHz	—	—	-54	
		87.5MHz<f<118MHz				
		174MHz<f<230MHz				
		470MHz<f<862MHz				
		2nd, 3rd Harmonic	—	—	-30	
Receiver						
$t_{ST,RX}$	RX Settling Time	Light Sleep mode to RX mode	—	150	—	μs
P_{Sens}	433MHz RX Sensitivity @ BER=0.1%	2kbps (f_{DEV}=8kHz)	—	-117	—	dBm
		10kbps (f_{DEV}=40kHz)	—	-110	—	
		50kbps (f_{DEV}=18.75kHz)	—	-107	—	
		125kbps (f_{DEV}=46.875kHz)	—	-103	—	
		250kbps (f_{DEV}=93.75kHz)	—	-100	—	
	868MHz RX Sensitivity @ BER=0.1%	2kbps (f_{DEV}=8kHz)	—	-116	—	dBm
		10kbps (f_{DEV}=40kHz)	—	-110	—	
		50kbps (f_{DEV}=18.75kHz)	—	-106	—	
		125kbps (f_{DEV}=46.875kHz)	—	-103	—	
		250kbps (f_{DEV}=93.75kHz)	—	-100	—	
$P_{IN,max}$	Maxmum Input Power	@ BER<0.1%	—	—	10	dBm
IR	Image Rejection	—	—	25	—	dB
S.E.$_{RX}$	RX Spurious	25MHz~1GHz	—	—	-57	dBm
		Above 1GHz	—	—	-47	
	RSSI Range	AGC on	-110	—	-10	dBm

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
LO Characteristics						
f_{LO}	RF Frequency Coverage Range	315MHz band	290	—	335	MHz
		433MHz band	415	—	490	
		470~510MHz band	470	—	510	
		868MHz band	830	—	1000	
		915MHz band	870	—	1050	
f_{STEP}	LO Frequency Resolution	—	—	—	1	kHz
PN_{LO}	433MHz Phase Noise	@ 100kHz offset	—	-91	—	dBc/Hz
		@ 1MHz offset	—	-110	—	
	868MHz Phase Noise	@ 100kHz offset	—	-82	—	
		@ 1MHz offset	—	-105	—	
Crystal Oscillator						
f_{XTAL}	X'tal Frequency	—	—	16	—	MHz
ESR	X'tal Equivalent Series Resistance	—	—	—	100	Ω
C_{LOAD}	X'tal Capacitor Load	—	8	12	16	pF
TOL	X'tal Tolerance (Note)	—	-20	—	+20	ppm
t_{SU}	X'tal Startup Time	49US XO	—	1	—	ms
		3225SMD XO	—	2	—	

Note: When the data rate is 2kbps at 315/433.92MHz, an X'tal with a tolerance of ±10ppm should be used.
When the data rate is 2kbps at 868/915MHz, an X'tal with a tolerance of ±5ppm should be used.

RF SPI Characteristics

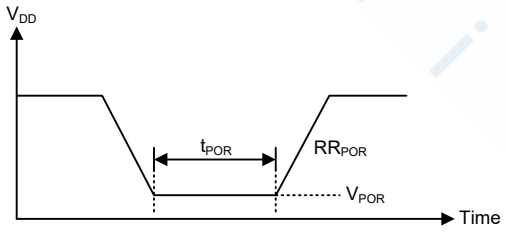
Ta=25°C

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
f_{SCK}	RF SCK Frequency	—	—	4	—	MHz
t_{SCKH}	RF SCK High Time	—	62.5	—	—	ns
t_{SCKL}	RF SCK Low Time	—	62.5	—	—	ns
t_{S_SDIO}	RF SDIO Input Setup Time	—	20	—	—	ns
t_{H_SDIO}	RF SDIO Input Hold Time	—	20	—	—	ns
t_{S_CSN}	RF CSN to SCK Active	—	30	—	—	ns
t_{H_CSN}	RF SCK Inactive to CSN Inactive	—	30	—	—	ns

Power-on Reset Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V_{DD}	Conditions				
V_{POR}	V_{DD} Start Voltage to Ensure Power-on Reset	—	—	—	—	100	mV
RR_{POR}	V_{DD} Rising Rate to Ensure Power-on Reset	—	—	0.035	—	—	V/ms
t_{POR}	Minimum Time for V_{DD} Stays at V_{POR} to Ensure Power-on Reset	—	—	1	—	—	ms



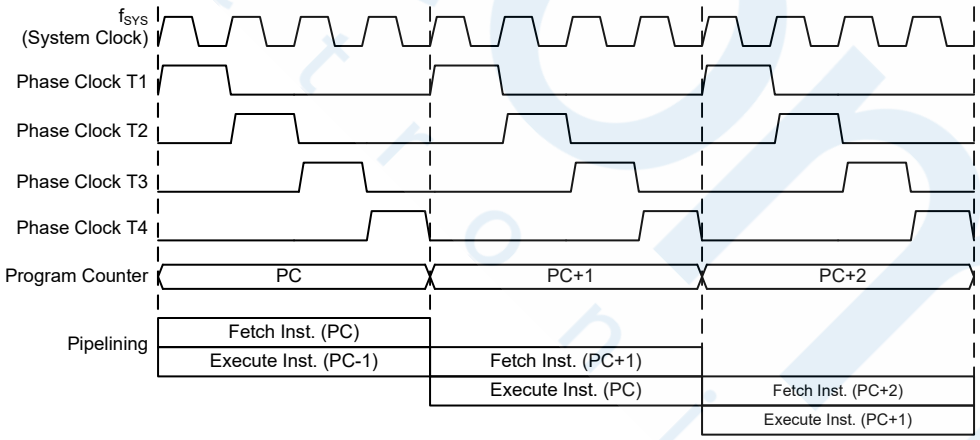
System Architecture

A key factor in the high-performance features of the Holtek range of microcontrollers is attributed to their internal system architecture. The device takes advantage of the usual features found within RISC microcontrollers providing increased speed of operation and enhanced performance. The pipelining scheme is implemented in such a way that instruction fetching and instruction execution are overlapped, hence instructions are effectively executed in one or two cycles for most of the standard or extended instructions respectively. The exceptions to these are branch or call instructions which need one more cycle. An 8-bit wide ALU is used in practically all instruction set operations, which carries out arithmetic operations, logic operations, rotation, increment, decrement, branch decisions, etc. The internal data path is simplified by moving data through the Accumulator and the ALU. Certain internal registers are implemented in the Data Memory and can be directly or indirectly addressed. The simple addressing methods of these registers along with additional architectural features ensure that a minimum of external components is required to provide a functional I/O and A/D control system with maximum reliability and flexibility. This makes the device suitable for low-cost, high-volume production for controller applications.

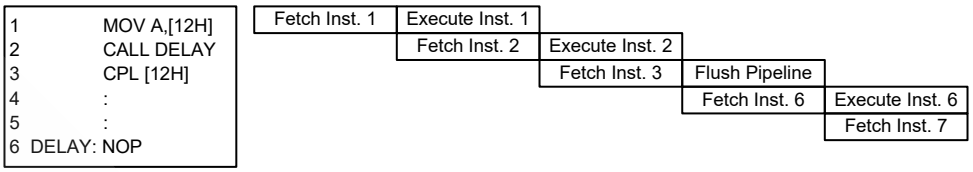
Clocking and Pipelining

The main system clock, derived from either an HXT, LXT, HIRC or LIRC oscillator is subdivided into four internally generated non-overlapping clocks, T1~T4. The Program Counter is incremented at the beginning of the T1 clock during which time a new instruction is fetched. The remaining T2~T4 clocks carry out the decoding and execution functions. In this way, one T1~T4 clock cycle forms one instruction cycle. Although the fetching and execution of instructions takes place in consecutive instruction cycles, the pipelining structure of the microcontroller ensures that instructions are effectively executed in one instruction cycle. The exception to this are instructions where the contents of the Program Counter are changed, such as subroutine calls or jumps, in which case the instruction will take one more instruction cycle to execute.

For instructions involving branches, such as jump or call instructions, two machine cycles are required to complete instruction execution. An extra cycle is required as the program takes one cycle to first obtain the actual jump or call address and then another cycle to actually execute the branch. The requirement for this extra cycle should be taken into account by programmers in timing sensitive applications.



System Clocking and Pipelining



Instruction Fetching

Program Counter

During program execution, the Program Counter is used to keep track of the address of the next instruction to be executed. It is automatically incremented by one each time an instruction is executed except for instructions, such as "JMP" or "CALL" that demands a jump to a non-consecutive Program Memory address. For the device with a Program Memory capacity in excess of 8K words, the Program Memory high byte address must be setup by selecting a certain program memoey bank which is implemented using the program memory bank pointer bit, PBP0. Only the lower 8 bits, known as the Program Counter Low Register, are directly addressable by the application program.

When executing instructions requiring jumps to non-consecutive addresses such as a jump instruction, a subroutine call, interrupt or reset, etc., the microcontroller manages program control by loading the required address into the Program Counter. For conditional skip instructions, once the condition has been met, the next instruction, which has already been fetched during the present instruction execution, is discarded and a dummy cycle takes its place while the correct instruction is obtained.

Program Counter	
High Byte	**Low Byte (PCL)**
PBP0, PC12~PC8	PCL7~PCL0

Program Counter

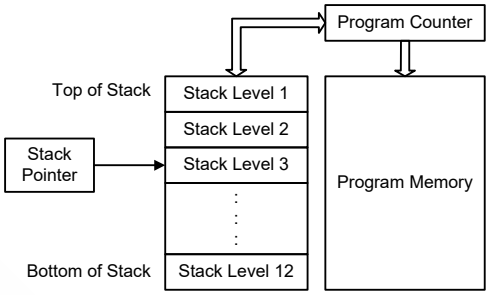
The lower byte of the Program Counter, known as the Program Counter Low register or PCL, is available for program control and is a readable and writeable register. By transferring data directly into this register, a short program jump can be executed directly. However, as only this low byte is available for manipulation, the jumps are limited to the present page of memory that is 256 locations. When such program jumps are executed it should also be noted that a dummy cycle will be inserted. Manipulating the PCL register may cause program branching, so an extra cycle is needed to pre-fetch.

Stack

This is a special part of the memory which is used to save the contents of the Program Counter only. The stack is organized into 12 levels and neither part of the data nor part of the program space, and is neither readable nor writeable. The activated level is indexed by the Stack Pointer, and is neither readable nor writeable. At a subroutine call or interrupt acknowledge signal, the contents of the Program Counter are pushed onto the stack. At the end of a subroutine or an interrupt routine, signaled by a return instruction, RET or RETI, the Program Counter is restored to its previous value from the stack. After a device reset, the Stack Pointer will point to the top of the stack.

If the stack is full and an enabled interrupt takes place, the interrupt request flag will be recorded but the acknowledge signal will be inhibited. When the Stack Pointer is decremented, by RET or RETI, the interrupt will be serviced. This feature prevents stack overflow allowing the programmer to use the structure more easily. However, when the stack is full, a CALL subroutine instruction can still be executed which will result in a stack overflow. Precautions should be taken to avoid such cases which might cause unpredictable program branching.

If the stack is overflow, the first Program Counter save in the stack will be lost.



Arithmetic and Logic Unit – ALU

The arithmetic-logic unit or ALU is a critical area of the microcontroller that carries out arithmetic and logic operations of the instruction set. Connected to the main microcontroller data bus, the ALU receives related instruction codes and performs the required arithmetic or logical operations after which the result will be placed in the specified register. As these ALU calculation or operations may result in carry, borrow or other status changes, the status register will be correspondingly updated to reflect these changes. The ALU supports the following functions:

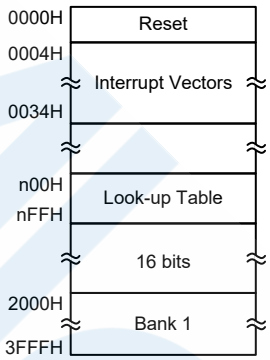
- Arithmetic operations:
  ADD, ADDM, ADC, ADCM, SUB, SUBM, SBC, SBCM, DAA,
  LADD, LADDM, LADC, LADCM, LSUB, LSUBM, LSBC, LSBCM, LDAA
- Logic operations:
  AND, OR, XOR, ANDM, ORM, XORM, CPL, CPLA,
  LAND, LOR, LXOR, LANDM, LORM, LXORM, LCPL, LCPLA
- Rotation:
  RRA, RR, RRCA, RRC, RLA, RL, RLCA, RLC,
  LRR, LRRA, LRRCA, LRRC, LRLA, LRL, LRLCA, LRLC
- Increment and Decrement:
  INCA, INC, DECA, DEC,
  LINCA, LINC, LDECA, LDEC
- Branch decision:
  JMP, SZ, SZA, SNZ, SIZ, SDZ, SIZA, SDZA, CALL, RET, RETI,
  LSZ, LSZA, LSNZ, LSIZ, LSDZ, LSIZA, LSDZA

Flash Program Memory

The Program Memory is the location where the user code or program is stored. For this device the Program Memory is Flash type, which means it can be programmed and re-programmed a large number of times, allowing the user the convenience of code modification on the same device. By using the appropriate programming tools, the Flash device offer users the flexibility to conveniently debug and develop their applications while also offering a means of field programming and updating.

Structure

The Program Memory has a capacity of 16K×16 bits. The Program Memory is addressed by the Program Counter and also contains data, table information and interrupt entries. Table data, which can be arranged in any location within the Program Memory, is addressed by a separate table pointer register.



Program Memory Structure

Special Vectors

Within the Program Memory, certain locations are reserved for the reset and interrupts. The location 0000H is reserved for use by the device reset for program initialisation. After a device reset is initiated, the program will jump to this location and begin execution.

Look-up Table

Any location within the Program Memory can be defined as a look-up table where programmers can store fixed data. To use the look-up table, the table pointer must first be configured by placing the address of the look up data to be retrieved in the table pointer registers, TBLP and TBHP. These registers define the total address of the look-up table.

After setting up the table pointer, the table data can be retrieved from the Program Memory using the corresponding table read instruction such as “TABRD [m]” or “TABRDL [m]” respectively when the memory [m] is located in Sector 0. If the memory [m] is located in other sectors except Sector 0, the data can be retrieved from the program memory using the corresponding extended table read instruction such as “LTABRD [m]” or “LTABRDL [m]” respectively. When the instruction is executed, the lower order table byte from the Program Memory will be transferred to the user defined Data Memory register [m] as specified in the instruction. The higher order table data byte from the Program Memory will be transferred to the TBLH special register. Any unused bits in this transferred higher order byte will be read as “0”.

The accompanying diagram illustrates the addressing data flow of the look-up table.

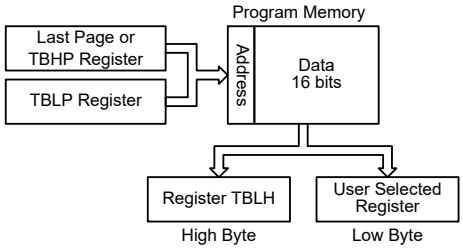


Table Program Example

The following example shows how the table pointer and table data is defined and retrieved from the microcontroller. This example uses raw table data located in the Program Memory which is stored there using the ORG statement. The value at this ORG statement is "1F00H" which is located in ROM Bank 1 and refers to the start address of the last page within the 16K words Program Memory of the device. The table pointer low byte register is set here to have an initial value of "06H". This will ensure that the first data read from the data table will be at the Program Memory address "3F06H" or 6 locations after the start of the last page. Note that the value for the table pointer is referenced to the specific address pointed by the TBLP and TBHP registers if the "TABRD [m]" or "LTABRD [m]" instruction is being used. The high byte of the table data which in this case is equal to zero will be transferred to the TBLH register automatically when the "TABRD [m]" or "LTABRD [m]" instruction is executed.

Because the TBLH register is a read/write register and can be restored, care should be taken to ensure its protection if both the main routine and Interrupt Service Routine use table read instructions. If using the table read instructions, the Interrupt Service Routines may change the value of the TBLH and subsequently cause errors if used again by the main routine. As a rule it is recommended that simultaneous use of the table read instructions should be avoided. However, in situations where simultaneous use cannot be avoided, the interrupts should be disabled prior to the execution of any main routine table-read instructions. Note that all table related instructions require two instruction cycles to complete their operation.

Table Read Program Example

```
rombank 1 code1
ds .section 'data'
tempreg1 db ?     ; temporary register #1
tempreg2 db ?     ; temporary register #2
:
:
mov a,06h         ; initialise low table pointer - note that this address is
                  ; referenced
mov tblp,a        ; to the last page or the page that tbhp pointed
mov a,3Fh         ; initialise high table pointer
mov tbhp,a
:
:
tabrd tempreg1    ; transfers value in table referenced by table pointer, data at
                  ; program memory address "3F06H" transferred to tempreg1 and TBLH
dec tblp          ; reduce value of table pointer by one
tabrd tempreg2    ; transfers value in table referenced by table pointer, data at
                  ; program memory address "3F05H" transferred to tempreg2 and TBLH
                  ; in this example the data "1AH" is transferred to tempreg1 and
                  ; data "0FH" to register tempreg2 the value "00H" will be
                  ; transferred to the high byte register TBLH
:
:
```

```
code1 .section 'code'
org 1F00h         ; sets initial address of program memory
dc 00Ah, 00Bh, 00Ch, 00Dh, 00Eh, 00Fh, 01Ah, 01Bh
:
```

In Circuit Programming – ICP

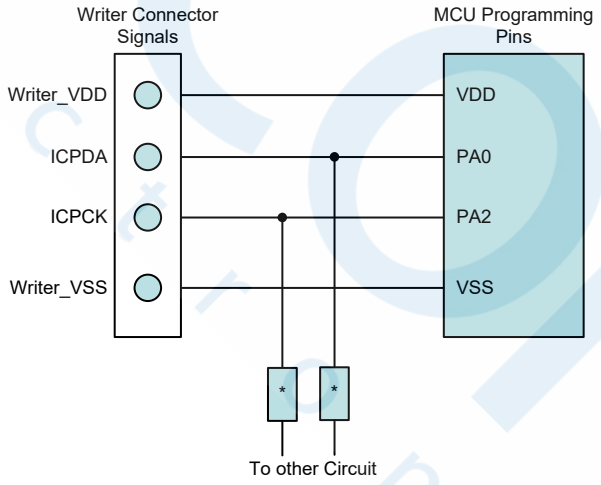
The provision of Flash type Program Memory provides the user with a means of convenient and easy upgrades and modifications to their programs on the same device.

As an additional convenience, a means of programming the microcontroller in-circuit has provided using a 4-pin interface. This provides manufacturers with the possibility of manufacturing their circuit boards complete with a programmed or un-programmed microcontroller, and then programming or upgrading the program at a later stage. This enables product manufacturers to easily keep their manufactured products supplied with the latest program releases without removal and re-insertion of the device.

Holtek Writer Pins	MCU Programming Pins	Pin Description
ICPDA	PA0	Programming Serial Data/Address
ICPCK	PA2	Programming Clock
VDD	VDD	Power Supply
VSS	VSS	Power Ground

The Program Memory can be programmed serially in-circuit using this 4-wire interface. Data is downloaded and uploaded serially on a single pin with an additional line for the clock. Two additional lines are required for the power supply. The technical details regarding the in-circuit programming of the device are beyond the scope of this document and will be supplied in supplementary literature.

During the programming process, the user can take care of the ICPDA and ICPCK pins for data and clock programming purposes to ensure that no other outputs are connected to these two pins.



Note: * may be resistor or capacitor. The resistance of * must be greater than 1kΩ or the capacitance of * must be less than 1nF.

On-Chip Debug Support – OCDS

There is an EV chip named BC66V3663 which is used to emulate the real MCU device named BC66F3663. The EV chip device also provides an "On-Chip Debug" function to debug the real MCU device during the development process. The EV chip and the real MCU device are almost functionally compatible except for "On-Chip Debug" function. Users can use the EV chip device to

emulate the real chip device behavior by connecting the OCDSDA and OCDSCK pins to the Holtek HT-IDE development tools. The OCDSDA pin is the OCDS Data/Address input/output pin while the OCDSCK pin is the OCDS clock input pin. When users use the EV chip for debugging, other functions which are shared with the OCDSDA and OCDSCK pins in the device will have no effect in the EV chip. However, the two OCDS pins which are pin-shared with the ICP programming pins are still used as the Flash Memory programming pins for ICP. For more detailed OCDS information, refer to the corresponding document named "Holtek e-Link for 8-bit MCU OCDS User's Guide".

Holtek e-Link Pins	EV Chip Pins	Pin Description
OCDSDA	OCDSDA	On-chip Debug Support Data/Address input/output
OCDSCK	OCDSCK	On-chip Debug Support Clock input
VDD	VDD	Power Supply
VSS	VSS	Power Ground

In Application Programming – IAP

Flash type Program Memory provides the user with a means of convenient and easy upgrades and modifications to their programs on the same device. The provision of the IAP function offers users the convenience of Flash Memory multi-programming features. The convenience of the IAP function is that it can execute the updated program procedure using its internal firmware, without requiring an external Program Writer or PC. In addition, the IAP interface can also be any type of communication protocol, such as UART, using I/O pins. Regarding the internal firmware, the user can select versions provided by Holtek or create their own. The following section illustrates the procedures regarding how to implement the IAP firmware.

Flash Memory Read/Write Size

The Flash memory Erase and Write operations are carried out in a page format while the Read operation is carried out in a word format. The page size and write buffer size are both assigned with a capacity of 32 words. Note that the Erase operation should be executed before the Write operation is executed.

When the Flash Memory Erase/Write Function is successfully enabled, the CFWEN bit will be set high. When the CFWEN bit is set high, the data can be written into the write buffer. The FWT bit is used to initiate the write process and then indicate the write operation status. This bit is set high by application programs to initiate a write process and will be cleared by hardware if the write process is finished.

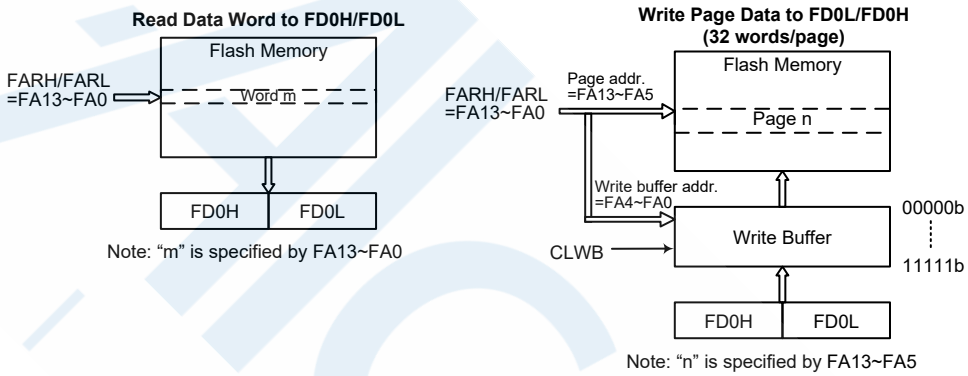
The Read operation can be carried out by executing a specific read procedure. The FRDEN bit is used to enable the read function and the FRD bit is used to initiate the read process by application programs and then indicate the read operation status. When the read process is finished, this bit will be cleared by hardware.

Operations	Format
Erase	32 words/page
Write	32 words/time
Read	1 word/time
Note: Page size=Write buffer size=32 words.	

IAP Operation Format

Page	FARH	FARL[7:5]	FARL[4:0]
0	0000 0000	000	Tag Address
1	0000 0000	001	
2	0000 0000	010	
3	0000 0000	011	
4	0000 0000	100	
5	0000 0000	101	
6	0000 0000	110	
7	0000 0000	111	
8	0000 0001	000	
:	:	:	
510	0011 1111	110	
511	0011 1111	111	

Page Number and Address Selection



Flash Memory IAP Read/Write Structure

Write Buffer

The write buffer is used to store the written data temporarily when executing the write operation. The Write Buffer can be filled with written data after the Flash Memory Erase/Write Function has been successfully enabled by executing the Flash Memory Erase/Write Function Enable procedure. The write buffer can be cleared by configuring the CLWB bit in the FC2 register. The CLWB bit can be set high to enable the Clear Write Buffer procedure. When the procedure is finished this bit will be cleared to low by the hardware. It is recommended that the write buffer should be cleared by setting the CLWB bit high before the write buffer is used for the first time or when the data in the write buffer is updated.

The write buffer size is 32 words corresponding to a page. The write buffer address is mapped to a specific flash memory page specified by the memory address bits, FA13~FA5. The data written into the FD0L and FD0H registers will be loaded into the write buffer. When data is written into the high byte data register, FD0H, it will result in the data stored in the high and low byte data registers both being written into the write buffer. It will also cause the flash memory address to be incremented by one, after which the new address will be loaded into the FARH and FARL address registers. When the flash memory address reaches the page boundary, 11111b of a page with 32 words, the address will now not be incremented but will stop at the last address of the page. At this point a new page address should be specified for any other erase/write operations.

After a write process is finished, the write buffer will automatically be cleared by the hardware. Note that the write buffer should be cleared manually by the application program when the data written into the flash memory is incorrect in the data verification step. The data should again be written into the write buffer after the write buffer has been cleared when the data is found to be incorrect during the data verification step.

IAP Flash Program Memory Registers

There are two address registers, four 16-bit data registers and three control registers. All the registers are located in Sector 0. Read and Write operations to the Flash memory are carried out by 16-bit data operations using the address and data registers and the control register. Several registers control the overall operation of the internal Flash Program Memory. The address registers are named FARL and FARH, the data registers are named FDnL and FDnH and the control registers are named FC0, FC1 and FC2.

Register Name	Bit							
	7	6	5	4	3	2	1	0
FC0	CFWEN	FMOD2	FMOD1	FMOD0	FWPEN	FWT	FRDEN	FRD
FC1	D7	D6	D5	D4	D3	D2	D1	D0
FC2	—	—	—	—	—	—	FWERTS	CLWB
FARL	FA7	FA6	FA5	FA4	FA3	FA2	FA1	FA0
FARH	—	—	FA13	FA12	FA11	FA10	FA9	FA8
FD0L	D7	D6	D5	D4	D3	D2	D1	D0
FD0H	D15	D14	D13	D12	D11	D10	D9	D8
FD1L	D7	D6	D5	D4	D3	D2	D1	D0
FD1H	D15	D14	D13	D12	D11	D10	D9	D8
FD2L	D7	D6	D5	D4	D3	D2	D1	D0
FD2H	D15	D14	D13	D12	D11	D10	D9	D8
FD3L	D7	D6	D5	D4	D3	D2	D1	D0
FD3H	D15	D14	D13	D12	D11	D10	D9	D8

IAP Register List

- **FARL Register**

Bit	7	6	5	4	3	2	1	0
Name	FA7	FA6	FA5	FA4	FA3	FA2	FA1	FA0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **FA7~FA0**: Flash Memory Address bit 7 ~ bit 0

- **FARH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	FA13	FA12	FA11	FA10	FA9	FA8
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as "0"

Bit 5~0 **FA13~FA8**: Flash Memory Address bit 13 ~ bit 8

• **FD0L Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: The first Flash Memory data bit 7 ~ bit 0

Note that data written into the low byte data register FD0L will only be stored in the FD0L register and not loaded into the lower 8-bit write buffer.

• **FD0H Register**

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D15~D8**: The first Flash Memory data bit 15 ~ bit 8

Note that when the 8-bit data is written into the high byte data register FD0H, the whole 16 bits of data stored in the FD0H and FD0L registers will simultaneously be loaded into the 16-bit write buffer after which the contents of the Flash memory address register pair, FARH and FARL, will be incremented by one.

• **FD1L Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: The second Flash Memory data bit 7 ~ bit 0

• **FD1H Register**

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D15~D8**: The second Flash Memory data bit 15 ~ bit 8

• **FD2L Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: The third Flash Memory data bit 7 ~ bit 0

• **FD2H Register**

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: The third Flash Memory data bit 15 ~ bit 8

• **FD3L Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: The fourth Flash Memory data bit 7 ~ bit 0

• **FD3H Register**

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D15~D8**: The fourth Flash Memory data bit 15 ~ bit 8

• **FC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	CFWEN	FMOD2	FMOD1	FMOD0	FWPEN	FWT	FRDEN	FRD
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **CFWEN**: Flash Memory Erase/Write function enable control
0: Flash Memory erase/write function is disabled
1: Flash Memory erase/write function has been successfully enabled

When this bit is cleared to zero by application program, the Flash Memory erase/write function is disabled. Note that this bit cannot be set high by application programs. Writing "1" into this bit results in no action. This bit is used to indicate that the Flash Memory erase/write function status. When this bit is set high by hardware, it means that the Flash Memory erase/write function is enabled successfully. Otherwise, the Flash Memory erase/write function is disabled as the bit content is zero.

Bit 6~4 **FMOD2~FMOD0**: Flash Memory Mode selection
000: Write Mode
001: Page Erase Mode
010: Reserved
011: Read Mode
100: Reserved
101: Reserved
110: Flash Memory Erase/Write function Enable Mode
111: Reserved

These bits are used to select the Flash Memory operation modes. Note that the "Flash memory Erase/Write function Enable Mode" should first be successfully enabled before the Erase or Write Flash memory operation is executed.

Bit 3 **FWPEN**: Flash Memory Erase/Write function enable procedure trigger
0: Erase/Write function enable procedure is not triggered or procedure timer times out
1: Erase/Write function enable procedure is triggered and procedure timer starts to count

This bit is used to activate the flash memory Erase/Write function enable procedure and an internal timer. It is set by the application programs and then cleared to zero by the hardware when the internal timer times out. The correct patterns must be written into the FD1L/FD1H, FD2L/FD2H and FD3L/FD3H register pairs respectively as soon as possible after the FWPEN bit is set high.

Bit 2 **FWT**: Flash Memory write initiate control
0: Do not initiate Flash Memory write or indicating that a Flash Memory write process has completed
1: Initiate a Flash Memory write process
This bit is set by software and cleared to zero by the hardware when the Flash memory write process has completed.

Bit 1 **FRDEN**: Flash Memory read enabled bit
0: Flash Memory read disable
1: Flash Memory read enable
This is the Flash memory Read Enable bit which must be set high before any Flash memory read operations are carried out. Clearing this bit to zero will inhibit Flash memory read operations.

Bit 0 **FRD**: Flash Memory read control bit
0: Do not initiate Flash Memory read or indicating that a Flash Memory read process has completed
1: Initiate a Flash Memory read process
This bit is set by software and cleared to zero by the hardware when the Flash memory read process has completed.

Note: 1. The FWT, FRDEN and FRD bits cannot be set to "1" at the same time with a single instruction.
2. Ensure that the f_{SUB} clock is stable before executing the erase or write operation.
3. Note that the CPU will be stopped when a read, erase or write operation is successfully activated.
4. Ensure that the read, erase or write operation is totally complete before executing other operations.

• **FC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: Chip Reset Pattern
When a specific value of "55H" is written into this register, a reset signal will be generated to reset the whole chip.

• **FC2 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	FWERTS	CLWB
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1 **FWERTS**: Erase time and Write time selection
0: Erase time is 3.2ms (t_{FER}) / Write time is 2.2ms (t_{FWR})
1: Erase time is 3.7ms (t_{FER}) / Write time is 3.0ms (t_{FWR})

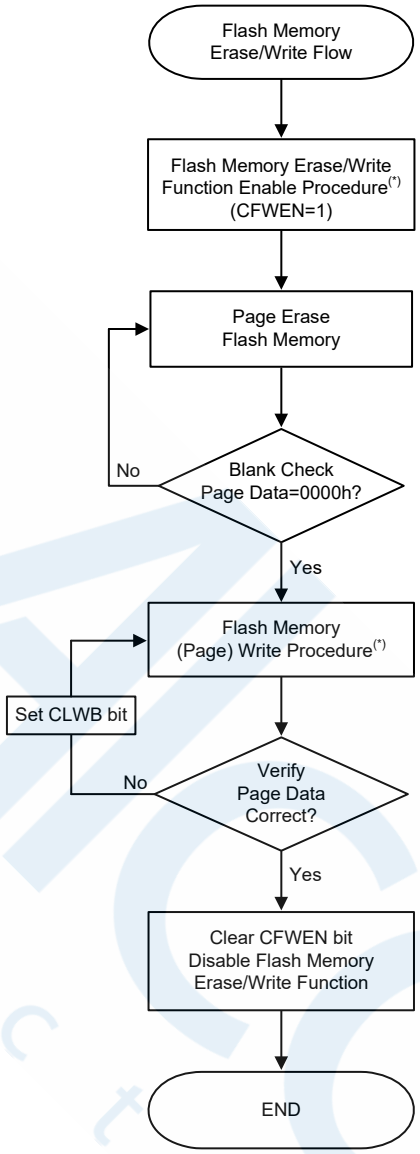
Bit 0 **CLWB**: Flash Memory Write buffer clear control
0: Do not initiate a Write Buffer Clear process or indicating that a Write Buffer Clear process has completed
1: Initiate a Write Buffer Clear process
This bit is set by software and cleared to zero by hardware when the Write Buffer Clear process has completed.

Flash Memory Erase/Write Flow

It is important to understand the Flash memory Erase/Write flow before the Flash memory contents are updated. Users can refer to the corresponding operation procedures when developing their IAP program to ensure that the flash memory contents are correctly updated.

Flash Memory Erase/Write Flow Descriptions

1. Activate the "Flash Memory Erase/Write function enable procedure" first. When the Flash Memory Erase/Write function is successfully enabled, the CFWEN bit in the FC0 register will automatically be set high by hardware. After this, Erase or Write operations can be executed on the Flash memory. Refer to the "Flash Memory Erase/Write Function Enable Procedure" for details.
2. Configure the flash memory address to select the desired erase page, tag address and then erase this page. For a page erase operation, set the FARL and FARH registers to specify the start address of the erase page, then write dummy data into the FD0H register to tag address. The current address will be internally incremented by one after each dummy data is written into the FD0H register. When the address reaches the page boundary, 11111b, the address will not be further incremented but stop at the last address of the page. Note that the write operation to the FD0H register is used to tag address, it must be implemented to determine which addresses to be erased.
3. Execute a Blank Check operation to ensure whether the page erase operation is successful or not. The "TABRD" instruction should be executed to read the flash memory contents and to check if the contents is 0000h or not. If the flash memory page erase operation fails, users should go back to Step 2 and execute the page erase operation again.
4. Write data into the specific page. Refer to the "Flash Memory Write Procedure" for details.
5. Execute the "TABRD" instruction to read the flash memory contents and check if the written data is correct or not. If the data read from the flash memory is different from the written data, it means that the page write operation has failed. The CLWB bit should be set high to clear the write buffer and then write the data into the specific page again if the write operation has failed.
6. Clear the CFWEN bit to disable the Flash Memory Erase/Write function enable mode if the current page Erase and Write operations are completed and no more pages need to be erased or written.



Flash Memory Erase/Write Flow

Note: * The Flash Memory Erase/Write Function Enable procedure and Flash Memory Write procedure will be described in the following sections.

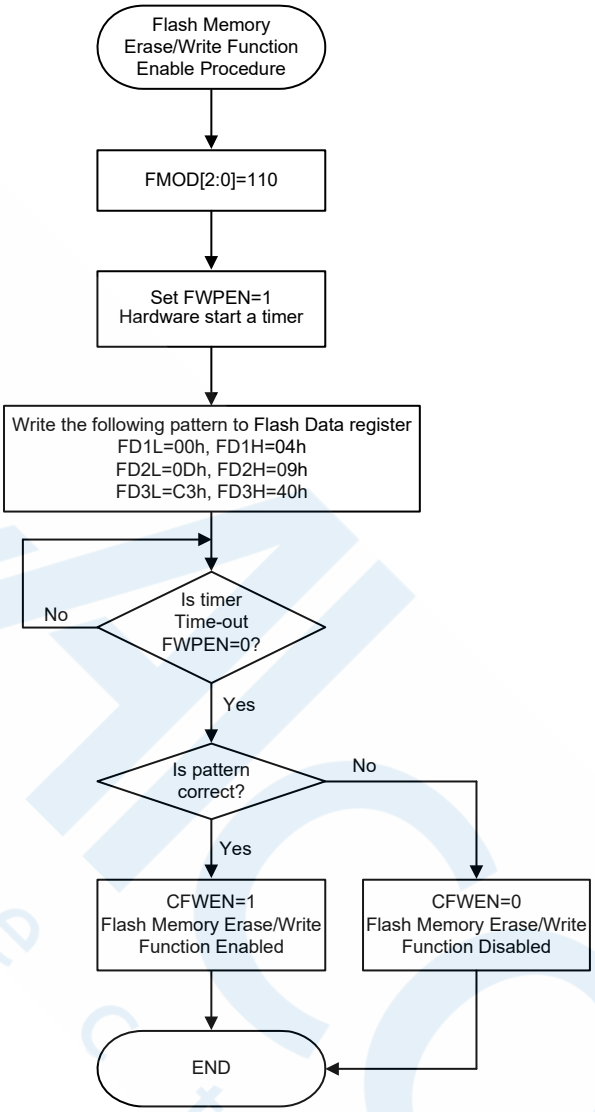
Flash Memory Erase/Write Function Enable Procedure

The Flash Memory Erase/Write Function Enable Mode is specially designed to prevent the flash memory contents from being wrongly modified. In order to allow users to change the Flash memory data using the IAP control registers, users must first enable the Flash memory Erase/Write function.

Flash Memory Erase/Write Function Enable Procedure Description

1. Write data "110" to the FMOD [2:0] bits in the FC0 register to select the Flash Memory Erase/ Write Function Enable Mode.
2. Set the FWPEN bit in the FC0 register to "1" to activate the Flash Memory Erase/Write Enable Function. This will also activate an internal timer.
3. Write the correct data pattern into the Flash data registers, FD1L~FD3L and FD1H~FD3H, as soon as possible after the FWPEN bit is set high. The data pattern to enable Flash memory erase/ write function is 00H, 0DH, C3H, 04H, 09H and 40H corresponding to the FD1L~FD3L and FD1H~FD3H registers respectively.
4. Once the timer has timed out, the FWPEN bit will automatically be cleared to zero by hardware regardless of the input data pattern.
5. If the written data pattern is incorrect, the Flash memory erase/write function will not be enabled successfully and the above steps should be repeated. If the written data pattern is correct, the Flash memory erase/write function will be enabled successfully.
6. Once the Flash memory erase/write function is enabled, the Flash memory contents can be updated by executing the page erase and write operations using the IAP control registers.

To disable the Flash memory erase/write function, the CFWEN bit in the FC0 register can be cleared. There is no need to execute the above procedure.



Flash Memory Erase/Write Function Enable Procedure

Flash Memory Write Procedure

After the Flash memory erase/write function has been successfully enabled as the CFWEN bit is set high, the data to be written into the flash memory can be loaded into the write buffer. The selected flash memory page data should be erased by properly configuring the IAP control registers before the data write procedure is executed.

The write buffer size is 32 words, known as a page, whose address is mapped to a specific flash memory page specified by the memory address bits, FA13~FA5. It is important to ensure that the page where the write buffer data is located is the same one which the memory address bits, FA13~FA5, specify.

Flash Memory Consecutive Write Description

The maximum amount of write data is 32 words for each write operation. The write buffer address will be automatically incremented by one when consecutive write operations are executed. The start address of a specific page should first be written into the FARL and FARH registers. Then the data word should first be written into the FD0L register and then the FD0H register. At the same time the write buffer address will be incremented by one and then the next data word can be written into the FD0L and FD0H registers for the next address without modifying the address register pair, FARH and FARL. When the write buffer address reaches the page boundary the address will not be further incremented but will stop at the last address of the page.

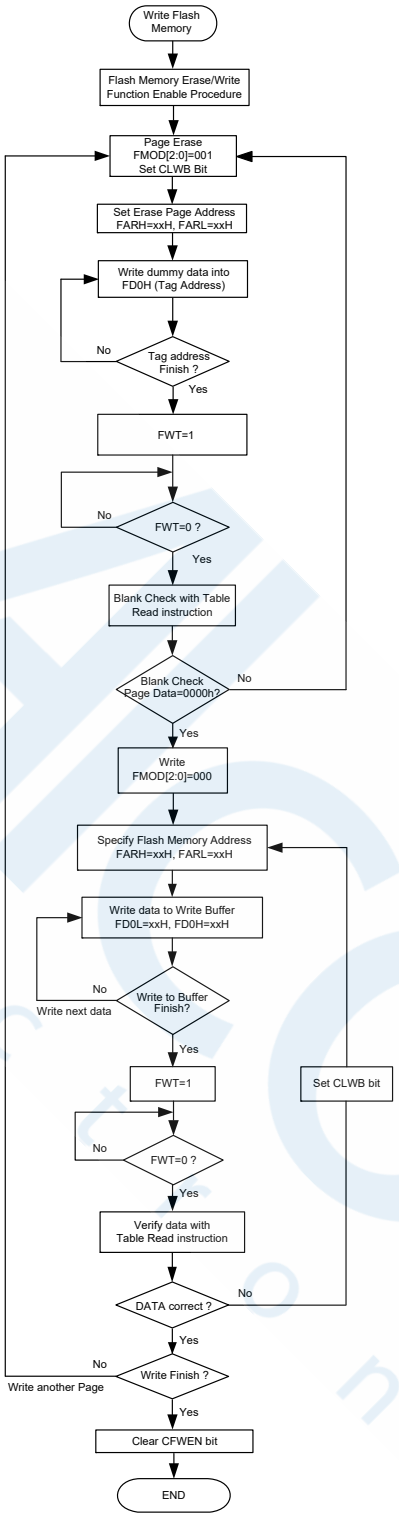
1. Activate the "Flash Memory Erase/Write function enable procedure". Check the CFWEN bit value and then execute the erase/write operations if the CFWEN bit is set high. Refer to the "Flash Memory Erase/Write function enable procedure" for more details.
2. Set the FMOD2~FMOD0 to "001" to select the erase operation and set the CLWB bit high to clear the write buffer. Set the FWT bit high to erase the desired page which is specified by the FARH and FARL registers and has been tagged address. Wait until the FWT bit goes low.
3. Execute a Blank Check operation using the table read instruction to ensure that the erase operation has successfully completed.

   Go to step 2 if the erase operation is not successful.

   Go to step 4 if the erase operation is successful.
4. Set the FMOD2~FMOD0 to "000" to select the write operation.
5. Setup the desired start address in the FARH and FARL registers. Write the desired data words consecutively into the FD0L and FD0H registers within a page as specified by their consecutive addresses. The maximum written data number is 32 words.
6. Set the FWT bit high to write the data words from the write buffer to the flash memory. Wait until the FWT bit goes low.
7. Verify the data using the table read instruction to ensure that the write operation has successfully completed.

   If the write operation has not successfully completed, set the CLWB bit high to clear the write buffer and then go to step 5.

   Go to step 8 if the write operation is successful.
8. Clear the CFWEN bit low to disable the Flash memory erase/write function.



Flash Memory Consecutive Write Procedure

Note: 1. When the erase or write operation is successfully activated, all CPU operations will temporarily cease.

2. It will take certain time for the FWT bit state changing from high to low in the erase or write operation, which can be selected by the FWERTS bit in the FC2 register.

Flash Memory Non-Consecutive Write Description

The main difference between Flash Memory Consecutive and Non-Consecutive Write operations is whether the data words to be written are located in consecutive addresses or not. If the data to be written is not located in consecutive addresses the desired address should be re-assigned after a data word is successfully written into the Flash Memory.

A two data word non-consecutive write operation is taken as an example here and described as follows:

1. Activate the "Flash Memory Erase/Write function enable procedure". Check the CFWEN bit value and then execute the erase/write operation if the CFWEN bit is set high. Refer to the "Flash Memory Erase/Write function enable procedure" for more details.
2. Set the FMOD2~FMOD0 to "001" to select the erase operation and set the CLWB bit high to clear the write buffer. Set the FWT bit high to erase the desired page which is specified by the FARH and FARL registers and has been tagged address. Wait until the FWT bit goes low.
3. Execute a Blank Check operation using the table read instruction to ensure that the erase operation has successfully completed.

   Go to step 2 if the erase operation is not successful.

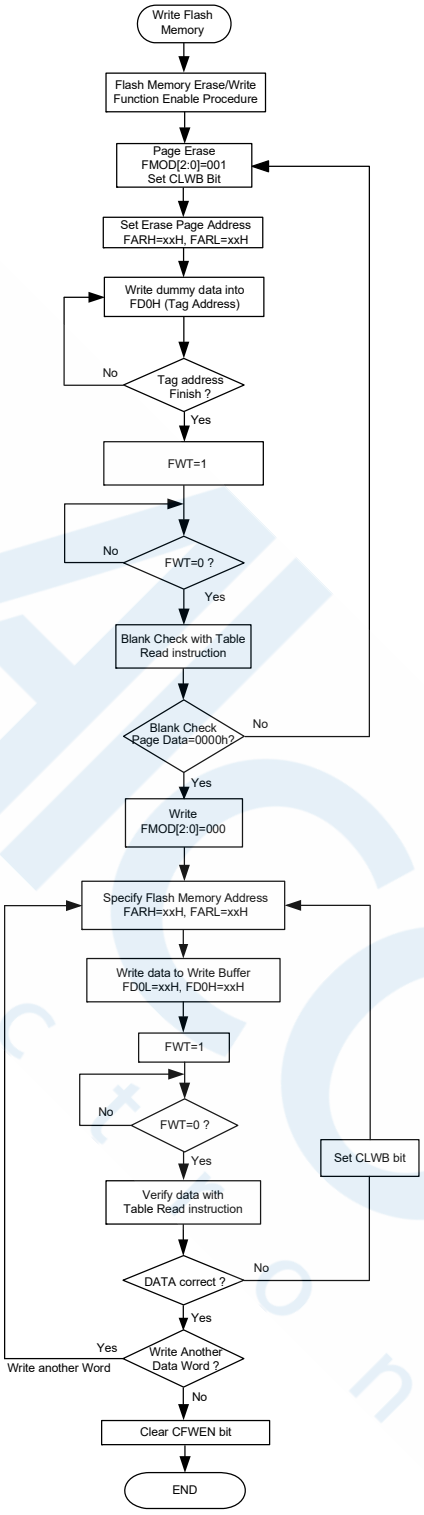
   Go to step 4 if the erase operation is successful.
4. Set the FMOD2~FMOD0 to "000" to select the write operation.
5. Setup the desired address ADDR1 in the FARH and FARL registers. Write the desired data word DATA1 first into the FD0L register and then into the FD0H register.
6. Set the FWT bit high to transfer the data word from the write buffer to the flash memory. Wait until the FWT bit goes low.
7. Verify the data using the table read instruction to ensure that the write operation has successfully completed.

   If the write operation has not successfully completed, set the CLWB bit high to clear the write buffer and then go to step 5.

   Go to step 8 if the write operation is successful.
8. Setup the desired address ADDR2 in the FARH and FARL registers. Write the desired data word DATA2 first into the FD0L register and then into the FD0H register.
9. Set the FWT bit high to transfer the data word from the write buffer to the flash memory. Wait until the FWT bit goes low.
10. Verify the data using the table read instruction to ensure that the write operation has successfully completed.

    If the write operation has not successfully completed, set the CLWB bit high to clear the write buffer and then go to step 8.

    Go to step 11 if the write operation is successful.
11. Clear the CFWEN bit low to disable the Flash memory erase/write function.



Flash Memory Non-Consecutive Write Procedure

Note: 1. When the erase or write operation is successfully activated, all CPU operations will temporarily cease.

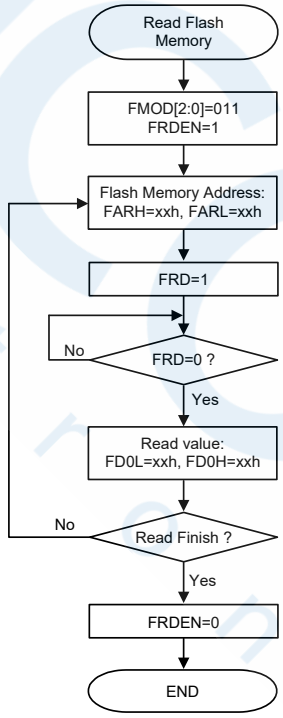
2. It will take certain time for the FWT bit state changing from high to low in the erase or write operation, which can be selected by the FWERTS bit in the FC2 register.

Important Points to Note for Flash Memory Write Operations

1. The "Flash Memory Erase/Write Function Enable Procedure" must be successfully activated before the Flash Memory erase/write operation is executed.
2. The Flash Memory erase operation is executed to erase a whole page.
3. The whole write buffer data will be written into the flash memory in a page format. The corresponding address cannot exceed the page boundary.
4. After the data is written into the flash memory the flash memory contents must be read out using the table read instruction, TABRD, and checked if it is correct or not. If the data written into the flash memory is incorrect, the write buffer should be cleared by setting the CLWB bit high and then write the data again into the write buffer. Then activate a write operation on the same flash memory page without erasing it. The data check, buffer clear and data re-write steps should be repeatedly executed until the data written into the flash memory is correct.
5. The system frequency should be setup to the maximum application frequency when data write and data check operations are executed using the IAP function.

Flash Memory Read Procedure

To activate the Flash Memory Read procedure, the FMOD field should be set to "011" to select the flash memory read mode and the FRDEN bit should be set high to enable the read function. The desired flash memory address should be written into the FARH and FARL registers and then the FRD bit should be set high. After this the flash memory read operation will be activated. The data stored in the specified address can be read from the data registers, FD0H and FD0L, when the FRD bit goes low. There is no need to first activate the Flash Memory Erase/Write Function Enable Procedure before the flash memory read operation is executed.



Flash Memory Read Procedure

Note: 1. When the read operation is successfully activated, all CPU operations will temporarily cease.
2. It will take a typical time of three instruction cycles for the FRD bit state changing from high to low.

Data Memory

The Data Memory is a volatile area of 8-bit wide RAM internal memory and is the location where temporary information is stored.

Categorised into two types, the first of these is an area of RAM, known as the Special Function Data Memory. These registers have fixed locations and are necessary for correct operation of the device. Many of these registers can be read from and written to directly under program control, however, some remain protected from user manipulation. The second area of Data Memory is known as the General Purpose Data Memory, which is reserved for general purpose use. All locations within this area are read and write accessible under program control.

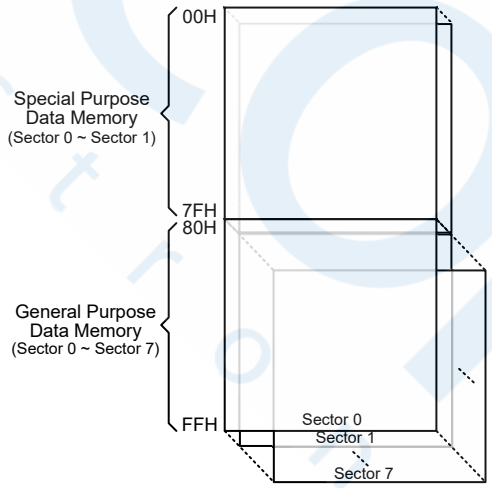
Switching between the different Data Memory sectors is achieved by properly setting the Memory Pointers to correct value when using the indirectly accessing method.

Structure

The Data Memory is subdivided into several sectors, all of which are implemented in 8-bit wide RAM. Each of the Data Memory sector is categorized into two types, the Special Purpose Data Memory and the General Purpose Data Memory. The address range of the Special Purpose Data Memory for the device is from 00H to 7FH while the General Purpose Data Memory address range is from 80H to FFH.

Special Purpose Data Memory	General Purpose Data Memory	
Located Sectors	Capacity	Sector: Address
0, 1	1K×8	0: 80H~FFH 1: 80H~FFH 2: 80H~FFH : 7: 80H~FFH

Data Memory Summary



Data Memory Structure

Data Memory Addressing

For the device that supports the extended instructions, there is no Bank Pointer for Data Memory addressing. The Bank Pointer, PBP, is only available for Program Mwmory. For Data Memory the desired sector is pointed by the MP1H or MP2H register and the certain Data Memory address in the selected sector is specified by the MP1L or MP2L register when using indirect addressing access.

Direct Addressing can be used in all sectors using the corresponding instruction which can address all available data memory space. For the accessed data memory which is located in any data memory sectors except Sector 0, the extended instructions can be used to access the data memory instead of using the indirect addressing access. The main difference between standard instructions and extended instructions is that the data memory address “m” in the extended instructions has 11 valid bits, the high byte indicates a sector and the low byte indicates a specific address within the sector.

General Purpose Data Memory

All microcontroller programs require an area of read/write memory where temporary data can be stored and retrieved for use later. It is this area of RAM memory that is known as General Purpose Data Memory. This area of Data Memory is fully accessible by the user programming for both reading and writing operations. By using the bit operation instructions individual bits can be set or reset under program control giving the user a large range of flexibility for bit manipulation in the Data Memory.

Special Purpose Data Memory

This area of Data Memory is where registers, necessary for the correct operation of the microcontroller, are stored. Most of the registers are both readable and writeable but some are protected and are readable only, the details of which are located under the relevant Special Function Register section. Note that for locations that are unused, any read instruction to these addresses will return the value “00H”.

	Sector 0	Sector 1
00H	IAR0	
01H	MP0	
02H	IAR1	
03H	MP1L	
04H	MP1H	
05H	ACC	
06H	PCL	
07H	TBLP	
08H	TBLH	
09H	TBHP	
0AH	STATUS	
0BH	PBP	
0CH	IAR2	
0DH	MP2L	
0EH	MP2H	
0FH	RSTFC	
10H	INTEG	
11H	INTC0	
12H	INTC1	
13H	INTC2	
14H	PA	
15H	PAC	
16H	PAPU	
17H	PAWU	
18H	LVDC	
19H	SCC	
1AH	WDTC	
1BH	TB0C	
1CH	HIRCC	
1DH	LVRC	
1EH	INTC3	
1FH		
20H	SADOL	
21H	SADOH	
22H	SADC0	
23H	SADC1	
24H	SADC2	
25H	PB	
26H	PBC	
27H	PBPU	
28H	CTMC0	
29H	CTMC1	
2AH	CTMDL	
2BH	CTMDH	
2CH	CTMAL	
2DH	CTMAH	
2EH	CTMRP	
2FH	STMC0	
30H	STMC1	
31H	STMDL	
32H	STMDH	
33H	STMAL	
34H	STMAH	
35H	STMRP	
36H	HXTC	
37H	PTM0C0	PTM1C0
38H	PTM0C1	PTM1C1
39H	PTM0DL	PTM1DL
3AH	PTM0DH	PTM1DH
3BH	PTM0AL	PTM1AL
3CH	PTM0AH	PTM1AH
3DH	PTM0RPL	PTM1RPL
3EH	PTM0RPH	PTM1RPH
3FH	CMPC	

	Sector 0	Sector 1
40H		EEC
41H	PC	
42H	PCC	
43H	PCPU	
44H	LXTC	
45H	SIMC0	
46H	SIMC1	
47H	SIMD	
48H	SIMC2/SIMA	
49H	SIMTOC	
4AH	VBGRC	
4BH	TB1C	
4CH	MFI0	U0SR
4DH	MFI1	U0CR1
4EH	MFI2	U0CR2
4FH	SLEDC0	U0CR3
50H	SLEDC1	BRDH0
51H	IFS0	BRDL0
52H	PD	UFCR0
53H	PDC	TXR_RXR0
54H	PDPU	RxCNT0
55H	ORMC	U1SR
56H	SLCDS3	U1CR1
57H	IFS1	U1CR2
58H	PDS1	U1CR3
59H		BRDH1
5AH	LVPUC	BRDL1
5BH	PAS0	UFCR1
5CH	PAS1	TXR_RXR1
5DH	PBS0	RxCNT1
5EH	PBS1	
5FH	PCS0	
60H	PCS1	
61H	PDS0	
62H	SLCDC0	
63H	SLCDS0	
64H	SLCDS1	
65H	SLCDS2	
66H	PSC0R	
67H	PSC1R	
68H	RSTC	
69H	FC0	
6AH	FC1	
6BH	FC2	
6CH	FARL	
6DH	FARH	
6EH	FD0L	
6FH	FD0H	
70H	FD1L	
71H	FD1H	
72H	FD2L	
73H	FD2H	
74H	FD3L	
75H	FD3H	
76H	EEAL	
77H	EEAH	
78H	EED	
79H		
7AH		
7BH		
7CH		
7DH		
7EH		
7FH		

: Unused, read as 00H

Special Purpose Data Memory Structure

Special Function Register Description

Most of the Special Function Register details will be described in the relevant functional section, however several registers require a separate description in this section.

Indirect Addressing Registers – IAR0, IAR1, IAR2

The Indirect Addressing Registers, IAR0, IAR1 and IAR2, although having their locations in normal RAM register space, do not actually physically exist as normal registers. The method of indirect addressing for RAM data manipulation uses these Indirect Addressing Registers and Memory Pointers, in contrast to direct memory addressing, where the actual memory address is specified. Actions on the IAR0, IAR1 and IAR2 registers will result in no actual read or write operation to these registers but rather to the memory location specified by their corresponding Memory Pointers, MP0, MP1L/MP1H or MP2L/MP2H. Acting as a pair, IAR0 and MP0 can together access data from Sector 0 while the IAR1 register together with the MP1L/MP1H register pair and IAR2 register together with the MP2L/MP2H register pair can access data from any Data Memory sector. As the Indirect Addressing Registers are not physically implemented, reading the Indirect Addressing Registers will return a result of "00H" and writing to the registers will result in no operation.

Memory Pointers – MP0, MP1L, MP1H, MP2L, MP2H

Five Memory Pointers, known as MP0, MP1L, MP1H, MP2L, MP2H, are provided. These Memory Pointers are physically implemented in the Data Memory and can be manipulated in the same way as normal registers providing a convenient way with which to address and track data. When any operation to the relevant Indirect Addressing Registers is carried out, the actual address that the microcontroller is directed to is the address specified by the related Memory Pointer. MP0, together with Indirect Addressing Register, IAR0, are used to access data from Sector 0, while MP1L/MP1H together with IAR1 and MP2L/MP2H together with IAR2 are used to access data from all sectors according to the corresponding MP1H or MP2H register. Direct Addressing can be used in all sectors using the extended instruction which can address all available data memory space.

The following example shows how to clear a section of four Data Memory locations already defined as locations adres1 to adres4.

Indirect Addressing Program Example

Example 1

```
data .section ´data´
adres1  db ?
adres2  db ?
adres3  db ?
adres4  db ?
block   db ?
code .section at 0 ´code´
org  00h
start:
     mov a, 04h                ; setup size of block
     mov block, a
     mov a, offset adres1      ; Accumulator loaded with first RAM address
     mov mp0, a                ; setup memory pointer with first RAM address
loop:
     clr IAR0                  ; clear the data at address defined by MP0
     inc mp0                   ; increment memory pointer
     sdz block                 ; check if last memory location has been cleared
     jmp loop
continue:
```

Example 2

```
data .section ´data´
adres1  db ?
adres2  db ?
adres3  db ?
adres4  db ?
block   db ?
code .section at 0 ´code´
org  00h
start:
     mov a, 04h             ; setup size of block
     mov block, a
     mov a, 01h             ; setup the memory sector
     mov mp1h, a
     mov a, offset adres1   ; Accumulator loaded with first RAM address
     mov mp1l, a            ; setup memory pointer with first RAM address
loop:
     clr IAR1               ; clear the data at address defined by MP1L
     inc mp1l               ; increment memory pointer MP1L
     sdz block              ; check if last memory location has been cleared
     jmp loop
continue:
```

The important point to note here is that in the example shown above, no reference is made to specific Data Memory addresses.

Direct Addressing Program Example using extended instructions

```
data .section ´data´
temp  db ?
code .section at 0 ´code´
org 00h
start:
     lmov a, [m]            ; move [m] data to acc
     lsub a, [m+1]          ; compare [m] and [m+1] data
     snz  c                 ; [m]>[m+1]?
     jmp  continue          ; no
     lmov a, [m]            ; yes, exchange [m] and [m+1] data
     mov  temp, a
     lmov a, [m+1]
     lmov [m], a
     mov  a, temp
     lmov [m+1], a
continue:
```

Note: Here “m” is a data memory address located in any data memory sectors. For example, m=1F0H, it indicates address 0F0H in Sector 1.

Program Memory Bank Pointer – PBP

For this device the Program Memory is divided into several banks. Selecting the required Program Memory area is achieved using the Program Memory Bank Pointer, PBP. The PBP register should be properly configured before the device executes the "Branch" operation using the "JMP" or "CALL" instruction. After that a jump to a non-consecutive Program Memory address which is located in a certain bank selected by the program memory bank pointer bits will occur.

• **PBP Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	PBP0
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

Bit 7~1 Unimplemented, read as "0"

Bit 0 **PBP0**: Program Memory Bank selection
0: Bank 0
1: Bank 1

Accumulator – ACC

The Accumulator is central to the operation of any microcontroller and is closely related with operations carried out by the ALU. The Accumulator is the place where all intermediate results from the ALU are stored. Without the Accumulator it would be necessary to write the result of each calculation or logical operation such as addition, subtraction, shift, etc., to the Data Memory resulting in higher programming and timing overheads. Data transfer operations usually involve the temporary storage function of the Accumulator; for example, when transferring data between one user-defined register and another, it is necessary to do this by passing the data through the Accumulator as no direct transfer between two registers is permitted.

Program Counter Low Byte Register – PCL

To provide additional program control functions, the low byte of the Program Counter is made accessible to programmers by locating it within the Special Purpose area of the Data Memory. By manipulating this register, direct jumps to other program locations are easily implemented. Loading a value directly into this PCL register will cause a jump to the specified Program Memory location, however, as the register is only 8-bit wide, only jumps within the current Program Memory page are permitted. When such operations are used, note that a dummy cycle will be inserted.

Look-up Table Registers – TBLP, TBHP, TBLH

These three special function registers are used to control operation of the look-up table which is stored in the Program Memory. TBLP and TBHP are the table pointers and indicate the location where the table data is located. Their value must be set before any table read commands are executed. Their value can be changed, for example using the "INC" or "DEC" instructions, allowing for easy table data pointing and reading. TBLH is the location where the high order byte of the table data is stored after a table read data instruction has been executed. Note that the lower order table data byte is transferred to a user defined location.

Option Memory Mapping Register – ORMC

The ORMC register is used to enable Option Memory Mapping function. The Option Memory capacity is 64 words. When a specific pattern of 55H and AAH is consecutively written into this register, the Option Memory Mapping function will be enabled and then the Option Memory code

can be read by using the table read instruction. The Option Memory addresses 00H~3FH will be mapped to Program Memory last page addresses C0H~FFH.

To successfully enable the Option Memory Mapping function, the specific pattern of 55H and AAH must be written into the ORMC register in two consecutive instruction cycles. It is therefore recommended that the global interrupt bit EMI should first be cleared before writing the specific pattern, and then set high again at a proper time according to users' requirements after the pattern is successfully written. An internal timer will be activated when the pattern is successfully written. The mapping operation will be automatically finished after a period of $4\times t_{LIRC}$. Therefore, users should read the data in time, otherwise the Option Memory Mapping function needs to be restarted. After the completion of each consecutive write operation to the ORMC register, the timer will recount.

When the table read instructions are used to read the Option Memory code, both "TABRD [m]" and "TABRDL [m]" instructions can be used. However, care must be taken if the "TABRD [m]" instruction is used, the table pointer defined by the TBHP register must be referenced to the last page. Refer to corresponding sections about the table read instruction for more details.

• **ORMC Register**

Bit	7	6	5	4	3	2	1	0
Name	ORMC7	ORMC6	ORMC5	ORMC4	ORMC3	ORMC2	ORMC1	ORMC0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **ORMC7~ORMC0**: Option Memory Mapping specific pattern
When a specific pattern of 55H and AAH is written into this register, the Option Memory Mapping function will be enabled. Note that the register content will be cleared after the MCU is woken up from the IDLE/SLEEP mode.

Status Register – STATUS

This 8-bit register contains the SC flag, CZ flag, zero flag (Z), carry flag (C), auxiliary carry flag (AC), overflow flag (OV), power down flag (PDF), and watchdog time-out flag (TO). These arithmetic/logical operation and system management flags are used to record the status and operation of the microcontroller.

With the exception of the TO and PDF flags, bits in the status register can be altered by instructions like most other registers. Any data written into the status register will not change the TO or PDF flag. In addition, operations related to the status register may give different results due to the different instruction operations. The TO flag can be affected only by a system power-up, a WDT time-out or by executing the "CLR WDT" or "HALT" instruction. The PDF flag is affected only by executing the "HALT" or "CLR WDT" instruction or during a system power-up.

The Z, OV, AC, C, SC and CZ flags generally reflect the status of the latest operations.

- C is set if an operation results in a carry during an addition operation or if a borrow does not take place during a subtraction operation; otherwise C is cleared. C is also affected by a rotate through carry instruction.
- AC is set if an operation results in a carry out of the low nibbles in addition, or no borrow from the high nibble into the low nibble in subtraction; otherwise AC is cleared.
- Z is set if the result of an arithmetic or logical operation is zero; otherwise Z is cleared.
- OV is set if an operation results in a carry into the highest-order bit but not a carry out of the highest-order bit, or vice versa; otherwise OV is cleared.
- PDF is cleared by a system power-up or executing the "CLR WDT" instruction. PDF is set by executing the "HALT" instruction.

- TO is cleared by a system power-up or executing the "CLR WDT" or "HALT" instruction. TO is set by a WDT time-out.
- CZ is the operational result of different flags for different instructions. Refer to register definitions for more details.
- SC is the result of the "XOR" operation which is performed by the OV flag and the MSB of the current instruction operation result.

In addition, on entering an interrupt sequence or executing a subroutine call, the status register will not be pushed onto the stack automatically. If the contents of the status registers are important and if the subroutine can corrupt the status register, precautions must be taken to correctly save it.

• STATUS Register

Bit	7	6	5	4	3	2	1	0
Name	SC	CZ	TO	PDF	OV	Z	AC	C
R/W	R/W	R/W	R	R	R/W	R/W	R/W	R/W
POR	x	x	0	0	x	x	x	x

"x": unknown

Bit 7 **SC**: The result of the "XOR" operation which is performed by the OV flag and the MSB of the instruction operation result.

Bit 6 **CZ**: The operational result of different flags for different instructions

For SUB/SUBM/LSUB/LSUBM instructions, the CZ flag is equal to the Z flag.

For SBC/SBCM/LSBC/LSBCM instructions, the CZ flag is the "AND" operation result which is performed by the previous operation CZ flag and current operation zero flag.

For other instructions, the CZ flag will not be affected.

Bit 5 **TO**: Watchdog Time-Out flag
0: After power up or executing the "CLR WDT" or "HALT" instruction
1: A watchdog time-out occurred

Bit 4 **PDF**: Power down flag
0: After power up or executing the "CLR WDT" instruction
1: By executing the "HALT" instruction

Bit 3 **OV**: Overflow flag
0: No overflow
1: An operation results in a carry into the highest-order bit but not a carry out of the highest-order bit or vice versa

Bit 2 **Z**: Zero flag
0: The result of an arithmetic or logical operation is not zero
1: The result of an arithmetic or logical operation is zero

Bit 1 **AC**: Auxiliary flag
0: No auxiliary carry
1: An operation results in a carry out of the low nibbles in addition, or no borrow from the high nibble into the low nibble in subtraction

Bit 0 **C**: Carry flag
0: No carry-out
1: An operation results in a carry during an addition operation or if a borrow does not take place during a subtraction operation

The C flag is also affected by a rotate through carry instruction.

EEPROM Data Memory

This device contains an area of internal EEPROM Data Memory. EEPROM is by its nature a non-volatile form of re-programmable memory, with data retention even when its power supply is removed. By incorporating this kind of data memory, a whole new host of application possibilities are made available to the designer. The availability of EEPROM storage allows information such as product identification numbers, calibration values, specific user data, system setup data or other product information to be stored directly within the product microcontroller. The process of reading and writing data to the EEPROM memory has been reduced to a very trivial affair.

EEPROM Data Memory Structure

The EEPROM Data Memory capacity is 1K×8 bits for the device. Unlike the Program Memory and RAM Data Memory, the EEPROM Data Memory is not directly mapped into memory space and is therefore not directly addressable in the same way as the other types of memory. Read and Write operations to the EEPROM are carried out in single byte operations using an address register pair and a data register in Sector 0 and a single control register in Sector 1.

EEPROM Registers

Four registers control the overall operation of the internal EEPROM Data Memory. These are the address registers, EEAL and EEAH, the data register, EED and a single control register, EEC. As both the EEAL, EEAH and EED registers are located in Sector 0, they can be directly accessed in the same way as any other Special Function Register. The EEC register however, being located in Sector 1, can only be read from or written to indirectly using the MP1L/MP1H or MP2L/MP2H Memory Pointer pairs and Indirect Addressing Register, IAR1/IAR2. Because the EEC control register is located at address 40H in Sector 1, the MP1L or MP2L Memory Pointer must first be set to the value 40H and the MP1H or MP2H Memory Pointer high byte set to the value, 01H, before any operations on the EEC register are executed.

Register Name	Bit							
	7	6	5	4	3	2	1	0
EEAL	EEAL7	EEAL6	EEAL5	EEAL4	EEAL3	EEAL2	EEAL1	EEAL0
EEAH	—	—	—	—	—	—	EEAH1	EEAH0
EED	EED7	EED6	EED5	EED4	EED3	EED2	EED1	EED0
EEC	EWERTS	EREN	ER	MODE	WREN	WR	RDEN	RD

EEPROM Register List

- **EEAL Register**

Bit	7	6	5	4	3	2	1	0
Name	EEAL7	EEAL6	EEAL5	EEAL4	EEAL3	EEAL2	EEAL1	EEAL0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **EEAL7~EEAL0**: Data EEPROM low byte address bit 7 ~ bit 0

• **EEAH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	EEAH1	EEAH0
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 **EEAH1~EEHA0**: Data EEPROM high byte address bit 1 ~ bit0

• **EED Register**

Bit	7	6	5	4	3	2	1	0
Name	EED7	EED6	EED5	EED4	EED3	EED2	EED1	EED0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **EED7~EED0**: Data EEPROM data bit 7 ~ bit 0

• **EEC Register**

Bit	7	6	5	4	3	2	1	0
Name	EWERTS	EREN	ER	MODE	WREN	WR	RDEN	RD
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **EWERTS**: EEPROM Erase time and Write time selection
0: Erase time is 3.2ms (t_{EEER}) / Write time is 2.2ms (t_{EEWR})
1: Erase time is 3.7ms (t_{EEER}) / Write time is 3.0ms (t_{EEWR})

Bit 6 **EREN**: Data EEPROM Erase Enable
0: Disable
1: Enable
This bit is used to enable data EEPROM erase function and must be set high before erase operations are carried out. This bit will be automatically reset to zero by the hardware after the erase cycle has finished. Clearing this bit to zero will inhibit data EEPROM erase operations.

Bit 5 **ER**: Data EEPROM Erase Control
0: Erase cycle has finished
1: Activate a erase cycle
When this bit is set high by the application program, an erase cycle will be activated. This bit will be automatically reset to zero by the hardware after the erase cycle has finished. Setting this bit high will have no effect if the EREN has not first been set high.

Bit 4 **MODE**: Data EEPROM Operation mode selection
0: Byte operation mode
1: Page operation mode
This is the EEPROM operation mode selection bit. When the bit is set high by the application program, the Page write, erase or read function will be selected. Otherwise, the byte write or read function will be selected. The EEPROM page buffer size is 16 bytes.

Bit 3 **WREN**: Data EEPROM Write Enable
0: Disable
1: Enable
This is the Data EEPROM Write Enable Bit, which must be set high before Data EEPROM write operations are carried out. Clearing this bit to zero will inhibit Data EEPROM write operations. Note that the WREN bit will automatically be cleared to zero after the write operation is finished.

Bit 2 **WR**: Data EEPROM Write Control
0: Write cycle has finished
1: Activate a write cycle
This is the Data EEPROM Write Control Bit and when set high by the application program will activate a write cycle. This bit will be automatically reset to zero by the hardware after the write cycle has finished. Setting this bit high will have no effect if the WREN has not first been set high.

Bit 1 **RDEN**: Data EEPROM Read Enable
0: Disable
1: Enable
This is the Data EEPROM Read Enable Bit which must be set high before Data EEPROM read operations are carried out. Clearing this bit to zero will inhibit Data EEPROM read operations.

Bit 0 **RD**: EEPROM Read Control
0: Read cycle has finished
1: Activate a read cycle
This is the Data EEPROM Read Control Bit and when set high by the application program will activate a read cycle. This bit will be automatically reset to zero by the hardware after the read cycle has finished. Setting this bit high will have no effect if the RDEN has not first been set high.

Note: 1. The EREN, ER, WREN, WR, RDEN and RD cannot be set high at the same time in one instruction.
2. Ensure that the f_{SUB} clock is stable before executing the erase or write operation.
3. Ensure that the erase or write operation is totally complete before changing the contents of the EEPROM related registers or activating the IAP function.

Read Operation from the EEPROM

Reading data from the EEPROM can be implemented by two modes for this device, byte read mode or page read mode, which is controlled by the EEPROM operation mode selection bit, MODE, in the EEC register.

Byte Read Mode

The EEPROM byte read operation can be executed when the mode selection bit, MODE, is cleared to zero. For a byte read operation the desired EEPROM address should first be placed in the EEAH and EEAL registers, as well as the read enable bit, RDEN, in the EEC register should be set high to enable the read function. Then setting the RD bit high will initiate the EEPROM byte read operation. Note that setting the RD bit high only will not initiate a read operation if the RDEN bit is not set high. When the read cycle terminates, the RD bit will automatically be cleared to zero and the EEPROM data can be read from the EED register. The data will remain in the EED register until another read or write operation is executed. The application program can poll the RD bit to determine when the data is valid for reading.

Page Read Mode

The EEPROM page read operation can be executed when the mode selection bit, MODE, is set high. The page size can be up to 16 bytes for the page read operation. For a page read operation the start address of the desired EEPROM page should first be placed in the EEAH and EEAL registers, as well as the read enable bit, RDEN, in the EEC register should be set high to enable the read function. Then setting the RD bit high will initiate the EEPROM page read operation. Note that setting the RD bit high only will not initiate a read operation if the RDEN bit is not set high. When the current byte read cycle terminates, the RD bit will automatically be cleared indicating that the EEPROM data can be read from the EED register, and the current address will be incremented by one by hardware. The data which is stored in the next EEPROM address can continuously be read from the EED register

when the RD bit is again set high without reconfiguring the EEPROM address and RDEN control bit. The application program can poll the RD bit to determine when the data is valid for reading.

The EEPROM address higher 6 bits are used to specify the desired page location while the lower 4 bits are used to point to the actual address. In the page operation mode the lower 4-bit address value will automatically be incremented by one. However, the higher 6-bit address value will not be incremented by hardware. When the EEPROM address lower 4-bit value which is internally incremented by one in the page mode reaches the page boundary, known as 0FH, the EEPROM address lower 4-bit value will stop at 0FH. The EEPROM address will not "roll over".

Page Erase Operation to the EEPROM

The EEPROM page erase operation can be executed when the mode selection bit, MODE, is set high. The EEPROM is capable of a 16-byte page erase. The internal page buffer will be cleared by hardware after power-on reset. When the EEPROM erase enable control bit, namely EREN, is changed from "1" to "0", the internal page buffer will also be cleared. Note that when the EREN bit is changed from "0" to "1", the internal page buffer will not be cleared. The EEPROM address higher 6 bits are used to specify the desired page location while the lower 4 bits are used to point to the actual address. In the page erase operation mode the lower 4-bit address value will automatically be incremented by one after each dummy data byte is written into the EED register. However, the higher 6-bit address value will not be incremented by hardware. When the EEPROM address lower 4-bit value which is internally incremented by one in the page mode reaches the page boundary, namely 0FH, the EEPROM address lower 4-bit value will stop at 0FH. The EEPROM address will not "roll over".

For page erase operations the start address of the desired EEPROM page should first be placed in the EEAH and EEAL registers and the dummy data to be written is placed in the EED register. The maximum data length for a page is 16 bytes. Note that the write operation to the EED register is used to tag address, it must be implemented to determine which addresses to be erased. When the page dummy data is completely written then the EREN bit in the EEC register should first be set high to enable erase operations and the ER bit must be immediately set high to initiate the EEPROM erase process. These two instructions must be executed in two consecutive instruction cycles to activate an erase operation successfully. The global interrupt enable bit EMI should also first be cleared before implementing an erase operation and then set again after a valid erase activation procedure has completed.

As the EEPROM erase cycle is controlled using an internal timer whose operation is asynchronous to microcontroller system clock, a certain time will elapse before the data will have been erased from the EEPROM. Detecting when the erase cycle has finished can be implemented either by polling the ER bit in the EEC register or by using the EEPROM interrupt. When the erase cycle terminates, the ER bit will be automatically cleared to zero by the microcontroller, informing the user that the page data has been erased. The application program can therefore poll the ER bit to determine when the erase cycle has ended. After the erase operation is finished, the EREN bit will be set low by hardware. The Data EEPROM erased page content will all be zero after a page erase operation.

Write Operation to the EEPROM

Writing data to the EEPROM can be implemented by two modes for this device, byte write mode or page write mode, which is controlled by the EEPROM operation mode selection bit, MODE, in the EEC register.

Byte Write Mode

The EEPROM byte write operation can be executed when the mode selection bit, MODE, is cleared to zero. For byte write operations the desired EEPROM address should first be placed in the EEAL and EEAH registers and the data to be written should be placed in the EED register. To write data to the EEPROM, the write enable bit, WREN, in the EEC register must first be set high to enable the write function. After this, the WR bit in the EEC register must be immediately set high to initiate a write cycle. These two instructions must be executed in two consecutive instruction cycles to activate a write operation successfully. The global interrupt bit EMI should also first be cleared before implementing any write operations, and then set high again after a valid write activation procedure has completed. Note that setting the WR bit high will not initiate a write cycle if the WREN bit has not been set.

As the EEPROM write cycle is controlled using an internal timer whose operation is asynchronous to microcontroller system clock, a certain time will elapse before the data will have been written into the EEPROM. Detecting when the write cycle has finished can be implemented either by polling the WR bit in the EEC register or by using the EEPROM interrupt. When the write cycle terminates, the WR bit will be automatically cleared to zero by the microcontroller, informing the user that the data has been written to the EEPROM. The application program can therefore poll the WR bit to determine when the write cycle has ended. Note that a byte erase operation will automatically be executed before a byte write operation is successfully activated.

Page Write Mode

Before a page write operation is executed, it is important to ensure that a relevant page erase operation has been successfully executed. The EEPROM page write operation can be executed when the mode selection bit, MODE, is set high. The EEPROM is capable of a 16-byte page write. The internal page buffer will be cleared by hardware after power on reset. When the EEPROM write enable control bit, namely WREN, is changed from “1” to “0”, the internal page buffer will also be cleared. Note that when the WREN bit is changed from “0” to “1”, the internal page buffer will not be cleared. A page write is initiated in the same way as a byte write initiation except that the EEPROM data can be written up to 16 bytes. The EEPROM address higher 6 bits are used to specify the desired page location while the lower 4 bits are used to point to the actual address. In the page write operation mode the lower 4-bit address value will automatically be incremented by one after each data byte is written into the EED register. However, the higher 6-bit address value will not be incremented by hardware. When the EEPROM address lower 4-bit value which is internally incremented by one in the page mode reaches the page boundary, namely 0FH, the EEPROM address lower 4-bit value will stop at 0FH. The EEPROM address will not “roll over”. At this point any data write operations to the EED register will be invalid.

For page write operations the the start address of the desired EEPROM page must first be placed in the EEAH and EEAL registers and the data placed in the EED register. The maximum data length for a page is 16 bytes. Note that when a data byte is written into the EED register, then the data in the EED register will be loaded into the internal page buffer and the current address value will automatically be incremented by one. When the page data is completely written into the page buffer, then the write enable bit, WREN, in the EEC register should first be set high to enable the write function. After this, the WR bit in the EEC register must be immediately set high to initiate a write cycle. These two instructions must be executed in two consecutive instruction cycles. The global interrupt bit EMI should also first be cleared before implementing any write operations, and then set again after a valid write activation procedure has completed. Note that setting the WR bit high will not initiate a write cycle if the WREN bit has not been set.

As the EEPROM write cycle is controlled using an internal timer whose operation is asynchronous to microcontroller system clock, a certain time will elapse before the data will have been written into the EEPROM. Detecting when the write cycle has finished can be implemented either by polling the WR bit in the EEC register or by using the EEPROM interrupt. When the write cycle terminates, the WR bit will be automatically cleared to zero by the microcontroller, informing the user that the data has been written to the EEPROM. The application program can therefore poll the WR bit to determine when the write cycle has ended. After the write operation is finished, the WREN bit will be set low by hardware.

Write Protection

Protection against inadvertent write operation is provided in several ways. After the device is powered-on the Write Enable bit in the control register will be cleared preventing any write operations. Also at power-on the Memory Pointer high byte register, MP1H or MP2H, will be reset to zero, which means that Data Memory Sector 0 will be selected. As the EEPROM control register is located in Sector 1, this adds a further measure of protection against spurious write operations. During normal program operation, ensuring that the Write Enable bit in the control register is cleared will safeguard against incorrect write operations.

EEPROM Interrupt

The EEPROM interrupt is generated when an EEPROM erase or write cycle has ended. The EEPROM interrupt must first be enabled by setting the DEE bit in the relevant interrupt register. However, as the EEPROM interrupt is contained within a Multi-function Interrupt, the associated multi-function interrupt enable bit must also be set. When an EEPROM erase or write cycle ends, the DEF request flag and its associated multi-function interrupt request flag will both be set high. If the global, EEPROM and multi-function interrupts are enabled and the stack is not full, a jump to the associated Multi-function Interrupt vector will take place. When the interrupt is serviced only the Multi-function interrupt flag will be automatically reset, the EEPROM interrupt flag must be manually reset by the application program. The EMI bit will also be automatically cleared to disable other interrupts. More details can be obtained in the Interrupts section.

Programming Considerations

Care must be taken that data is not inadvertently written to the EEPROM. Protection can be enhanced by ensuring that the Write Enable bit is normally cleared to zero when not writing. Also the Memory Pointer high byte register, MP1H or MP2H, could be normally cleared to zero as this would inhibit access to Sector 1 where the EEPROM control register exists. Although certainly not necessary, consideration might be given in the application program to the checking of the validity of new write data by a simple read back process.

When writing or erasing data the WR/ER bit must be set high immediately after the WREN/EREN bit has been set high, to ensure the write cycle or erase executes correctly. The global interrupt bit EMI should also be cleared before a write or erase cycle is executed and then set again after a valid write or erase activation procedure has completed. Note that the device should not enter the IDLE or SLEEP mode until the EEPROM read or write operation is totally complete. Otherwise, the EEPROM read, erase or write operation will fail.

Programming Examples

Reading a Data Byte from the EEPROM – polling method

```
MOV  A, 40H                ; setup memory mointer lower byte MP1L
MOV  MP1L, A               ; MP1L points to EEC register
MOV  A, 01H                ; setup memory pointer high byte MP1H
MOV  MP1H, A
CLR  IAR1.4                ; clear MODE bit, select byte operation mode
MOV  A, EEPROM_ADRES_H     ; user defined high byte address
MOV  EEAH, A
MOV  A, EEPROM_ADRES_L     ; user defined low byte address
MOV  EEAL, A
SET  IAR1.1                ; set RDEN bit, enable read operations
SET  IAR1.0                ; start Read Cycle - set RD bit
BACK:
SZ   IAR1.0                ; check for read cycle end
JMP  BACK
CLR  IAR1                  ; disable EEPROM read function
CLR  MP1H
MOV  A, EED                ; move read data to register
MOV  READ_DATA, A
```

Reading a Data Page from the EEPROM – polling method

```
MOV  A, 40H                ; set memory pointer low byte MP1L
MOV  MP1L, A               ; MP1L points to EEC register
MOV  A, 01H                ; set memory pointer high byte MP1H
MOV  MP1H, A
SET  IAR1.4                ; set MODE bit, select page operation mode
MOV  A, EEPROM_ADRES_H     ; user defined high byte address
MOV  EEAH, A
MOV  A, EEPROM_ADRES_L     ; user defined low byte address
MOV  EEAL, A
SET  IAR1.1                ; set RDEN bit, enable read operations
; ~~~~ The data length can be up to 16 bytes (Start) ~~~~
CALL READ
CALL READ
:
:
JMP PAGE_READ_FINISH
; ~~~~ The data length can be up to 16 bytes (End) ~~~~
READ:
SET  IAR1.0                ; start Read Cycle - set RD bit
BACK:
SZ   IAR1.0                ; check for read cycle end
JMP  BACK
MOV  A, EED                ; move read data to register
MOV  READ_DATA, A
RET
:
PAGE_READ_FINISH:
CLR  IAR1                  ; disable EEPROM read function
CLR  MP1H
```

Erasing a Data Page to the EEPROM − polling method

```
MOV  A, 40H                ; set memory pointer low byte MP1L
MOV  MP1L, A               ; MP1 points to EEC register
MOV  A, 01H                ; set memory pointer high byte MP1H
MOV  MP1H, A
SET  IAR1.4                ; set MODE bit, select page operation mode
MOV  A, EEPROM_ADRES_H     ; user defined high byte address
MOV  EEAH, A
MOV  A, EEPROM_ADRES_L     ; user defined low byte address
MOV  EEAL, A
; ~~~~ The data length can be up to 16 bytes (Start) ~~~~
CALL WRITE_BUF
CALL WRITE_BUF
:
:
JMP  Erase_START
; ~~~~ The data length can be up to 16 bytes (End) ~~~~
WRITE_BUF:
MOV  A, EEPROM_DATA        ; user defined data, erase mode don't care data value
MOV  EED, A
RET
:
Erase_START:
CLR  EMI
SET  IAR1.6                ; set EREN bit, enable write operations
SET  IAR1.5                ; start Write Cycle - set ER bit - executed immediately
                           ; after setting EREN bit
SET  EMI
BACK:
SZ   IAR1.5                ; check for write cycle end
JMP  BACK
CLR  MP1H
```

Writing a Data Byte to the EEPROM – polling method

```
MOV  A, 40H                ; set memory pointer low byte MP1L
MOV  MP1L, A               ; MP1L points to EEC register
MOV  A, 01H                ; set memory pointer high byte MP1H
MOV  MP1H, A
CLR  IAR1.4                ; clear MODE bit, select byte write mode
MOV  A, EEPROM_ADRES       ; user defined high byte address
MOV  EEAH, A
MOV  A, EEPROM_ADRES       ; user defined low byte address
MOV  EEAL, A
MOV  A, EEPROM_DATA        ; user defined data
MOV  EED, A
CLR  EMI
SET  IAR1.3                ; set WREN bit, enable write operations
SET  IAR1.2                ; start Write Cycle – set WR bit – executed immediately
                           ; after setting WREN bit
SET  EMI
BACK:
SZ   IAR1.2                ; check for write cycle end
JMP  BACK
CLR  MP1H
```

Writing a Data Page to the EEPROM – polling method

```
MOV  A, 40H                ; set memory pointer low byte MP1L
MOV  MP1L, A               ; MP1L points to EEC register
MOV  A, 01H                ; set memory pointer high byte MP1H
MOV  MP1H, A
SET  IAR1.4                ; set MODE bit, select page operation mode
MOV  A, EEPROM_ADRES_H     ; user defined high byte address
MOV  EEAH, A
MOV  A, EEPROM_ADRES_L     ; user defined low byte address
MOV  EEAL, A
; ~~~~ The data length can be up to 16 bytes (Start) ~~~~
CALL WRITE_BUF
CALL WRITE_BUF
:
:
JMP  WRITE_START
; ~~~~ The data length can be up to 16 bytes (End) ~~~~
WRITE_BUF:
MOV  A, EEPROM_DATA        ; user define data
MOV  EED, A
RET
:
WRITE_START:
CLR  EMI
SET  IAR1.3                ; set WREN bit, enable write operations
SET  IAR1.2                ; start Write Cycle - set WR bit - executed immediately
                           ; after setting WREN bit
SET  EMI
BACK:
SZ   IAR1.2                ; check for write cycle end
JMP  BACK
CLR  MP1H
```

Oscillators

Various oscillator options offer the user a wide range of functions according to their various application requirements. The flexible features of the oscillator functions ensure that the best optimisation can be achieved in terms of speed and power saving. Oscillator selections and operation are selected through a combination of configuration option and relevant control registers.

Oscillator Overview

In addition to being the source of the main system clock the oscillators also provide clock sources for the Watchdog Timer and Time Base Interrupts. External oscillators requiring some external components as well as fully integrated internal oscillators, requiring no external components, are provided to form a range of both fast and slow system oscillators. The higher frequency oscillators provide higher performance but carry with it the disadvantage of higher power requirements, while the opposite is of course true for the lower frequency oscillators. With the capability of dynamically switching between fast and slow system clock, the device has the flexibility to optimize the performance/power ratio, a feature especially important in power sensitive portable applications.

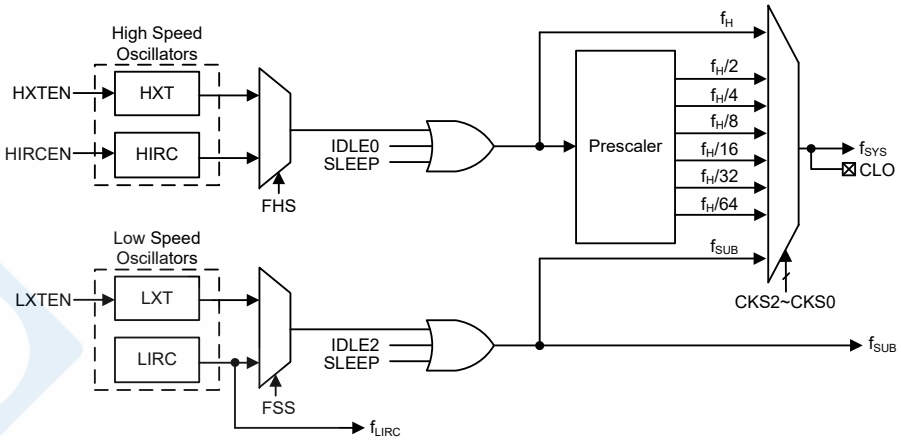
Type	Name	Frequency	Pins
External High Speed Crystal Oscillator	HXT	400kHz~16MHz	OSC1/OSC2
Internal High Speed RC Oscillator	HIRC	8/12/16MHz	—
External Low Speed Crystal Oscillator	LXT	32.768kHz	XT1/XT2
Internal Low Speed RC Oscillator	LIRC	32kHz	—

Oscillator Types

System Clock Configurations

There are several oscillator sources, two high speed oscillators and two low speed oscillators for the device. The high speed oscillators are the external crystal/ceramic oscillator, HXT, and the internal 8/12/16MHz RC oscillator, HIRC. The low speed oscillators are the internal 32kHz RC oscillator, LIRC, and the external 32.768kHz crystal oscillator, LXT. Selecting whether the low or high speed oscillator is used as the system oscillator is implemented using the CKS2~CKS0 bits in the SCC register and the system clock can be dynamically selected.

The actual source clock used for the low speed oscillator is chosen via the FSS bit in the SCC register while for the high speed oscillator the source clock is selected by the FHS bit in the SCC register. The frequency of the slow speed or high speed system clock is also determined using the CKS2~CKS0 bits in the SCC register. Note that two oscillator selections must be made namely one high speed and one low speed system oscillators. It is not possible to choose a no-oscillator selection for either the high or low speed oscillator.

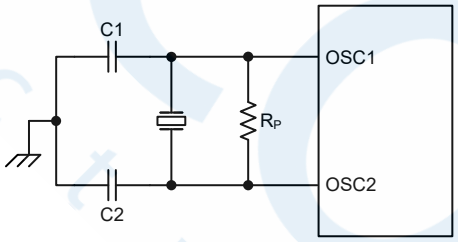


System Clock Configurations

External Crystal/Ceramic Oscillator – HXT

The External Crystal/Ceramic System Oscillator is one of the high frequency oscillator choices, which is selected via a software control bit, FHS. For most crystal oscillator configurations, the simple connection of a crystal across OSC1 and OSC2 will create the necessary phase shift and feedback for oscillation, without requiring external capacitors. However, for some crystal types and frequencies, to ensure oscillation, it may be necessary to add two small value capacitors, C1 and C2. Using a ceramic resonator will usually require two small value capacitors, C1 and C2, to be connected as shown for oscillation to occur. The values of C1 and C2 should be selected in consultation with the crystal or resonator manufacturer's specification.

For oscillator stability and to minimise the effects of noise and crosstalk, it is important to ensure that the crystal and any associated resistors and capacitors along with interconnecting lines are all located as close to the MCU as possible.



Note: 1. R_P is normally not required. C1 and C2 are required.
2. Although not shown OSC1/OSC2 pins have a parasitic capacitance of around 7pF.

Crystal/Ceramic Oscillator – HXT

HXT Oscillator C1 and C2 Values		
Crystal Frequency	**C1**	**C2**
16MHz	0pF	0pF
12MHz	0pF	0pF
8MHz	0pF	0pF
6MHz	0pF	0pF
4MHz	0pF	0pF
1MHz	100pF	100pF
Note: C1 and C2 values are for guidance only.		

Crystal Recommended Capacitor Values

Internal High Speed RC Oscillator – HIRC

The internal RC oscillator is one of the high frequency oscillator choices, which is selected via a software control bit, FHS. It is a fully integrated system oscillator requiring no external components. The internal RC oscillator has three fixed frequencies of 8MHz, 12MHz and 16MHz, which are selected by HIRC1~HIRC0 bits in the HIRCC register. These bits must be setup to match the selected configuration option frequency to ensure that the HIRC frequency accuracy specified in the A.C. Characterisitcs is achieved. Device trimming during the manufacturing process and the inclusion of internal frequency compensation circuits are used to ensure that the influence of the power supply voltage, temperature and process variations on the oscillation frequency are minimised. Note that if this internal system clock option requires no external pins for its operation.

External 32.768kHz Crystal Oscillator – LXT

The External 32.768kHz Crystal System Oscillator is one of the low frequency oscillator choices, which is selected via a software control bit, FSS. This clock source has a fixed frequency of 32.768kHz and requires a 32.768kHz crystal to be connected between pins XT1 and XT2. The external resistor and capacitor components connected to the 32.768kHz crystal are necessary to provide oscillation. For applications where precise frequencies are essential, these components may be required to provide frequency compensation due to different crystal manufacturing tolerances. After the LXT oscillator is enabled by setting the LXTEN bit to 1, there is a time delay associated with the LXT oscillator waiting for it to start-up.

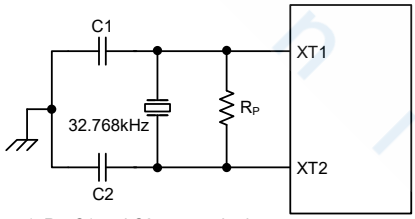
When the microcontroller enters the SLEEP or IDLE Mode, the system clock is switched off to stop microcontroller activity and to conserve power. However, in many microcontroller applications it may be necessary to keep the internal timers operational even when the microcontroller is in the SLEEP or IDLE Mode. To do this, another clock, independent of the system clock, must be provided.

However, for some crystals, to ensure oscillation and accurate frequency generation, it is necessary to add two small value external capacitors, C1 and C2. The exact values of C1 and C2 should be selected in consultation with the crystal or resonator manufacturer's specification. The external parallel feedback resistor, R_P, is required.

The pin-shared software control bits determine if the XT1/XT2 pins are used for the LXT oscillator or as I/O or other pin-shared functional pins.

- If the LXT oscillator is not used for any clock source, the XT1/XT2 pins can be used as normal I/O or other pin-shared functional pins.
- If the LXT oscillator is used for any clock source, the 32.768kHz crystal should be connected to the XT1/XT2 pins.

For oscillator stability and to minimise the effects of noise and crosstalk, it is important to ensure that the crystal and any associated resistors and capacitors along with interconnecting lines are all located as close to the MCU as possible.



Note: 1. R_P, C1 and C2 are required.
2. Although not shown XT1/XT2 pins have a parasitic capacitance of around 7pF.

External LXT Oscillator

LXT Oscillator C1 and C2 Values		
Crystal Frequency	C1	C2
32.768kHz	10pF	10pF
Note: 1. C1 and C2 values are for guidance only. 2. R_P=5MΩ~10MΩ is recommended.		

32.768kHz Crystal Recommended Capacitor Values

LXT Oscillator Low Power Function

The LXT oscillator can function in one of two modes, the Speed Up Mode and the Low Power Mode. The mode selection is executed using the LXTSP bit in the register.

LXTSP	LXT Mode
0	Low Power
1	Speed Up

When the LXTSP bit is set to high, the LXT Speed Up Mode will be enabled. In the Speed Up Mode the LXT oscillator will power up and stabilise quickly. However, after the LXT oscillator has fully powered up, it can be placed into the Low Power Mode by clearing the LXTSP bit to zero and the oscillator will continue to run but with reduced current consumption. It is important to note that the LXT operating mode switching must be properly controlled before the LXT oscillator clock is selected as the system clock source. Once the LXT oscillator clock is selected as the system clock source using the CKS2~CKS0 bits and FSS bit in the SCC register, the LXT oscillator operating mode cannot be changed.

It should be noted that, no matter what condition the LXTSP bit is set to, the LXT oscillator will be always function normally, the only difference is that it will take more time to start up if it is in the low power mode.

Internal 32kHz Oscillator – LIRC

The Internal 32kHz Oscillator is one of the low frequency oscillator choices, which is selected by the FSS bit in the SCC register. It is a fully integrated RC oscillator with a typical frequency of 32kHz, requiring no external components for its implementation. Device trimming during the manufacturing process and the inclusion of internal frequency compensation circuits are used to ensure that the influence of the power supply voltage, temperature and process variations on the oscillation frequency are minimised.

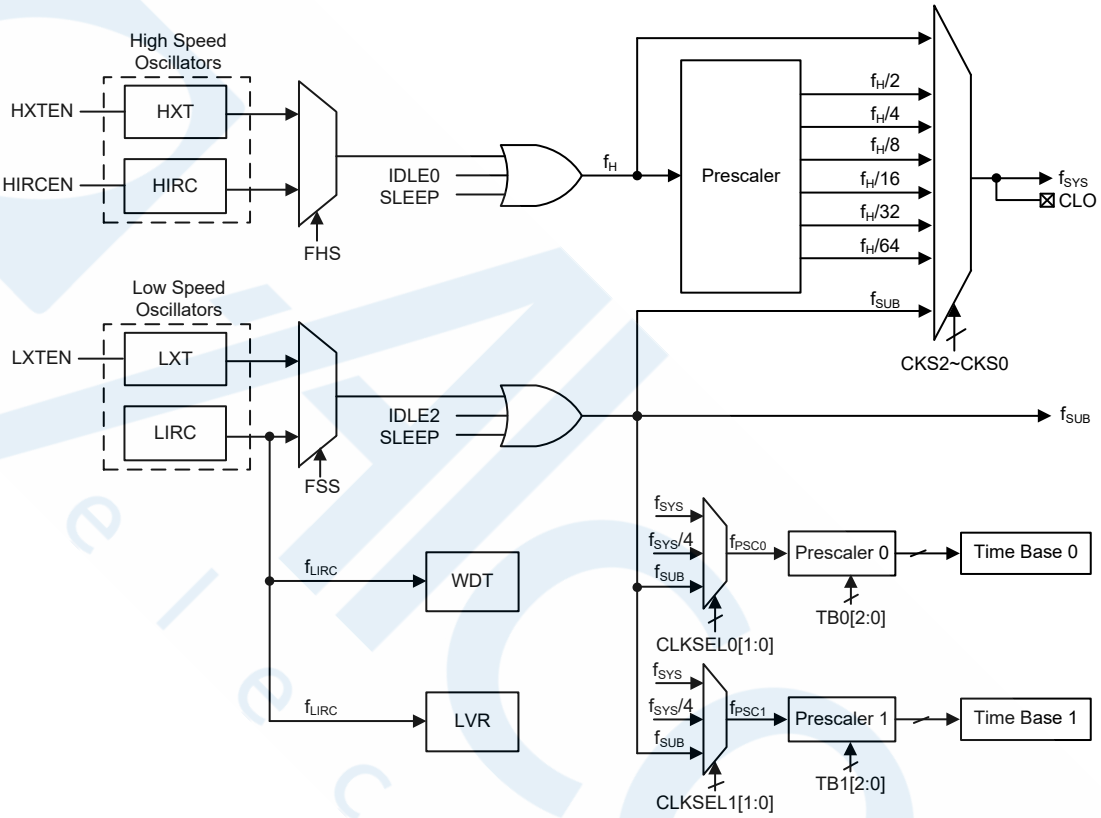
Operating Modes and System Clocks

Present day applications require that their microcontrollers have high performance but often still demand that they consume as little power as possible, conflicting requirements that are especially true in battery powered portable applications. The fast clocks required for high performance will by their nature increase current consumption and of course vice-versa, lower speed clocks reduce current consumption. As Holtek has provided the device with both high and low speed clock sources and the means to switch between them dynamically, the user can optimise the operation of their microcontroller to achieve the best performance/power ratio.

System Clocks

The device has many different clock sources for both the CPU and peripheral function operation. By providing the user with a wide range of clock selections using register programming, a clock system can be configured to obtain maximum application performance.

The main system clock, can come from either a high frequency, f_H, or low frequency, f_{SUB}, source, and is selected using the CKS2~CKS0 bits in the SCC register. The high speed system clock can be sourced from either an HXT or HIRC oscillator, selected via configuring the FHS bit in the SCC register. The low speed system clock source can be sourced from internal clock f_{SUB}. If f_{SUB} is selected then it can be sourced by either the LXT or LIRC oscillator, selected via configuring the FSS bit in the SCC register. The other choice, which is a divided version of the high speed system oscillator has a range of $f_H/2$~$f_H/64$.



Device Clock Configurations

Note: When the system clock source f_{SYS} is switched to f_{SUB} from f_H, the high speed oscillator will stop to conserve the power or continue to oscillate to provide the clock source, f_H~$f_H/64$, for peripheral circuit to use, which is determined by configuring the corresponding high speed oscillator enable control bit.

System Operation Modes

There are six different modes of operation for the microcontroller, each one with its own special characteristics and which can be chosen according to the specific performance and power requirements of the application. There are two modes allowing normal operation of the microcontroller, the FAST Mode and SLOW Mode. The remaining four modes, the SLEEP, IDLE0, IDLE1 and IDLE2 Mode, are used when the microcontroller CPU is switched off to conserve power.

Operation Mode	CPU	Register Setting			f_{SYS}	f_H	f_{SUB}	f_{LIRC}
		FHIDEN	FSIDEN	CKS2~CKS0				
FAST	On	x	x	000~110	f_H~f_H/64	On	On	On
SLOW	On	x	x	111	f_{SUB}	On/Off (1)	On	On
IDLE0	Off	0	1	000~110	Off	Off	On	On
				111	On			
IDLE1	Off	1	1	xxx	On	On	On	On
IDLE2	Off	1	0	000~110	On	On	Off	On
				111	Off			
SLEEP	Off	0	0	xxx	Off	Off	Off	On/Off (2)

"x": Don't care

Note: 1. The f_H clock will be switched on or off by configuring the corresponding oscillator enable bit in the SLOW mode.

2. The f_{LIRC} clock can be switched on or off which is controlled by the WDT function being enabled or disabled in the SLEEP mode.

FAST Mode

This is one of the main operating modes where the microcontroller has all of its functions operational and where the system clock is provided by one of the high speed oscillators. This mode operates allowing the microcontroller to operate normally with a clock source will come from one of the high speed oscillators, either the HXT or HIRC oscillator, selected by the FHS bit in the SCC register. The high speed oscillator will however first be divided by a ratio ranging from 1 to 64, the actual ratio being selected by the CKS2~CKS0 bit in the SCC register. Although a high speed oscillator is used, running the microcontroller at a divided clock ratio reduces the operating current.

SLOW Mode

This is also a mode where the microcontroller operates normally although now with a slower speed clock source. The clock source used will be from f_{SUB}. The f_{SUB} clock is derived from either the LIRC or LXT oscillator determined by the FSS bit in the SCC register.

SLEEP Mode

The SLEEP Mode is entered when a HALT instruction is executed and when the FHIDEN and FSIDEN bit are low. In the SLEEP mode the CPU will be stopped. The f_{SUB} clock provided to the peripheral function will also be stopped, too. However the f_{LIRC} clock can continue to operate if the WDT function is enabled by the WDTC register.

IDLE0 Mode

The IDLE0 Mode is entered when a HALT instruction is executed and when the FHIDEN bit in the SCC register is low and the FSIDEN bit in the SCC register is high. In the IDLE0 Mode the CPU will be switched off but the low speed oscillator will be turned on to drive some peripheral functions.

IDLE1 Mode

The IDLE1 Mode is entered when a HALT instruction is executed and when the FHIDEN bit in the SCC register is high and the FSIDEN bit in the SCC register is high. In the IDLE1 Mode the CPU will be switched off but both the high and low speed oscillators will be turned on to provide a clock source to keep some peripheral functions operational.

IDLE2 Mode

The IDLE2 Mode is entered when a HALT instruction is executed and when the FHIDEN bit in the SCC register is high and the FSIDEN bit in the SCC register is low. In the IDLE2 Mode the CPU and low speed oscillator will be switched off but the high speed oscillator will be turned on to provide a clock source to keep some peripheral functions operational.

Control Registers

The registers, SCC, HIRCC, HXTC and LXTC, are used to control the system clock and the corresponding oscillator configurations.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SCC	CKS2	CKS1	CKS0	—	FHS	FSS	FHIDEN	FSIDEN
HIRCC	—	—	—	—	HIRC1	HIRC0	HIRCF	HIRCEN
HXTC	—	—	—	—	—	HXTM	HXTF	HXTEN
LXTC	—	—	—	—	—	LXTSP	LXTF	LXTEN

System Operating Mode Control Register List

- **SCC Register**

Bit	7	6	5	4	3	2	1	0
Name	CKS2	CKS1	CKS0	—	FHS	FSS	FHIDEN	FSIDEN
R/W	R/W	R/W	R/W	—	R/W	R/W	R/W	R/W
POR	0	0	0	—	0	0	0	0

Bit 7~5 **CKS2~CKS0**: System clock selection
000: f_H
001: $f_H/2$
010: $f_H/4$
011: $f_H/8$
100: $f_H/16$
101: $f_H/32$
110: $f_H/64$
111: f_{SUB}
These three bits are used to select which clock is used as the system clock source. In addition to the system clock source directly derived from f_H or f_{SUB}, a divided version of the high speed system oscillator can also be chosen as the system clock source.

Bit 4 Unimplemented, read as "0"

Bit 3 **FHS**: High Frequency oscillator selection
0: HIRC
1: HXT

Bit 2 **FSS**: Low Frequency oscillator selection
0: LIRC
1: LXT

Bit 1 **FHIDEN**: High Frequency oscillator control when CPU is switched off
0: Disable
1: Enable
This bit is used to control whether the high speed oscillator is activated or stopped when the CPU is switched off by executing a "HALT" instruction.

Bit 0 **FSIDEN**: Low Frequency oscillator control when CPU is switched off
0: Disable
1: Enable
This bit is used to control whether the low speed oscillator is activated or stopped when the CPU is switched off by executing a "HALT" instruction.

Note: A certain delay is required before the relevant clock is successfully switched to the target clock source after any clock switching setup using the CKS2~CKS0 bits, FHS bit or FSS bit. A proper delay time must be arranged before executing the following operations which require immediate reaction with the target clock source.

Clock switching delay time=$4\times t_{SYS}+[0\sim(1.5\times t_{Curr.}+0.5\times t_{Tar.})]$, where $t_{Curr.}$ indicates the current clock period, $t_{Tar.}$ indicates the target clock period and t_{SYS} indicates the current system clock period.

• **HIRCC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	HIRC1	HIRC0	HIRCF	HIRCEN
R/W	—	—	—	—	R/W	R/W	R	R/W
POR	—	—	—	—	0	0	0	1

Bit 7~4 Unimplemented, read as "0"

Bit 3~2 **HIRC1~HIRC0**: HIRC frequency selection
00: 8MHz
01: 12MHz
10: 16MHz
11: 8MHz

When the HIRC oscillator is enabled or the HIRC frequency selection is changed by application program, the clock frequency will automatically be changed after the HIRCF flag is set high.

It is recommended that the HIRC frequency selected by these two bits should be same with the frequency determined by the configuration option to achieve the HIRC frequency accuracy specified in the A.C. Characteristics.

Bit 1 **HIRCF**: HIRC oscillator stable flag
0: HIRC unstable
1: HIRC stable

This bit is used to indicate whether the HIRC oscillator is stable or not. When the HIRCEN bit is set high to enable the HIRC oscillator or the HIRC frequency selection is changed by application program, the HIRCF bit will first be cleared to zero and then set high after the HIRC oscillator is stable.

Bit 0 **HIRCEN**: HIRC oscillator enable control
0: Disable
1: Enable

• **HXTC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	HXTM	HXTF	HXTEN
R/W	—	—	—	—	—	R/W	R	R/W
POR	—	—	—	—	—	0	0	0

Bit 7~3 Unimplemented, read as "0"

Bit 2 **HXTM**: HXT mode selection
0: HXT frequency ≤ 10MHz (sink/source current is smaller)
1: HXT frequency > 10MHz (sink/source current is largler)

Note that this bit should be configured correctly according to the used HXT frequency. If HXTM=0 while the HXT frequency is larger than 10MHz, the oscillation performance at a low voltage condition may be not well. If HXTM=1 while the HXT frequency is less than 10MHz, the oscillator frequency and current may be obnormal.

This bit must be properly configured before the HXT is enabled. When the OSC1 and OSC2 pin functions have been enabled using relevant pin-shared control bits and the HXTEN bit is set to 1 to enable the HXT oscillator, it is invalid to change the value of this bit. When the OSC1 and OSC2 pin functions are disabled, then the HXTM bit can be changed by software, regardless of the HXTEN bit value.

Bit 1 **HXTF**: HXT oscillator stable flag
0: HXT unstable
1: HXT stable
This bit is used to indicate whether the HXT oscillator is stable or not. When the HXTEN bit is set high to enable the HXT oscillator, the HXTF bit will first be cleared to zero and then set high after the HXT oscillator is stable.

Bit 0 **HXTEN**: HXT oscillator enable control
0: Disable
1: Enable

• **LXTC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	LXTSP	LXTF	LXTEN
R/W	—	—	—	—	—	R/W	R	R/W
POR	—	—	—	—	—	0	0	0

Bit 7~3 Unimplemented, read as "0"

Bit 2 **LXTSP**: LXT Speed up control
0: Disable – Low power
1: Enable – Speed up
This bit is used to control whether the LXT oscillator is operating in the low power or speed up mode. When the LXTSP bit is set high, the LXT oscillator will oscillate quickly but consume more power. If the LXTSP bit is cleared to zero, the LXT oscillator will consume less power but take longer time to stablise. It is important to note that this bit cannot be changed after the LXT oscillator is selected as the system clock source using the CKS2~CKS0 and FSS bits in the SCC register.

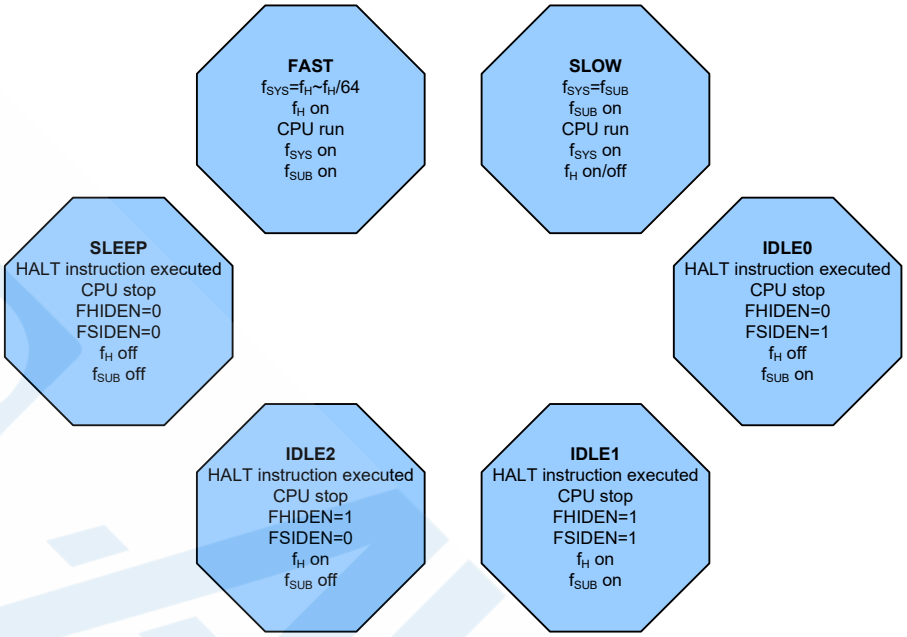
Bit 1 **LXTF**: LXT oscillator stable flag
0: LXT unstable
1: LXT stable
This bit is used to indicate whether the LXT oscillator is stable or not. When the LXTEN bit is set high to enable the LXT oscillator, the LXTF bit will first be cleared to zero and then set high after the LXT oscillator is stable.

Bit 0 **LXTEN**: LXT oscillator enable control
0: Disable
1: Enable

Operating Mode Switching

The device can switch between operating modes dynamically allowing the user to select the best performance/power ratio for the present task in hand. In this way microcontroller operations that do not require high performance can be executed using slower clocks thus requiring less operating current and prolonging battery life in portable applications.

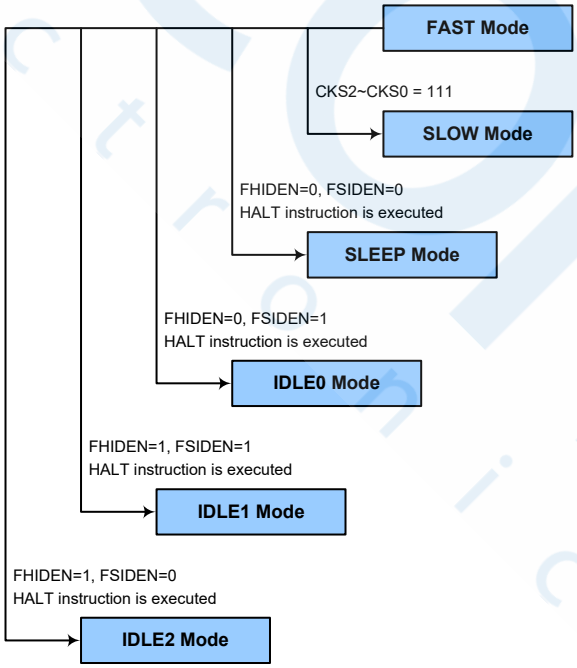
In simple terms, mode switching between the FAST Mode and SLOW Mode is executed using the CKS2~CKS0 bits in the SCC register while mode switching from the FAST/SLOW Modes to the SLEEP/IDLE Modes is executed via the HALT instruction. When a HALT instruction is executed, whether the device enters the IDLE Mode or the SLEEP Mode is determined by the condition of the FHIDEN and FSIDEN bits in the SCC register.



FAST Mode to SLOW Mode Switching

When running in the FAST Mode, which uses the high speed system oscillator, and therefore consumes more power, the system clock can switch to run in the SLOW Mode by setting the CKS2~CKS0 bits to "111" in the SCC register. This will then use the low speed system oscillator which will consume less power. Users may decide to do this for certain operations which do not require high performance and can subsequently reduce power consumption.

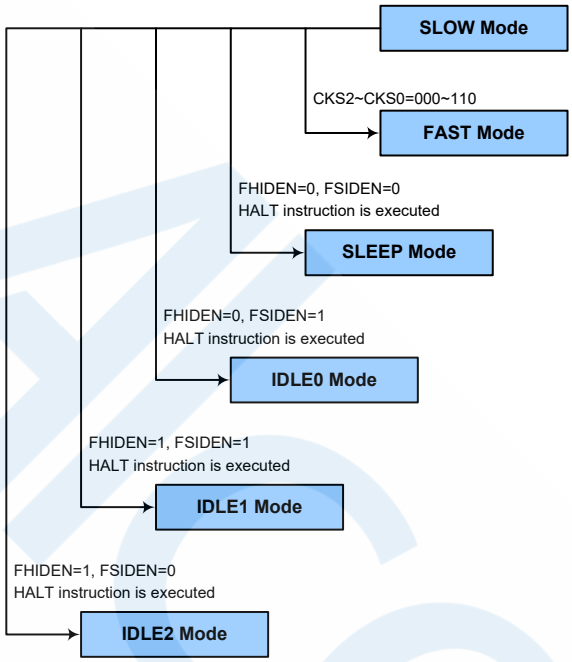
The SLOW Mode system clock is sourced from the LXT or LIRC oscillator determined by the FSS bit in the SCC register and therefore requires the selected oscillator to be stable before full mode switching occurs.



SLOW Mode to FAST Mode Switching

In SLOW mode the system clock is derived from f_{SUB}. When system clock is switched back to the FAST mode from f_{SUB}, the CKS2~CKS0 bits should be set to "000"~"110" and then the system clock will respectively be switched to f_H~f_H/64.

However, if f_H is not used in SLOW mode and thus switched off, it will take some time to re-oscillate and stabilise when switching to the FAST mode from the SLOW Mode. This is monitored using the HXTF bit in the HXTC register or the HIRCF bit in the HIRCC register. The time duration required for the high speed system oscillator stabilization is specified in the System Start Up Time Characteristics.



Entering the SLEEP Mode

There is only one way for the device to enter the SLEEP Mode and that is to execute the "HALT" instruction in the application program with both the FHIDEN and FSIDEN bits in the SCC register equal to "0". In this mode all the clocks and functions will be switched off except the WDT function. When this instruction is executed under the conditions described above, the following will occur:

- The system clock will be stopped and the application program will stop at the "HALT" instruction.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag PDF will be set, and WDT timeout flag TO will be cleared.
- The WDT will be cleared and resume counting if the WDT function is enabled. If the WDT function is disabled, the WDT will be cleared and then stopped.

Entering the IDLE0 Mode

There is only one way for the device to enter the IDLE0 Mode and that is to execute the "HALT" instruction in the application program with the FHIDEN bit in the SCC register equal to "0" and the FSIDEN bit in the SCC register equal to "1". When this instruction is executed under the conditions described above, the following will occur:

- The f_H clock will be stopped and the application program will stop at the "HALT" instruction, but the f_{SUB} clock will be on.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag PDF will be set, and WDT timeout flag TO will be cleared.
- The WDT will be cleared and resume counting if the WDT function is enabled. If the WDT function is disabled, the WDT will be cleared and then stopped.

Entering the IDLE1 Mode

There is only one way for the device to enter the IDLE1 Mode and that is to execute the "HALT" instruction in the application program with both the FHIDEN and FSIDEN bits in the SCC register equal to "1". When this instruction is executed under the conditions described above, the following will occur:

- The f_H and f_{SUB} clocks will be on but the application program will stop at the "HALT" instruction.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag PDF will be set, and WDT timeout flag TO will be cleared.
- The WDT will be cleared and resume counting if the WDT function is enabled. If the WDT function is disabled, the WDT will be cleared and then stopped.

Entering the IDLE2 Mode

There is only one way for the device to enter the IDLE2 Mode and that is to execute the "HALT" instruction in the application program with the FHIDEN bit in the SCC register equal to "1" and the FSIDEN bit in the SCC register equal to "0". When this instruction is executed under the conditions described above, the following will occur:

- The f_H clock will be on but the f_{SUB} clock will be off and the application program will stop at the "HALT" instruction.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting if the WDT function is enabled. If the WDT function is disabled, the WDT will be cleared and then stopped.

Standby Current Considerations

As the main reason for entering the SLEEP or IDLE Mode is to keep the current consumption of the device to as low a value as possible, perhaps only in the order of several micro-amps except in the IDLE1 and IDLE2 Mode, there are other considerations which must also be taken into account by the circuit designer if the power consumption is to be minimised. Special attention must be made to the I/O pins on the device. All high-impedance input pins must be connected to either a fixed high or low level as any floating input pins could create internal oscillations and result in increased current consumption. This also applies to the device which has different package types, as there may be unbonded pins. These must either be set as outputs or if set as inputs must have pull-high resistors connected.

Care must also be taken with the loads, which are connected to I/O pins, which are set as outputs. These should be placed in a condition in which minimum current is drawn or connected only to external circuits that do not draw current, such as other CMOS inputs. Also note that additional standby current will also be required if the LXT or LIRC oscillator has enabled via configuration option.

In the IDLE1 and IDLE2 Mode the system oscillator is on, if the peripheral function clock source is derived from the high speed oscillator, the additional standby current will also be perhaps in the order of several hundred micro-amps.

Wake-up

To minimise power consumption the device can enter the SLEEP or any IDLE Mode, where the CPU will be switched off. However, when the device is woken up again, it will take a considerable time for the original system oscillator to restart, stablise and allow normal operation to resume.

After the system enters the SLEEP or IDLE Mode, it can be woken up from one of various sources listed as follows:

- An external falling edge on Port A
- A system interrupt
- A WDT overflow

When the device executes the “HALT” instruction, it will enter the IDLE or SLEEP mode and the PDF flag will be set to 1. The PDF flag will be cleared to 0 if the device experiences a system power-up or executes the clear Watchdog Timer instruction. If the system is woken up by a WDT overflow, a Watchdog Timer reset will be initiated and the TO flag will be set to 1. The TO flag is set if a WDT time-out occurs and causes a wake-up that only resets the Program Counter and Stack Pointer, other flags remain in their original status.

Each pin on Port A can be set using the PAWU register to permit a negative transition on the pin to wake up the system. When a Port A pin wake-up occurs, the program will resume execution at the instruction following the “HALT” instruction. If the system is woken up by an interrupt, then two possible situations may occur. The first is where the related interrupt is disabled or the interrupt is enabled but the stack is full, in which case the program will resume execution at the instruction following the “HALT” instruction. In this situation, the interrupt which woke up the device will not be immediately serviced, but will rather be serviced later when the related interrupt is finally enabled or when a stack level becomes free. The other situation is where the related interrupt is enabled and the stack is not full, in which case the regular interrupt response takes place. If an interrupt request flag is set high before entering the SLEEP or IDLE Mode, the wake-up function of the related interrupt will be disabled.

Watchdog Timer

The Watchdog Timer is provided to prevent program malfunctions or sequences from jumping to unknown locations, due to certain uncontrollable external events such as electrical noise.

Watchdog Timer Clock Source

The Watchdog Timer clock source is provided by the internal clock, f_{LIRC}, which is sourced from the LIRC oscillator. The LIRC internal oscillator has an approximate frequency of 32kHz and this specified internal clock period can vary with V_{DD}, temperature and process variations. The Watchdog Timer source clock is then subdivided by a ratio of 2^8 to 2^{18} to give longer timeouts, the actual value being chosen using the WS2~WS0 bits in the WDTC register.

Watchdog Timer Control Register

A single register, WDTC, controls the required time-out period as well as the Watchdog Timer the enable/disable and the MCU reset operation.

• **WDTC Register**

Bit	7	6	5	4	3	2	1	0
Name	WE4	WE3	WE2	WE1	WE0	WS2	WS1	WS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	1	0	1	0	0	1	1

Bit 7~3 **WE4~WE0**: WDT function software control
10101: Disable
01010: Enable
Other values: Reset MCU
When these bits are changed to any other values due to environmental noise the microcontroller will be reset; this reset operation will be activated after a delay time, t_{SRESET}, and the WRF bit in the RSTFC register will be set high.

Bit 2~0 **WS2~WS0**: WDT time-out period selection
000: $2^8/f_{LIRC}$
001: $2^{10}/f_{LIRC}$
010: $2^{12}/f_{LIRC}$
011: $2^{14}/f_{LIRC}$
100: $2^{15}/f_{LIRC}$
101: $2^{16}/f_{LIRC}$
110: $2^{17}/f_{LIRC}$
111: $2^{18}/f_{LIRC}$
These three bits determine the division ratio of the Watchdog Timer source clock, which in turn determines the timeout period.

• **RSTFC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	RSTF	LVRF	LRF	WRF
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	x	0	0

"x": unknown

Bit 7~4 Unimplemented, read as "0"

Bit 3 **RSTF**: Reset control register software reset flag
Refer to Internal Reset Control section.

Bit 2 **LVRF**: LVR function reset flag
Refer to the Low Voltage Reset section.

Bit 1 **LRF**: LVR control register software reset flag
Refer to the Low Voltage Reset section.

Bit 0 **WRF**: WDT control register software reset flag
0: Not occurred
1: Occurred
This bit is set high by the WDT Control register software reset and cleared to zero by the application program. Note that this bit can only be cleared to zero by the application program.

Watchdog Timer Operation

The Watchdog Timer operates by providing a device reset when its timer overflows. This means that in the application program and during normal operation the user has to strategically clear the Watchdog Timer before it overflows to prevent the Watchdog Timer from executing a reset. This is done using the clear watchdog instruction. If the program malfunctions for whatever reason, jumps to an unknown location, or enters an endless loop, the clear instruction will not be executed in the correct manner, in which case the Watchdog Timer will overflow and reset the device. There are five bits, WE4~WE0, in the WDTC register to offer the enable/disable control of the Watchdog Timer and the MCU reset. The WDT function will be disabled when the WE4~WE0 bits are set to a value of 10101B while the WDT function will be enabled if the WE4~WE0 bits are equal to 01010B. If the WE4~WE0 bits are set to any other values, other than 01010B and 10101B, it will reset the device after a delay time, t_{SRESET}. After power-on these bits will have the value of 01010B.

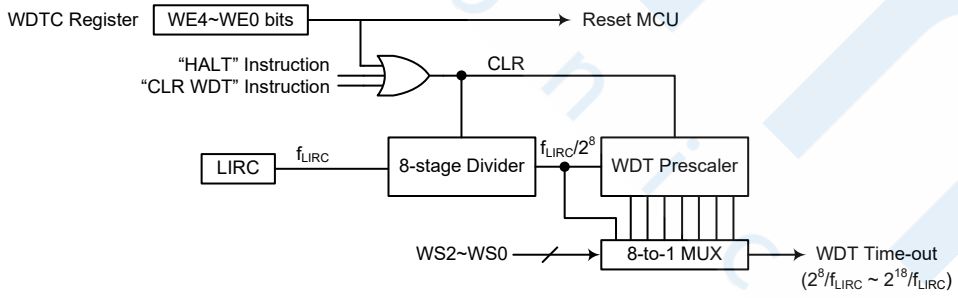
WE4~WE0 Bits	WDT Function
10101B	Disable
01010B	Enable
Any other value	Reset MCU

Watchdog Timer Function Control

Under normal program operation, a Watchdog Timer time-out will initialise a device reset and set the status bit TO. However, if the system is in the SLEEP or IDLE Mode, when a Watchdog Timer time-out occurs, the TO bit in the status register will be set and only the Program Counter and Stack Pointer will be reset. Three methods can be adopted to clear the contents of the Watchdog Timer. The first is a WDTC software reset, which means a certain value except 01010B and 10101B written into the WE4~WE0 bit, the second is using the Watchdog Timer software clear instruction and the third is via a HALT instruction.

There is only one method of using software instruction to clear the Watchdog Timer. That is to use the single "CLR WDT" instruction to clear the WDT.

The maximum time-out period is when the 2^{18} division ratio is selected. As an example, with a 32kHz LIRC oscillator as its source clock, this will give a maximum watchdog period of around 8 seconds for the 2^{18} division ratio, and a minimum timeout of 8ms for the 2^8 division ration.



Watchdog Timer

Reset and Initialisation

A reset function is a fundamental part of any microcontroller ensuring that the device can be set to some predetermined condition irrespective of outside parameters. The most important reset condition is after power is first applied to the microcontroller. In this case, internal circuitry will ensure that the microcontroller, after a short delay, will be in a well-defined state and ready to execute the first program instruction. After this power-on reset, certain important internal registers will be set to defined states before the program commences. One of these registers is the Program Counter, which will be reset to zero forcing the microcontroller to begin program execution from the lowest Program Memory address.

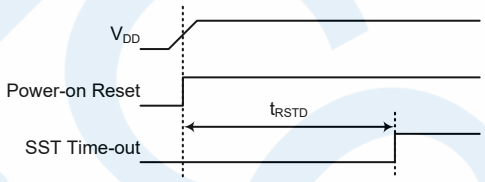
In addition to the power-on reset, another reset exists in the form of a Low Voltage Reset, LVR, where a full reset is implemented in situations where the power supply voltage falls below a certain threshold. Another type of reset is when the Watchdog Timer overflows and resets the microcontroller. All types of reset operations result in different register conditions being setup.

Reset Functions

There are several ways in which a microcontroller reset can occur, through events occurring internally.

Power-on Reset

The most fundamental and unavoidable reset is the one that occurs after power is first applied to the microcontroller. As well as ensuring that the Program Memory begins execution from the first memory address, a power-on reset also ensures that certain other registers are preset to known conditions. All the I/O port and port control registers will power up in a high condition ensuring that all pins will be first set to inputs.



Power-On Reset Timing Chart

Internal Reset Control

There is an internal reset control register, RSTC, which is used to provide a reset when the device operates abnormally due to the environmental noise interference. If the content of the RSTC register is set to any value other than 01010101B or 10101010B, it will reset the device after a delay time, t_{SRESET}. After power-on the register will have a value of 01010101B.

RSTC7~RSTC0 Bits	Reset Function
01010101B	No operation
10101010B	No operation
Any other value	Reset MCU

Internal Reset Function Control

• **RSTC Register**

Bit	7	6	5	4	3	2	1	0
Name	RSTC7	RSTC6	RSTC5	RSTC4	RSTC3	RSTC2	RSTC1	RSTC0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	1	0	1	0	1	0	1

Bit 7~0 **RSTC7~RSTC0**: Reset function control
01010101: No operation
10101010: No operation
Other values: Reset MCU
If these bits are changed due to adverse environmental conditions, the microcontroller will be reset. The reset operation will be activated after a delay time, t_{SRESET} and the RSTF bit in the RSTFC register will be set to 1. All resets will reset this register to POR value except the WDT time-out reset during SLEEP/IDLE mode.

• **RSTFC Register**

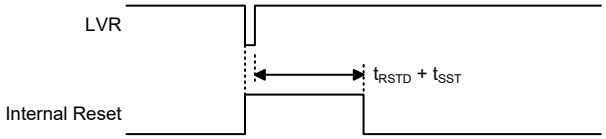
Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	RSTF	LVRF	LRF	WRF
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	x	0	0

"x": unknown

Bit 7~4 Unimplemented, read as "0"
Bit 3 **RSTF**: Reset control register software reset flag
0: Not occurred
1: Occurred
This bit is set high by the RSTC control register software reset and cleared to zero by the application program. Note that this bit can only be cleared to 0 by the application program.
Bit 2 **LVRF**: LVR function reset flag
Refer to the Low Voltage Reset section.
Bit 1 **LRF**: LVR control register software reset flag
Refer to the Low Voltage Reset section.
Bit 0 **WRF**: WDT control register software reset flag
Refer to the Watchdog Timer Control Register section.

Low Voltage Reset – LVR

The microcontroller contains a low voltage reset circuit in order to monitor the supply voltage of the device and provides an MCU reset should the value fall below a certain predefined level. The LVR function can be enabled or disabled by the LVRC control register. If the LVRC control register is configured to enable the LVR function, the LVR function will be always enabled except in the SLEEP or IDLE mode. If the supply voltage of the device drops to within a range of 0.9V~V_{LVR} such as might occur when changing the battery in battery powered applications, the LVR will automatically reset the device internally and the LVRF bit in the RSTFC register will also be set high. For a valid LVR signal, a low supply voltage, i.e., a voltage in the range between 0.9V~V_{LVR} must exist for a time greater than that specified by t_{LVR} in the LVD/LVR Electrical characteristics. If the low supply voltage state does not exceed this value, the LVR will ignore the low supply voltage and will not perform a reset function. The actual V_{LVR} value can be selected by the LVS7~LVS0 bits in the LVRC register. If the LVS7~LVS0 bits are changed to some different values by environmental noise, the LVR will reset the device after a delay time, t_{SRESET}. When this happens, the LRF bit in the RSTFC register will be set high. After power-on the register will have the value of 01100110B. Note that the LVR function will be automatically disabled when the device enters the IDLE or SLEEP mode.



Low Voltage Reset Timing Chart

• **LVRC Register**

Bit	7	6	5	4	3	2	1	0
Name	LVS7	LVS6	LVS5	LVS4	LVS3	LVS2	LVS1	LVS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	1	1	0	0	1	1	0

Bit 7~0 **LVS7~LVS0**: LVR voltage selection
01100110: 1.7V
01010101: 1.9V
00110011: 2.55V
10011001: 3.15V
10101010: Reserved
11110000: LVR disable
Any other value: Generates MCU reset – register is reset to POR value

When an actual low voltage condition as specified above occurs, an MCU reset will be generated. The reset operation will be activated after the low voltage condition keeps more than a t_{LVR} time. In this situation the register contents will remain the same after such a reset occurs.

Any register value, other than the four defined LVR values and 10101010B/11110000B above, will also result in the generation of an MCU reset. The reset operation will be activated after a delay time, t_{SRESET}. However in this situation the register contents will be reset to the POR value.

• **RSTFC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	RSTF	LVRF	LRF	WRF
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	x	0	0

"x": unknown

Bit 7~4 Unimplemented, read as "0"

Bit 3 **RSTF**: Reset control register software reset flag
Refer to the Internal Reset Control section.

Bit 2 **LVRF**: LVR function reset flag
0: Not occur
1: Occurred
This bit is set high when a specific low voltage reset situation condition occurs. This bit can only be cleared to zero by the application program.

Bit 1 **LRF**: LVR control register software reset flag
0: Not occurred
1: Occurred
This bit is set high if the LVRC register containing any non-defined LVR voltage register values. This in effect acts like a software reset function. This bit can only be cleared to zero by the application program.

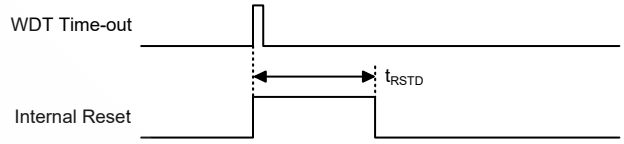
Bit 0 **WRF**: WDT control register software reset flag
Refer to the Watchdog Timer Control Register section.

IAP Reset

When a specific value of "55H" is written into the FC1 register, a reset signal will be generated to reset the whole device. Refer to the In Application Programming section for more associated details.

Watchdog Time-out Reset during Normal Operation

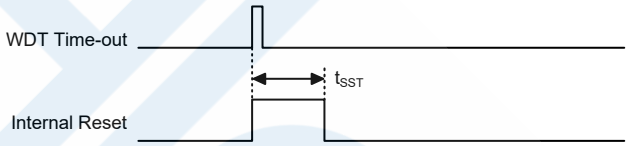
When the Watchdog time-out Reset during normal operations in the FAST or SLOW mode occurs, the Watchdog time-out flag TO will be set to "1".



WDT Time-out Reset during Normal Operation Timing Chart

Watchdog Time-out Reset during SLEEP or IDLE Mode

The Watchdog time-out Reset during SLEEP or IDLE Mode is a little different from other kinds of reset. Most of the conditions remain unchanged except that the Program Counter and the Stack Pointer will be cleared to "0" and the TO and PDF flags will be set to "1". Refer to the System Start Up Time Characteristics for t_{SST} details.



WDT Time-out Reset during SLEEP or IDLE Timing Chart

Reset Initial Conditions

The different types of reset described affect the reset flags in different ways. These flags, known as PDF and TO are located in the status register and are controlled by various microcontroller operations, such as the SLEEP or IDLE Mode function or Watchdog Timer. The reset flags are shown in the table:

TO	PDF	Reset Conditions
0	0	Power-on reset
u	u	LVR reset during FAST or SLOW Mode operation
1	u	WDT time-out reset during FAST or SLOW Mode operation
1	1	WDT time-out reset during IDLE or SLEEP Mode operation

"u" stands for unchanged

The following table indicates the way in which the various components of the microcontroller are affected after a power-on reset occurs.

Item	Condition after Reset
Program Counter	Reset to zero
Interrupts	All interrupts will be disabled
WDT, Time Bases	Cleared after reset, WDT begins counting
Timer Modules	Timer Modules will be turned off
Input/Output Ports	I/O ports will be set as inputs
Stack Pointer	Stack Pointer will point to the top of the stack

The different kinds of resets all affect the internal registers of the microcontroller in different ways. To ensure reliable continuation of normal program execution after a reset occurs, it is important to

know what condition the microcontroller is in after a particular reset occurs. The following table describes how each type of reset affects each of the microcontroller internal registers.

Name	Power-On Reset	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
IAR0	0000 0000	0000 0000	uuuu uuuu
MP0	0000 0000	0000 0000	uuuu uuuu
IAR1	0000 0000	0000 0000	uuuu uuuu
MP1L	0000 0000	0000 0000	uuuu uuuu
MP1H	0000 0000	0000 0000	uuuu uuuu
ACC	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCL	0000 0000	0000 0000	0000 0000
TBLP	xxxx xxxx	uuuu uuuu	uuuu uuuu
TBLH	xxxx xxxx	uuuu uuuu	uuuu uuuu
TBHP	--xx xxxx	--uu uuuu	--uu uuuu
STATUS	xx00 xxxx	uu1u uuuu	uu11 uuuu
PBP	---- ---0	---- ---0	---- ---u
IAR2	0000 0000	0000 0000	uuuu uuuu
MP2L	0000 0000	0000 0000	uuuu uuuu
MP2H	0000 0000	0000 0000	uuuu uuuu
RSTFC	---- 0x00	---- uuuu	---- uuuu
INTEG	--00 0000	--00 0000	--uu uuuu
INTC0	-000 0000	-000 0000	-uuu uuuu
INTC1	0000 0000	0000 0000	uuuu uuuu
INTC2	0000 0000	0000 0000	uuuu uuuu
PA	1111 1111	1111 1111	uuuu uuuu
PAC	1111 1111	1111 1111	uuuu uuuu
PAPU	0000 0000	0000 0000	uuuu uuuu
PAWU	0000 0000	0000 0000	uuuu uuuu
LVDC	--00 -000	--00 -000	--uu -uuu
SCC	000- 0000	000- 0000	uuu- uuuu
WDTC	0101 0011	0101 0011	uuuu uuuu
TB0C	0--- -000	0--- -000	u--- -uuu
HIRCC	---- 0001	---- 0001	---- uuuu
LVRC	0110 0110	0110 0110	uuuu uuuu
INTC3	--00 --00	--00 --00	--uu --uu
SADOL	xxxx ----	xxxx ----	uuuu ---- (ADRFS=0) uuuu uuuu (ADRFS=1)
SADOH	xxxx xxxx	xxxx xxxx	uuuu uuuu (ADRFS=0) ---- uuuu (ADRFS=1)
SADC0	0000 0000	0000 0000	uuuu uuuu
SADC1	0000 -000	0000 -000	uuuu -uuu
SADC2	0--0 0000	0--0 0000	u--u uuuu
PB	-111 1111	-111 1111	-uuu uuuu
PBC	-111 1111	-111 1111	-uuu uuuu
PBPU	-000 0000	-000 0000	-uuu uuuu
CTMC0	0000 0---	0000 0---	uuuu u---

Name	Power-On Reset	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
CTMC1	0000 0000	0000 0000	uuuu uuuu
CTMDL	0000 0000	0000 0000	uuuu uuuu
CTMDH	0000 0000	0000 0000	uuuu uuuu
CTMAL	0000 0000	0000 0000	uuuu uuuu
CTMAH	0000 0000	0000 0000	uuuu uuuu
CTMRP	0000 0000	0000 0000	uuuu uuuu
STMC0	0000 0---	0000 0---	uuuu u---
STMC1	0000 0000	0000 0000	uuuu uuuu
STMDL	0000 0000	0000 0000	uuuu uuuu
STMDH	0000 0000	0000 0000	uuuu uuuu
STMAL	0000 0000	0000 0000	uuuu uuuu
STMAH	0000 0000	0000 0000	uuuu uuuu
STMRP	0000 0000	0000 0000	uuuu uuuu
HXTC	---- -000	---- -000	---- -uuu
PTM0C0	0000 0---	0000 0---	uuuu u---
PTM0C1	0000 0000	0000 0000	uuuu uuuu
PTM0DL	0000 0000	0000 0000	uuuu uuuu
PTM0DH	---- --00	---- --00	---- --uu
PTM0AL	0000 0000	0000 0000	uuuu uuuu
PTM0AH	---- --00	---- --00	---- --uu
PTM0RPL	0000 0000	0000 0000	uuuu uuuu
PTM0RPH	---- --00	---- --00	---- --uu
CMPC	-000 ---1	-000 ---1	-uuu ---u
PC	-111 1111	-111 1111	-uuu uuuu
PCC	-111 1111	-111 1111	-uuu uuuu
PCPU	-000 0000	-000 0000	-uuu uuuu
LXTC	---- -000	---- -000	---- -uuu
SIMC0	111- 0000	111- 0000	uuu- uuuu
SIMC1	1000 0001	1000 0001	uuuu uuuu
SIMD	xxxx xxxx	xxxx xxxx	uuuu uuuu
SIMA/SIMC2	0000 0000	0000 0000	uuuu uuuu
SIMTOC	0000 0000	0000 0000	uuuu uuuu
VBGRC	---- ---0	---- ---0	---- ---u
TB1C	0--- -000	0--- -000	u--- -uuu
MFI0	0000 0000	0000 0000	uuuu uuuu
MFI1	0000 0000	0000 0000	uuuu uuuu
MFI2	--00 --00	--00 --00	--uu --uu
SLEDC0	0000 0000	0000 0000	uuuu uuuu
SLEDC1	0000 0000	0000 0000	uuuu uuuu
IFS0	0000 0000	0000 0000	uuuu uuuu
PD	1111 1111	1111 1111	uuuu uuuu
PDC	1111 1111	1111 1111	uuuu uuuu
PDPU	0000 0000	0000 0000	uuuu uuuu
ORMC	0000 0000	0000 0000	0000 0000
SLCDS3	--00 0000	--00 0000	--uu uuuu
IFS1	---- ---0	---- ---0	---- ---u
PSD1	0000 0000	0000 0000	uuuu uuuu

Name	Power-On Reset	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
LVPUC	---- ---0	---- ---0	---- ---u
PAS0	0000 0000	0000 0000	uuuu uuuu
PAS1	0000 0000	0000 0000	uuuu uuuu
PBS0	0000 0000	0000 0000	uuuu uuuu
PBS1	--00 0000	--00 0000	--uu uuuu
PCS0	0000 0000	0000 0000	uuuu uuuu
PCS1	--00 0000	--00 0000	--uu uuuu
PDS0	0000 0000	0000 0000	uuuu uuuu
SLCDC0	0000 ----	0000 ----	uuuu ----
SLCDS0	0000 0000	0000 0000	uuuu uuuu
SLCDS1	0000 0000	0000 0000	uuuu uuuu
SLCDS2	0000 0000	0000 0000	uuuu uuuu
PSC0R	---- --00	---- --00	---- --uu
PSC1R	---- --00	---- --00	---- --uu
RSTC	0101 0101	0101 0101	uuuu uuuu
FC0	0000 0000	0000 0000	uuuu uuuu
FC1	0000 0000	0000 0000	uuuu uuuu
FC2	---- --00	---- --00	---- --uu
FARL	0000 0000	0000 0000	uuuu uuuu
FARH	--00 0000	--00 0000	--uu uuuu
FD0L	0000 0000	0000 0000	uuuu uuuu
FD0H	0000 0000	0000 0000	uuuu uuuu
FD1L	0000 0000	0000 0000	uuuu uuuu
FD1H	0000 0000	0000 0000	uuuu uuuu
FD2L	0000 0000	0000 0000	uuuu uuuu
FD2H	0000 0000	0000 0000	uuuu uuuu
FD3L	0000 0000	0000 0000	uuuu uuuu
FD3H	0000 0000	0000 0000	uuuu uuuu
EEAL	0000 0000	0000 0000	uuuu uuuu
EEAH	---- --00	---- --00	---- --uu
EED	0000 0000	0000 0000	uuuu uuuu
PTM1C0	0000 0---	0000 0---	uuuu u---
PTM1C1	0000 0000	0000 0000	uuuu uuuu
PTM1DL	0000 0000	0000 0000	uuuu uuuu
PTM1DH	---- --00	---- --00	---- --uu
PTM1AL	0000 0000	0000 0000	uuuu uuuu
PTM1AH	---- --00	---- --00	---- --uu
PTM1RPL	0000 0000	0000 0000	uuuu uuuu
PTM1RPH	---- --00	---- --00	---- --uu
EEC	0000 0000	0000 0000	uuuu uuuu
U0SR	0000 1011	0000 1011	uuuu uuuu
U0CR1	0000 00x0	0000 00x0	uuuu uuuu
U0CR2	0000 0000	0000 0000	uuuu uuuu
U0CR3	---- ---0	---- ---0	---- ---u
BRDH0	0000 0000	0000 0000	uuuu uuuu
BRDL0	0000 0000	0000 0000	uuuu uuuu
UFCR0	--00 0000	--00 0000	--uu uuuu

Name	Power-On Reset	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
TXR_RXR0	xxxx xxxx	xxxx xxxx	uuuu uuuu
RxCNT0	---- -000	---- -000	---- -uuu
U1SR	0000 1011	0000 1011	uuuu uuuu
U1CR1	0000 00x0	0000 00x0	uuuu uuuu
U1CR2	0000 0000	0000 0000	uuuu uuuu
U1CR3	---- ---0	---- ---0	---- ---u
BRDH1	0000 0000	0000 0000	uuuu uuuu
BRDL1	0000 0000	0000 0000	uuuu uuuu
UFCR1	--00 0000	--00 0000	--uu uuuu
TXR_RXR1	xxxx xxxx	xxxx xxxx	uuuu uuuu
RxCNT1	---- -000	---- -000	---- -uuu

Note: "u" stands for unchanged
"x" stands for unknown
"-" stands for unimplemented

Input/Output Ports

Holtek microcontrollers offer considerable flexibility on their I/O ports. With the input or output designation of every pin fully under user program control, pull-high selections for all ports and wake-up selections on certain pins, the user is provided with an I/O structure to meet the needs of a wide range of application possibilities.

The device provides bidirectional input/output lines labeled with port name PA~PD. These I/O ports are mapped to the RAM Data Memory with specific addresses as shown in the Special Purpose Data Memory table. All of these I/O ports can be used for input and output operations. For input operation, these ports are non-latching, which means the inputs must be ready at the T2 rising edge of instruction "MOV A, [m]", where m denotes the port address. For output operation, all the data is latched and remains unchanged until the output latch is rewritten.

Register Name	Bit							
	7	6	5	4	3	2	1	0
PA	PA7	PA6	PA5	PA4	PA3	PA2	PA1	PA0
PAC	PAC7	PAC6	PAC5	PAC4	PAC3	PAC2	PAC1	PAC0
PAPU	PAPU7	PAPU6	PAPU5	PAPU4	PAPU3	PAPU2	PAPU1	PAPU0
PAWU	PAWU7	PAWU6	PAWU5	PAWU4	PAWU3	PAWU2	PAWU1	PAWU0
PB	—	PB6	PB5	PB4	PB3	PB2	PB1	PB0
PBC	—	PBC6	PBC5	PBC4	PBC3	PBC2	PBC1	PBC0
PBPU	—	PBPU6	PBPU5	PBPU4	PBPU3	PBPU2	PBPU1	PBPU0
PC	—	PC6	PC5	PC4	PC3	PC2	PC1	PC0
PCC	—	PCC6	PCC5	PCC4	PCC3	PCC2	PCC1	PCC0
PCPU	—	PCPU6	PCPU5	PCPU4	PCPU3	PCPU2	PCPU1	PCPU0
PD	PD7	PD6	PD5	PD4	PD3	PD2	PD1	PD0
PDC	PDC7	PDC6	PDC5	PDC4	PDC3	PDC2	PDC1	PDC0
PDPU	PDPU7	PDPU6	PDPU5	PDPU4	PDPU3	PDPU2	PDPU1	PDPU0

"—": Unimplemented, read as "0"

I/O Logic Function Register List

Pull-high Resistors

Many product applications require pull-high resistors for their switch inputs usually requiring the use of an external resistor. To eliminate the need for these external resistors, all I/O pins, when configured as a digital input have the capability of being connected to an internal pull-high resistor. These pull-high resistors are selected using the LVPUC and PxPU registers, and are implemented using weak PMOS transistors. The PxPU register is used to determine whether the pull-high function is enabled or not while the LVPUC register is used to select the pull-high resistors value for low voltage power supply applications.

Note that the pull-high resistor can be controlled by the relevant pull-high control register only when the pin-shared functional pin is selected as a digital input or NMOS output. Otherwise, the pull-high resistors can not be enabled.

• **PxPU Register**

Bit	7	6	5	4	3	2	1	0
Name	PxPU7	PxPU6	PxPU5	PxPU4	PxPU3	PxPU2	PxPU1	PxPU0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

PxPUn: I/O Port x Pin pull-high function control
0: Disable
1: Enable

The PxPUn bit is used to control the pin pull-high function. Here the "x" can be A, B, C or D. However, the actual available bits for each I/O Port may be different.

Special attention should be paid to the unbounded lines, PC3, PC6, PD3~PD4 and PD6~PD7. The relevant PxPUn bit should be fixed at 0 and the corresponding line should be set as an ouput by clearing the PxCn bit to zero to avoid unwanted power consumption result from floating input conditions. .

• **LVPUC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	LVPU
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

Bit 7~1 Unimplemented, read as "0"

Bit 0 **LVPU**: Pull-high resistor selection when low voltage power supply
0: All pin pull-high resistors are 60kΩ @ 3V
1: All pin pull-high resistors are 15kΩ @ 3V

This bit is used to select the pull-high resistor value for low voltage power supply applications. The LVPU bit is only available when the corresponding pin pull-high function is enabled by setting the relevant pull-high control bit high. This bit will have no effect when the pull-high function is disabled.

Port A Wake-up

The HALT instruction forces the microcontroller into the SLEEP or IDLE Mode which preserves power, a feature that is important for battery and other low-power applications. Various methods exist to wake up the microcontroller, one of which is to change the logic condition on one of the Port A pins from high to low. This function is especially suitable for applications that can be woken up via external switches. Each pin on Port A can be selected individually to have this wake-up feature using the PAWU register.

Note that the wake-up function can be controlled by the wake-up control register only when the pin is selected as a general purpose input and the MCU enters the IDLE or SLEEP mode.

• **PAWU Register**

Bit	7	6	5	4	3	2	1	0
Name	PAWU7	PAWU6	PAWU5	PAWU4	PAWU3	PAWU2	PAWU1	PAWU0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **PAWU7~PAWU0**: PA7~PA0 wake-up function control
0: Disable
1: Enable

I/O Port Control Registers

Each I/O port has its own control register known as PAC~PDC, to control the input/output configuration. With this control register, each CMOS output or input can be reconfigured dynamically under software control. Each pin of the I/O ports is directly mapped to a bit in its associated port control register. For the I/O pin to function as an input, the corresponding bit of the control register must be written as a "1". This will then allow the logic state of the input pin to be directly read by instructions. When the corresponding bit of the control register is written as a "0", the I/O pin will be set as a CMOS output. If the pin is currently set as an output, instructions can still be used to read the output register. However, it should be noted that the program will in fact only read the status of the output data latch and not the actual logic status of the output pin.

• **PxC Register**

Bit	7	6	5	4	3	2	1	0
Name	PxC7	PxC6	PxC5	PxC4	PxC3	PxC2	PxC1	PxC0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	1	1	1	1	1	1	1	1

PxCn: I/O Port x Pin type selection
0: Output
1: Input

The PxCn bit is used to control the pin type selection. Here the "x" can be A, B, C or D. However, the actual available bits for each I/O Port may be different.

Special attention should be paid to the unbounded lines, PC3, PC6, PD3~PD4 and PD6~PD7. The relevant PxCn bit should be fixed at 0 to avoid unwanted power consumption result from floating input conditions. .

I/O Port Source Current Selection

The device supports different output source current driving capability for each I/O port. With the selection register, SLEDCn, specific I/O port can support four levels of the source current driving capability. These source current selection bits are available only when the corresponding pin is configured as a CMOS output. Otherwise, these select bits have no effect. Users should refer to the Input/Output Characteristics section to select the desired output source current for different applications.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SLEDC0	SLEDC07	SLEDC06	SLEDC05	SLEDC04	SLEDC03	SLEDC02	SLEDC01	SLEDC00
SLEDC1	SLEDC17	SLEDC16	SLEDC15	SLEDC14	SLEDC13	SLEDC12	SLEDC11	SLEDC10

I/O Port Source Current Control Register List

• **SLEDC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	SLEDC07	SLEDC06	SLEDC05	SLEDC04	SLEDC03	SLEDC02	SLEDC01	SLEDC00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **SLEDC07~SLEDC06**: PB6~PB4 Source Current Selection
00: Source current=Level 0 (min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (max.)

Bit 5~4 **SLEDC05~SLEDC04**: PB3~PB0 Source Current Selection
00: Source current=Level 0 (min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (max.)

Bit 3~2 **SLEDC03~SLEDC02**: PA7~PA4 Source Current Selection
00: Source current=Level 0 (min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (max.)

Bit 1~0 **SLEDC01~SLEDC00**: PA3~PA0 Source Current Selection
00: Source current=Level 0 (min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (max.)

• **SLEDC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	SLEDC17	SLEDC16	SLEDC15	SLEDC14	SLEDC13	SLEDC12	SLEDC11	SLEDC10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **SLEDC17~SLEDC16**: PD5 Source Current Selection
00: Source current=Level 0 (min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (max.)

Bit 5~4 **SLEDC15~SLEDC14**: PD2~PD0 Source Current Selection
00: Source current=Level 0 (min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (max.)

Bit 3~2 **SLEDC13~SLEDC12**: PC5~PC4 Source Current Selection
00: Source current=Level 0 (min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (max.)

Bit 1~0 **SLEDC11~SLEDC10**: PC2~PC0 Source Current Selection
00: Source current=Level 0 (min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (max.)

Pin-shared Functions

The flexibility of the microcontroller range is greatly enhanced by the use of pins that have more than one function. Limited numbers of pins can force serious design constraints on designers but by supplying pins with multi-functions, many of these difficulties can be overcome. For these pins, the desired function of the multi-function I/O pins is selected by a series of registers via the application program control.

Pin-shared Function Selection Registers

The limited number of supplied pins in a package can impose restrictions on the amount of functions a certain device can contain. However by allowing the same pins to share several different functions and providing a means of function selection, a wide range of different functions can be incorporated into even relatively small package sizes. The device includes Port "x" output function Selection register "n", labeled as PxSn, and Input Function Selection register, labeled as IFSi, which can select the desired functions of the multi-function pin-shared pins.

The most important point to note is to make sure that the desired pin-shared function is properly selected and also deselected. For most pin-shared functions, to select the desired pin-shared function, the pin-shared function should first be correctly selected using the corresponding pin-shared control register. After that the corresponding peripheral functional setting should be configured and then the peripheral function can be enabled. However, a special point must be noted for some digital input pins, such as INTn, xTCKn, etc, which share the same pin-shared control configuration with their corresponding general purpose I/O functions when setting the relevant pin-shared control bit fields. To select these pin functions, in addition to the necessary pin-shared control and peripheral functional setup aforementioned, they must also be setup as an input by setting the corresponding bit in the I/O port control register. To correctly deselect the pin-shared function, the peripheral function should first be disabled and then the corresponding pin-shared function control register can be modified to select other pin-shared functions.

Register Name	Bit							
	7	6	5	4	3	2	1	0
PAS0	PAS07	PAS06	PAS05	PAS04	PAS03	PAS02	PAS01	PAS00
PAS1	PAS17	PAS16	PAS15	PAS14	PAS13	PAS12	PAS11	PAS10
PBS0	PBS07	PBS06	PBS05	PBS04	PBS03	PBS02	PBS01	PBS00
PBS1	—	—	PBS15	PBS14	PBS13	PBS12	PBS11	PBS10
PCS0	PCS07	PCS06	PCS05	PCS04	PCS03	PCS02	PCS01	PCS00
PCS1	—	—	PCS15	PCS14	PCS13	PCS12	PCS11	PCS10
PDS0	PDS07	PDS06	PDS05	PDS04	PDS03	PDS02	PDS01	PDS00
PDS1	PDS17	PDS16	PDS15	PDS14	PDS13	PDS12	PDS11	PDS10
IFS0	RX1PS	INT2PS	INT1PS	INT0PS	SDI_SDAPS	SCK_SCLPS	SCSBPS	RX0PS
IFS1	—	—	—	—	—	—	—	PTCK1PS

Pin-shared Function Selection Register List

• **PAS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PAS07	PAS06	PAS05	PAS04	PAS03	PAS02	PAS01	PAS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PAS07~PAS06**: PA3 Pin-Shared Function Selection
00: PA3
01: SDI/SDA
10: CMPO
11: SCOM3/SSEG3

Bit 5~4 **PAS05~PAS04**: PA2 Pin-Shared Function Selection
00: PA2/INT1
01: PA2/INT1
10: PA2/INT1
11: SCOM2/SSEG2

Bit 3~2 **PAS03~PAS02**: PA1 Pin-Shared Function Selection
00: PA1/INT0
01: PA1/INT0
10: SDO
11: SCOM1/SSEG1

Bit 1~0 **PAS01~PAS00**: PA0 Pin-Shared Function Selection
00: PA0
01: PA0
10: STP
11: SCOM0/SSEG0

• **PAS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	PAS17	PAS16	PAS15	PAS14	PAS13	PAS12	PAS11	PAS10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PAS17~PAS16**: PA7 Pin-Shared Function Selection
00: PA7
01: PTP0
10: SCOM7/SSEG7
11: AN6

Bit 5~4 **PAS15~PAS14**: PA6 Pin-Shared Function Selection
00: PA6/CTCK
01: SCOM6/SSEG6
10: AN5
11: VREF

Bit 3~2 **PAS13~PAS12**: PA5 Pin-Shared Function Selection
00: PA5
01: SCOM5/SSEG5
10: AN4
11: VREFI

Bit 1~0 **PAS11~PAS10**: PA4 Pin-Shared Function Selection
00: PA4/PTCK0
01: PA4/PTCK0
10: SCOM4/SSEG4
11: AN3

• **PBS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PBS07	PBS06	PBS05	PBS04	PBS03	PBS02	PBS01	PBS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PBS07~PBS06**: PB3 Pin-Shared Function Selection
00: PB3
01: CTP
10: SCOM11/SSEG11
11: AN7

Bit 5~4 **PBS05~PBS04**: PB2 Pin-Shared Function Selection
00: PB2/STCK
01: STP
10: SCOM10/SSEG10
11: AN2

Bit 3~2 **PBS03~PBS02**: PB1 Pin-Shared Function Selection
00: PB1/INT1
01: SCOM9/SSEG9
10: AN1
11: XT2

Bit 1~0 **PBS01~PBS00**: PB0 Pin-Shared Function Selection
00: PB0/INT0
01: SCOM8/SSEG8
10: AN0
11: XT1

• **PBS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	PBS15	PBS14	PBS13	PBS12	PBS11	PBS10
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as "0"

Bit 5~4 **PBS15~PBS14**: PB6 Pin-Shared function selection
00: PB6
01: SCK/SCL
10: CMPINP
11: SCOM14/SSEG14

Bit 3~2 **PBS13~PBS12**: PB5 Pin-Shared function selection
00: PB5
01: $\overline{\text{SCS}}$
10: CMPINN
11: SCOM13/SSEG13

Bit 1~0 **PBS11~PBS10**: PB4 Pin-Shared function selection
00: PB4
01: CLO
10: SCOM12/SSEG12
11: AN8

• **PCS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PCS07	PCS06	PCS05	PCS04	PCS03	PCS02	PCS01	PCS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PCS07~PCS06**: Reserved, must be fixed at "00"

Bit 5~4 **PCS05~PCS04**: PC2 Pin-Shared Function Selection
00: PC2
01: PC2
10: SDO
11: SCOM17/SSEG17

Bit 3~2 **PCS05~PCS04**: PC1 Pin-Shared Function Selection
00: PC1
01: RX0/TX0
10: SCOM16/SSEG16
11: OSC2

Bit 1~0 **PCS01~PCS00**: PC0 Pin-Shared Function Selection
00: PC0
01: TX0
10: SCOM15/SSEG15
11: OSC1

• **PCS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	PCS15	PCS14	PCS13	PCS12	PCS11	PCS10
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as "0"

Bit 5~4 **PCS15~PCS14**: Reserved, must be fixed at "00"

Bit 3~2 **PCS13~PCS12**: PC5 Pin-Shared Function Selection
00: PC5
01: PC5
10: SCK/SCL
11: SCOM20/SSEG20

Bit 1~0 **PCS11~PCS10**: PC4 Pin-Shared Function Selection
00: PC4
01: PC4
10: SDI/SDA
11: SCOM19/SSEG19

• **PDS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PDS07	PDS06	PDS05	PDS04	PDS03	PDS02	PDS01	PDS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PDS07~PDS06**: Reserved, must be fixed at "00"

Bit 5~4 **PDS05~PDS04**: PD2 Pin-Shared Function Selection
00: PD2
01: TX0
10: SCOM24/SSEG24
11: AN10

Bit 3~2 **PDS03~PCS02**: PD1 Pin-Shared Function Selection
00: PD1
01: RX0/TX0
10: SCOM23/SSEG23
11: AN11

Bit 1~0 **PCS01~PCS00**: PD0 Pin-Shared Function Selection
00: PD0
01: PD0
10: PTP0
11: SCOM22/SSEG22

• **PDS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	PDS17	PDS16	PDS15	PDS14	PDS13	PDS12	PDS11	PDS10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PDS17~PDS16**: Reserved, must be fixed at "00"

Bit 5~4 **PDS15~PDS14**: Reserved, must be fixed at "00"

Bit 3~2 **PDS13~PCS12**: PD5 Pin-Shared Function Selection
00: PD5/INT2/PTCK1
01: PD5/INT2/PTCK1
10: RX1/TX1
11: SCOM27/SSEG27

Bit 1~0 **PCS11~PCS10**: Reserved, must be fixed at "00"

• **IFS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	RX1PS	INT2PS	INT1PS	INT0PS	SDI_SDAPS	SCK_SCLPS	SCSBPS	RX0PS
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **RX1PS**: RX1/TX1 input source pin selection
0: PD5
1: Reserved

Bit 6 **INT2PS**: INT2 input source pin selection
0: Reserved
1: PD5

Bit 5 **INT1PS**: INT1 input source pin selection
0: PB1
1: PA2

Bit 4 **INT0PS**: INT0 input source pin selection
0: PB0
1: PA1

Bit 3 **SDI_SDAPS**: SDI/SDA input source pin selection
0: PC4
1: PA3

Bit 2 **SCK_SCLPS**: SCK/SCL input source pin selection
0: PC5
1: PB6

Bit 1 **SCSBPS**: $\overline{\text{SCS}}$ input source pin selection
0: Reserved
1: PB5

Bit 0 **RX0PS**: RX0/TX0 input source pin selection
0: PD1
1: PC1

• **IFS1 Register**

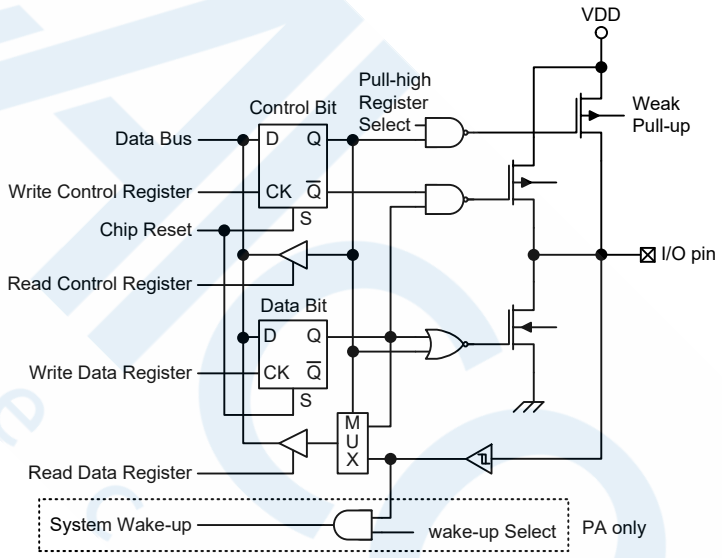
Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	PTCK1PS
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

Bit 7~1 Unimplemented, read as "0"

Bit 0 **PTCK1PS**: PTCK1 input source pin selection
0: PD5
1: Reserved

I/O Pin Structures

The accompanying diagram illustrates the internal structure of the I/O logic function. As the exact logical construction of the I/O pin will differ from this drawing, it is supplied as a guide only to assist with the functional understanding of the I/O logic function. The wide range of pin-shared structures does not permit all types to be shown.



Logic Function Input/Output Structure

Programming Considerations

Within the user program, one of the first things to consider is port initialisation. After a reset, all of the I/O data and port control registers will be set high. This means that all I/O pins will default to an input state, the level of which depends on the other connected circuitry and whether pull-high selections have been chosen. If the port control registers are then programmed to set some pins as outputs, these output pins will have an initial high output value unless the associated port data registers are first programmed. Selecting which pins are inputs and which are outputs can be achieved byte-wide by loading the correct values into the appropriate port control register or by programming individual bits in the port control register using the "SET [m].i" and "CLR [m].i" instructions. Note that when using these bit control instructions, a read-modify-write operation takes place. The microcontroller must first read in the data on the entire port, modify it to the required new bit values and then rewrite this data back to the output ports.

Port A has the additional capability of providing wake-up function. When the device is in the SLEEP or IDLE Mode, various methods are available to wake the device up. One of these is a high to low transition of any of the Port A pins. Single or multiple pins on Port A can be set to have this function.

Timer Modules – TM

One of the most fundamental functions in any microcontroller device is the ability to control and measure time. To implement time related functions the device includes several Timer Modules, abbreviated to the name TM. The TMs are multi-purpose timing units and serves to provide operations such as Timer/Counter, Compare Match Output and Single Pulse Output as well as being the functional unit for the generation of PWM signals. Each of the TMs has two individual interrupts. The addition of input and output pins for the TM ensures that users are provided with timing units with a wide and flexible range of features.

The common features of the different TM types are described here with more detailed information provided in the individual Compact, Standard and Periodic Type TM sections.

Introduction

The device contains several TMs and each individual TM can be categorised as a certain type, namely Compact Type TM, Standard Type TM or Periodic Type TM. Although similar in nature, the different TM types vary in their feature complexity. The common features to all of the Compact, Standard and Periodic TMs will be described in this section and the detailed operation regarding each of the TM types will be described in separate sections. The main features and differences between the three types of TMs are summarised in the accompanying table.

TM Function	CTM	STM	PTM
Timer/Counter	√	√	√
Compare Match Output	√	√	√
PWM Output	√	√	√
Single Pulse Output	—	√	√
PWM Alignment	Edge	Edge	Edge
PWM Adjustment Period & Duty	Duty or Period	Duty or Period	Duty or Period

TM Function Summary

TM Operation

The different types of TM offer a diverse range of functions, from simple timing operations to PWM signal generation. The key to understanding how the TM operates is to see it in terms of a free running counter whose value is then compared with the value of pre-programmed internal comparators. When the free running count-up counter has the same value as the pre-programmed comparator, known as a compare match situation, a TM interrupt signal will be generated which can clear the counter and perhaps also change the condition of the TM output pin. The internal TM counter is driven by a user selectable clock source, which can be an internal clock or an external pin.

TM Clock Source

The clock source which drives the main counter in each TM can originate from various sources. The selection of the required clock source is implemented using the xTnCK2~xTnCK0 bits in the xTMn control registers, where “x” stands for C, S or P type TM and “n” stands for the specific TM serial number. For the CTM and STM there are no serial number “n” in the revelant pins, registers and control bits since there is only one CTM and one STM in the device. The clock source can be a ratio of the system clock, f_{SYS}, or the internal high clock, f_H, the f_{SUB} clock source or the external xTCKn pin. The xTCKn pin clock source is used to allow an external signal to drive the TM as an external clock source for event counting.

TM Interrupts

The Compact Type, Standard Type and Periodic Type TMs each have two internal interrupts, one for each of the internal comparator A or comparator P, which generate a TM interrupt when a compare match condition occurs. When a TM interrupt is generated it can be used to clear the counter and also to change the state of the TM output pin.

TM External Pins

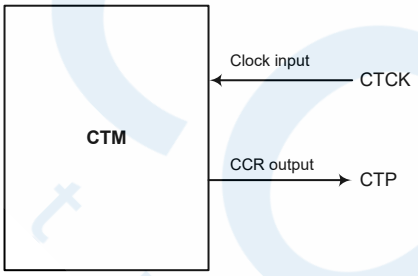
Each of the TMs, irrespective of what type, has one TM input pin, with the label xTCKn. The xTMn input pin, xTCKn, is essentially a clock source for the xTMn and is selected using the xTnCK2~xTnCK0 bits in the xTMnC0 register. This external TM input pin allows an external clock source to drive the internal TM. The xTCKn input pin can be chosen to have either a rising or falling active edge. The STCK and PTCKn pins are also used as the external trigger input pin in single pulse output mode for the STM and PTMn respectively.

The TMs each has one output pin, xTPn. When the TM is in the Compare Match Output Mode, the pin can be controlled by the xTMn to switch to a high or low level or to toggle when a compare match situation occurs. The output pin is also the pin where the TM generates the PWM output waveform.

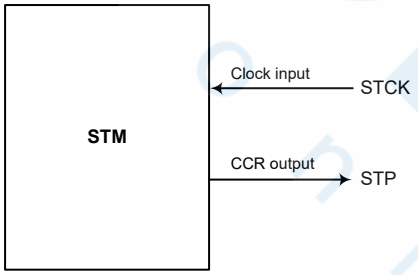
As the TM input and output pins are pin-shared with other functions, the TM input and output functions must first be setup using the relevant pin-shared function selection bits described in the Pin-shared Function section. The details of the pin-shared function selection are described in the pin-shared function section.

CTM		STM		PTM0		PTM1	
Input	Output	Input	Output	Input	Output	Input	Output
CTCK	CTP	STCK	STP	PTCK0	PTP0	PTCK1	—

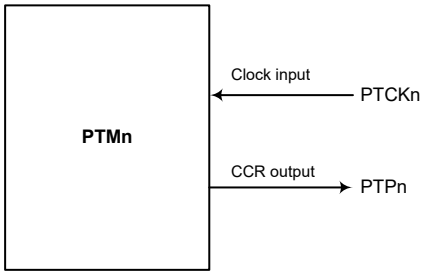
TM External Pins



CTM Function Pin Block Diagram



STM Function Pin Block Diagram

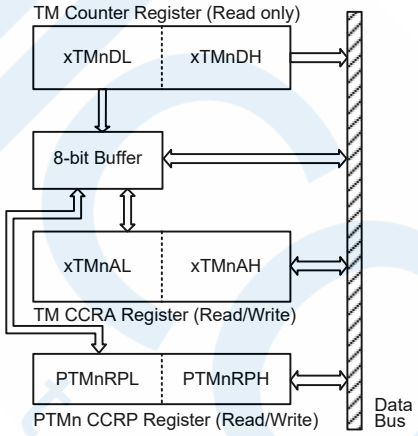


PTMn Function Pin Block Diagram (n=0~1)

Programming Considerations

The TM Counter Registers and the Compare CCRA and CCRP registers, all have a low and high byte structure. The high bytes can be directly accessed, but as the low bytes can only be accessed via an internal 8-bit buffer, reading or writing to these register pairs must be carried out in a specific way. The important point to note is that data transfer to and from the 8-bit buffer and its related low byte only takes place when a write or read operation to its corresponding high byte is executed.

As the CCRA and CCRP registers are implemented in the way shown in the following diagram and accessing these register pairs is carried out in a specific way as described above, it is recommended to use the "MOV" instruction to access the CCRA and CCRP low byte registers, named xTMnAL and PTMnRPL, using the following access procedures. Accessing the CCRA or CCRP low byte registers without following these access procedures will result in unpredictable values.

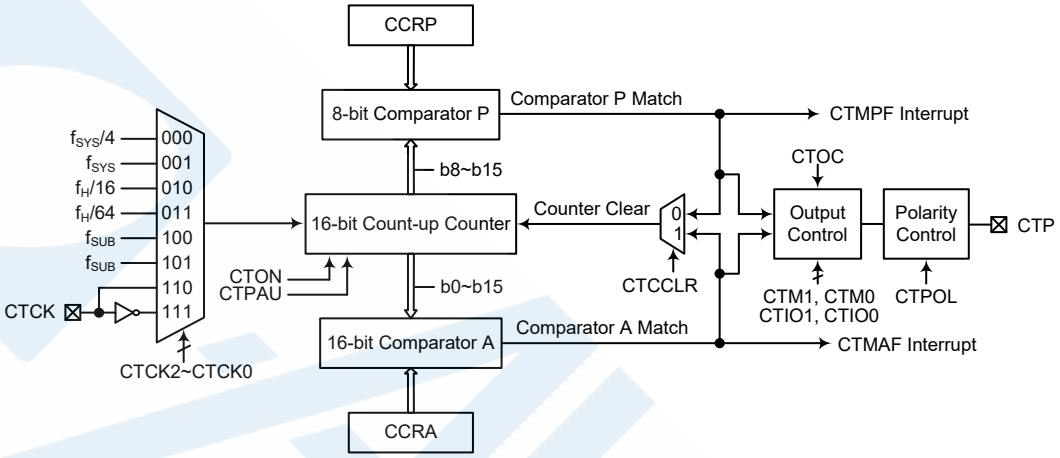


The following steps show the read and write procedures:

- Writing Data to CCRA or CCRP
  - Step 1. Write data to Low Byte xTMnAL or PTMnRPL
    - Note that here data is only written to the 8-bit buffer.
  - Step 2. Write data to High Byte xTMnAH or PTMnRPH
    - Here data is written directly to the high byte registers and simultaneously data is latched from the 8-bit buffer to the Low Byte registers.
- Reading Data from the Counter Registers and CCRA or CCRP
  - Step 1. Read data from the High Byte xTMnDH, xTMnAH or PTMnRPH
    - Here data is read directly from the High Byte registers and simultaneously data is latched from the Low Byte register into the 8-bit buffer.
  - Step 2. Read data from the Low Byte xTMnDL, xTMnAL or PTMnRPL
    - This step reads data from the 8-bit buffer.

Compact Type TM – CTM

Although the simplest form of the three TM types, the Compact TM type still contains three operating modes, which are Compare Match Output, Timer/Event Counter and PWM Output modes. The Compact TM can also be controlled with an external input pin and can drive one external output pin.



Note: The CTM external pins are pin-shared with other functions, so before using the CTM function, ensure that the relevant pin-shared function registers have been set properly to enable the CTM pin function. The CTCK pin, if used, must also be set as an input by setting the corresponding bit in the port control register.

16-bit Compact Type TM Block Diagram

Compact Type TM Operation

At its core is a 16-bit count-up counter which is driven by a user selectable internal or external clock source. There are also two internal comparators with the names, Comparator A and Comparator P. These comparators will compare the value in the counter with CCRP and CCRA registers. The CCRP is 8-bit wide whose value is compared with the highest eight bits in the counter while the CCRA is 16-bit wide and therefore compares with all counter bits.

The only way of changing the value of the 16-bit counter using the application program, is to clear the counter by changing the CTON bit from low to high. The counter will also be cleared automatically by a counter overflow or a compare match with one of its associated comparators. When these conditions occur, a CTM interrupt signal will also usually be generated. The Compact Type TM can operate in a number of different operational modes, can be driven by different clock sources including an input pin and can also control one output pin. All operating setup conditions are selected using relevant internal registers.

Compact Type TM Register Description

Overall operation of the Compact TM is controlled using a series of registers. A read only register pair exists to store the internal counter 16-bit value, while a read/write register pair exists to store the internal 16-bit CCRA value. The CTMRP register is used to store the 8-bit CCRP value. The remaining two registers are control registers which setup the different operating and control modes.

Register Name	Bit							
	7	6	5	4	3	2	1	0
CTMC0	CTPAU	CTCK2	CTCK1	CTCK0	CTON	—	—	—
CTMC1	CTM1	CTM0	CTIO1	CTIO0	CTOC	CTPOL	CTDPX	CTCCLR
CTMDL	D7	D6	D5	D4	D3	D2	D1	D0
CTMDH	D15	D14	D13	D12	D11	D10	D9	D8
CTMAL	D7	D6	D5	D4	D3	D2	D1	D0
CTMAH	D15	D14	D13	D12	D11	D10	D9	D8
CTMRP	D7	D6	D5	D4	D3	D2	D1	D0

16-bit Compact TM Register List

- **CTMC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	CTPAU	CTCK2	CTCK1	CTCK0	CTON	—	—	—
R/W	R/W	R/W	R/W	R/W	R/W	—	—	—
POR	0	0	0	0	0	—	—	—

Bit 7 **CTPAU**: CTM counter pause control
0: Run
1: Pause
The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the CTM will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

Bit 6~4 **CTCK2~CTCK0**: CTM counter clock selection
000: $f_{SYS}/4$
001: f_{SYS}
010: $f_H/16$
011: $f_H/64$
100: f_{SUB}
101: f_{SUB}
110: CTCK rising edge clock
111: CTCK falling edge clock
These three bits are used to select the clock source for the CTM. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source f_{SYS} is the system clock, while f_H and f_{SUB} are other internal clocks, the details of which can be found in the oscillator section.

Bit 3 **CTON**: CTM counter on/off control
0: Off
1: On
This bit controls the overall on/off function of the CTM. Setting the bit high enables the counter to run while clearing the bit disables the CTM. Clearing this bit to zero will stop the counter from counting and turn off the CTM which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again. If the CTM is in the Compare Match Output Mode or the PWM Output Mode then the CTM output pin will be reset to its initial condition, as specified by the CTOC bit, when the CTON bit changes from low to high.

Bit 2~0 Unimplemented, read as “0”

• **CTMC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	CTM1	CTM0	CTIO1	CTIO0	CTOC	CTPOL	CTDPX	CTCCLR
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **CTM1~CTM0**: CTM operating mode selection
00: Compare Match Output Mode
01: Undefined
10: PWM Output Mode
11: Timer/Counter Mode

These bits set the required operating mode for the CTM. To ensure reliable operation the CTM should be switched off before any changes are made to the CTM1 and CTM0 bits. In the Timer/Counter Mode, the CTM output pin state is undefined.

Bit 5~4 **CTIO1~CTIO0**: Select CTM external pin function

Compare Match Output Mode
00: No change
01: Output low
10: Output high
11: Toggle output

PWM Output Mode
00: PWM output inactive state
01: PWM output active state
10: PWM output
11: Undefined

Timer/Counter Mode
Unused

These two bits are used to determine how the CTM external pin changes state when a certain condition is reached. The function that these bits select depends upon in which mode the CTM is running.

In the Compare Match Output Mode, the CTIO1 and CTIO0 bits determine how the CTM output pin changes state when a compare match occurs from the Comparator A. The CTM output pin can be set to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the CTM output pin should be configured using the CTOC bit in the CTMC1 register. Note that the output level requested by the CTIO1 and CTIO0 bits must be different from the initial value setup using the CTOC bit otherwise no change will occur on the CTM output pin when a compare match occurs. After the CTM output pin changes state, it can be reset to its initial level by changing the level of the CTON bit from low to high.

In the PWM Output Mode, the CTIO1 and CTIO0 bits determine how the CTM output pin changes state when a certain compare match condition occurs. The PWM output function is modified by changing these two bits. It is necessary to change the values of the CTIO1 and CTIO0 bits only after the CTM has been switched off. Unpredictable PWM outputs will occur if the CTIO1 and CTIO0 bits are changed when the CTM is running.

Bit 3 **CTOC**: CTM CTP output control

Compare Match Output Mode
0: Initial low
1: Initial high

PWM Output Mode
0: Active low
1: Active high

This is the output control bit for the CTM output pin. Its operation depends upon whether CTM is being used in the Compare Match Output Mode or in the PWM

Output Mode. It has no effect if the CTM is in the Timer/Counter Mode. In the Compare Match Output Mode it determines the logic level of the CTM output pin before a compare match occurs. In the PWM Output Mode it determines if the PWM signal is active high or active low.

Bit 2 **CTPOL**: CTM CTP output polarity control
0: Non-invert
1: Invert
This bit controls the polarity of the CTP output pin. When the bit is set high the CTM output pin will be inverted and not inverted when the bit is zero. It has no effect if the CTM is in the Timer/Counter Mode.

Bit 1 **CTDPX**: CTM PWM duty/period control
0: CCRP – period; CCRA – duty
1: CCRP – duty; CCRA – period
This bit determines which of the CCRA and CCRP registers are used for period and duty control of the PWM waveform.

Bit 0 **CTCCLR**: CTM counter clear condition selection
0: CTM Comparator P match
1: CTM Comparator A match
This bit is used to select the method which clears the counter. Remember that the Compact TM contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the CTCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The CTCCLR bit is not used in the PWM Output Mode.

• CTMDL Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: CTM Counter Low Byte Register bit 7 ~ bit 0
CTM 16-bit Counter bit 7 ~ bit 0

• CTMDH Register

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0

Bit 7~2 **D15~D8**: CTM Counter High Byte Register bit 7 ~ bit 0
CTM 16-bit Counter bit 15 ~ bit 8

• CTMAL Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: CTM CCRA Low Byte Register bit 7 ~ bit 0
CTM 16-bit CCRA bit 7 ~ bit 0

• **CTMAH Register**

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D15~D8**: CTM CCRA High Byte Register bit 7 ~ bit 0
CTM 16-bit CCRA bit 15 ~ bit 8

• **CTMRP Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: CTM CCRP 8-bit register, compared with the CTM Counter bit 15 ~ bit 8
Comparator P Match Period=
0: 65536 CTM clocks
1~255: 256×(1~255) CTM clocks
These eight bits are used to setup the value on the internal CCRP 8-bit register, which are then compared with the internal counter's highest eight bits. The result of this comparison can be selected to clear the internal counter if the CTCCLR bit is set to zero. Setting the CTCCLR bit to zero ensures that a compare match with the CCRP values will reset the internal counter. As the CCRP bits are only compared with the highest eight counter bits, the compare values exist in 256 clock cycle multiples. Clearing all eight bits to zero is in effect allowing the counter to overflow at its maximum value.

Compact Type TM Operating Modes

The Compact Type TM can operate in one of three operating modes, Compare Match Output Mode, PWM Output Mode or Timer/Counter Mode. The operating mode is selected using the CTM1 and CTM0 bits in the CTMC1 register.

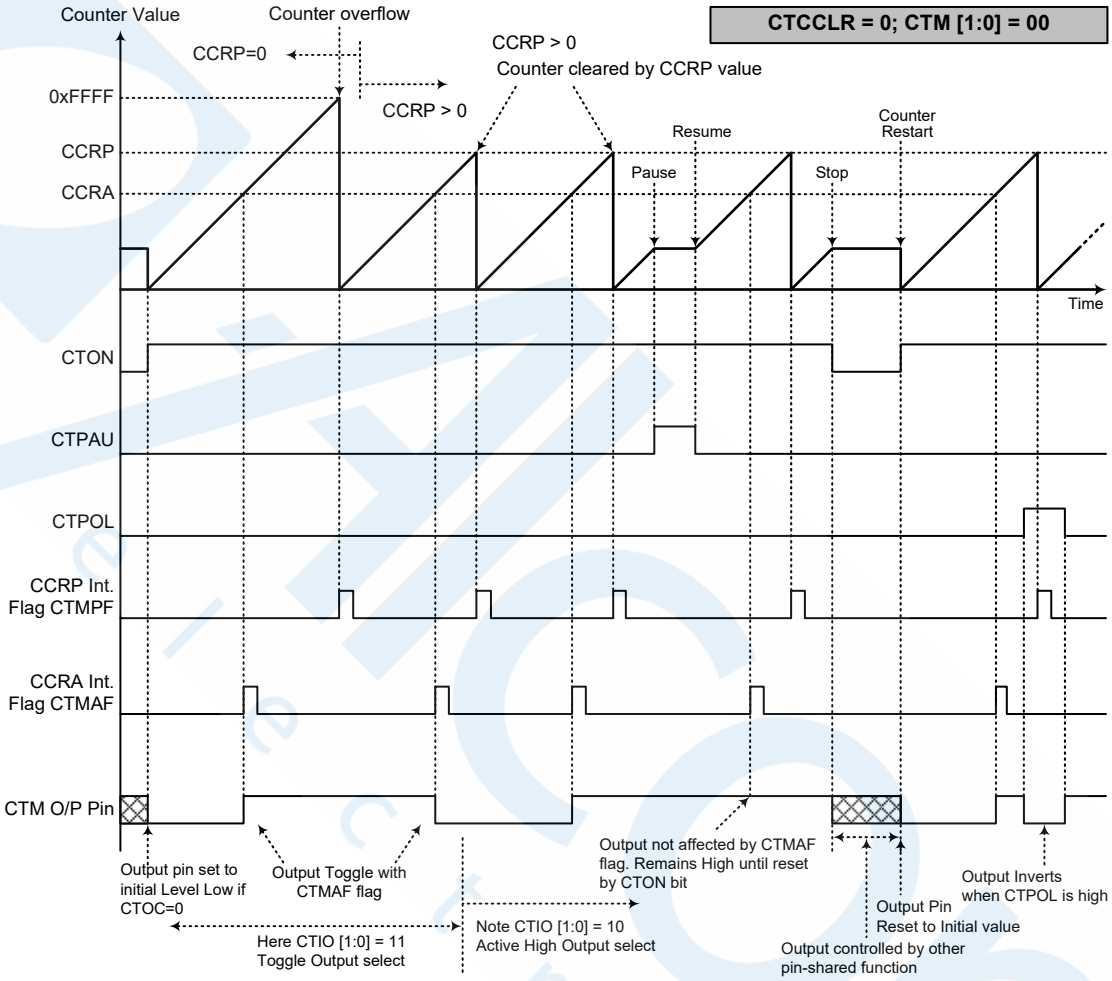
Compare Match Output Mode

To select this mode, bits CTM1 and CTM0 in the CTMC1 register, should be set to "00" respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the CTCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match occurs from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both CTMAF and CTMPF interrupt request flags for the Comparator A and Comparator P respectively, will both be generated.

If the CTCCLR bit in the CTMC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the CTMAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when CTCCLR is high no CTMPF interrupt request flag will be generated. If the CCRA bits are all zero, the counter will overflow when it reaches its maximum 16-bit, FFFF Hex, value, however here the CTMAF interrupt request flag will not be generated.

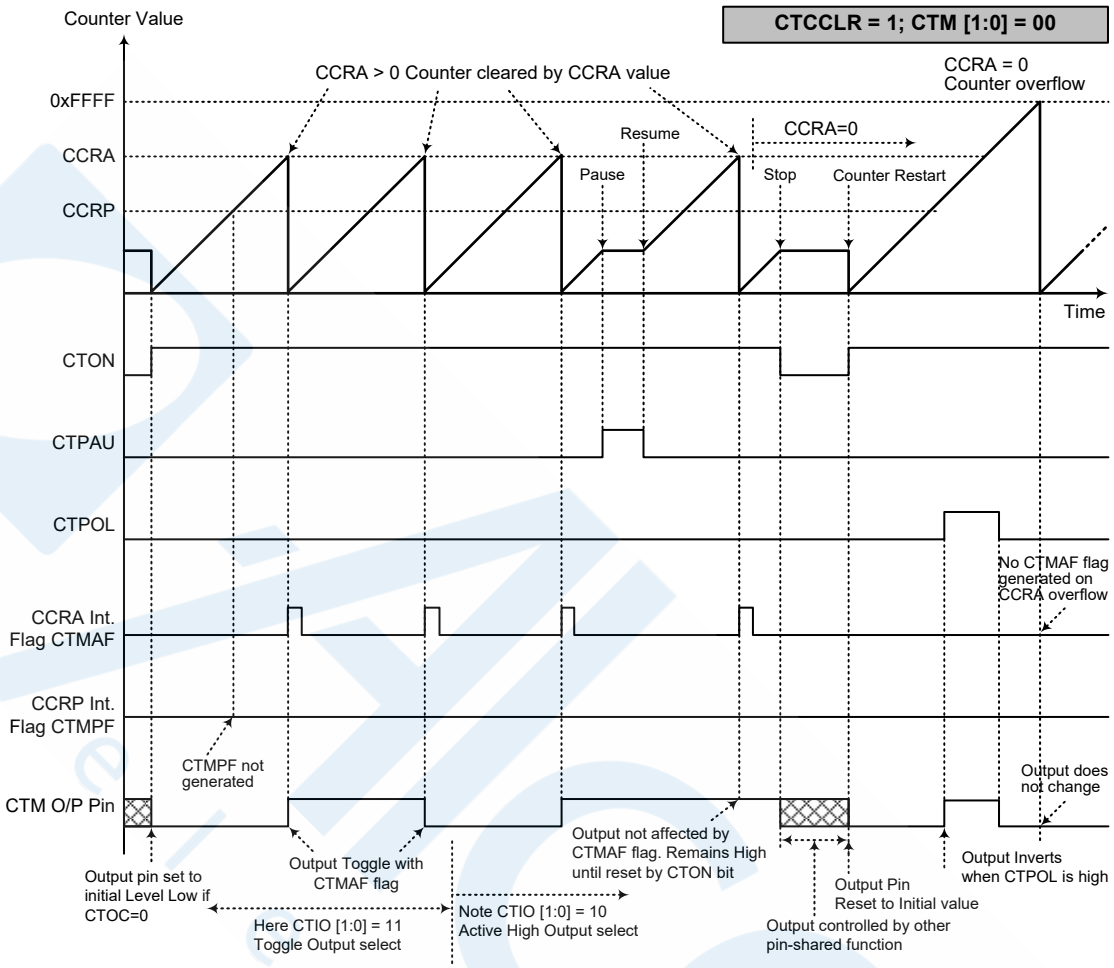
As the name of the mode suggests, after a comparison is made, the CTM output pin will change state. The CTM output pin condition however only changes state when a CTMAF interrupt request flag is generated after a compare match occurs from Comparator A. The CTMPF interrupt request flag, generated from a compare match occurs from Comparator P, will have no effect on the CTM

output pin. The way in which the CTM output pin changes state are determined by the condition of the CTIO1 and CTIO0 bits in the CTMC1 register. The CTM output pin can be selected using the CTIO1 and CTIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the CTM output pin, which is setup after the CTON bit changes from low to high, is setup using the CTOC bit. Note that if the CTIO1 and CTIO0 bits are zero then no pin change will take place.



Compare Match Output Mode – CTCCLR=0

Note: 1. With CTCCLR=0, a Comparator P match will clear the counter
2. The CTM output pin is controlled only by the CTMAF flag
3. The output pin is reset to its initial state by a CTON bit rising edge



Compare Match Output Mode – CTCCLR=1

Note: 1. With CTCCLR=1, a Comparator A match will clear the counter

2. The CTM output pin is controlled only by the CTMAF flag
3. The output pin is reset to its initial state by a CTON bit rising edge
4. The CTMPF flag is not generated when CTCCLR=1

Timer/Counter Mode

To select this mode, bits CTM1 and CTM0 in the CTMC1 register should be set to "11" respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the CTM output pin is not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the CTM output pin is not used in this mode, the pin can be used as a normal I/O pin or other pin-shared functions.

PWM Output Mode

To select this mode, bits CTM1 and CTM0 in the CTMC1 register should be set to "10" respectively. The PWM function within the CTM is useful for applications which require functions such as motor control, heating control, illumination control etc. By providing a signal of fixed frequency but of varying duty cycle on the CTM output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output Mode, the CTCCLR bit has no effect on the PWM operation. Both of the CCRA and CCRP registers are used to generate the PWM waveform, one register is used to clear the internal counter and thus control the PWM waveform frequency, while the other one is used to control the duty cycle. Which register is used to control either frequency or duty cycle is determined using the CTDPX bit in the CTMC1 register. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRA and CCRP registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The CTOC bit in the CTMC1 register is used to select the required polarity of the PWM waveform while the two CTIO1 and CTIO0 bits are used to enable the PWM output or to force the CTM output pin to a fixed high or low level. The CTPOL bit is used to reverse the polarity of the PWM output waveform.

• 16-bit CTM, PWM Output Mode, Edge-aligned Mode, CTDPX=0

CCRP	1~255	0
Period	CCRP×256	65536
Duty	CCRA	

If f_{SYS}=16MHz, CTM clock source is f_{SYS}/4, CCRP=2, CCRA=128,

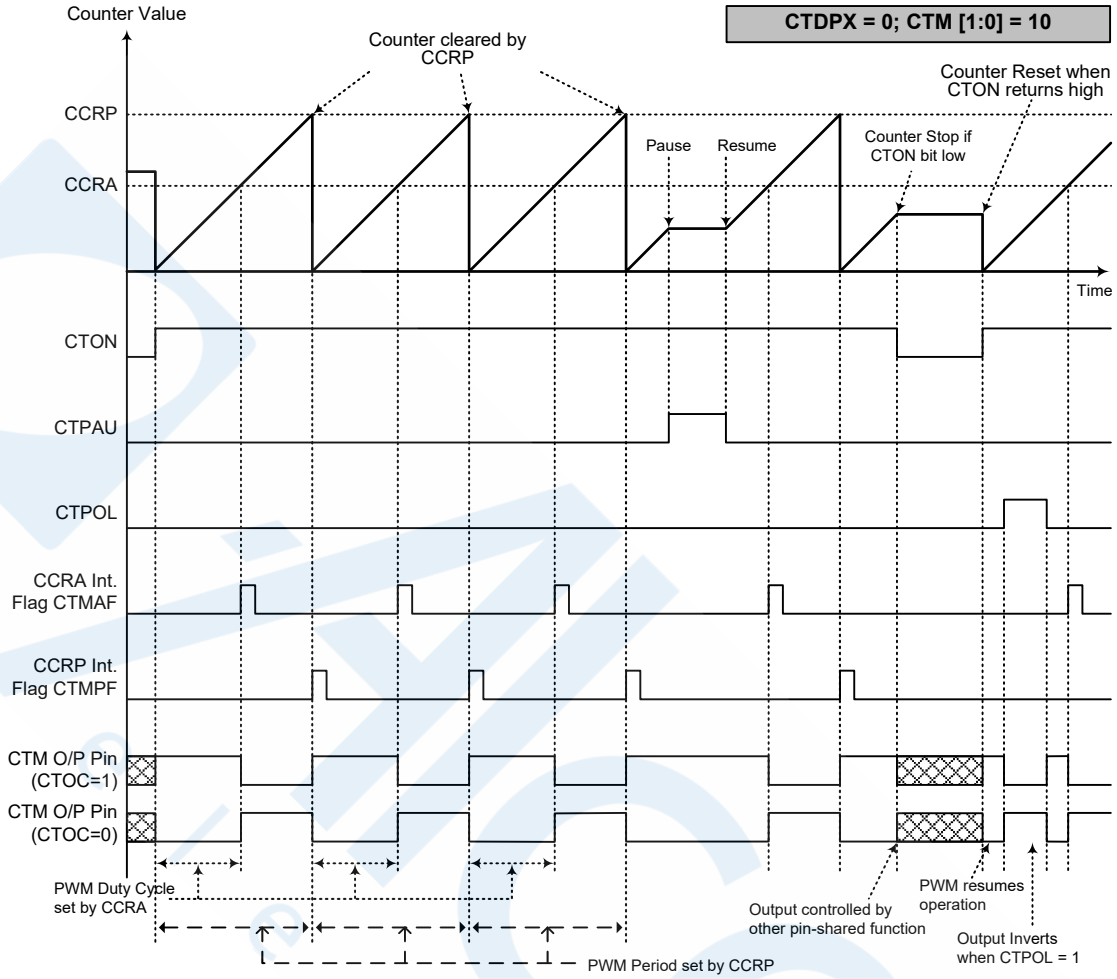
The CTM PWM output frequency=(f_{SYS}/4)/512=f_{SYS}/2048=7.8125kHz, duty=128/512=25%,

If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.

• 16-bit CTM, PWM Output Mode, Edge-aligned Mode, CTDPX=1

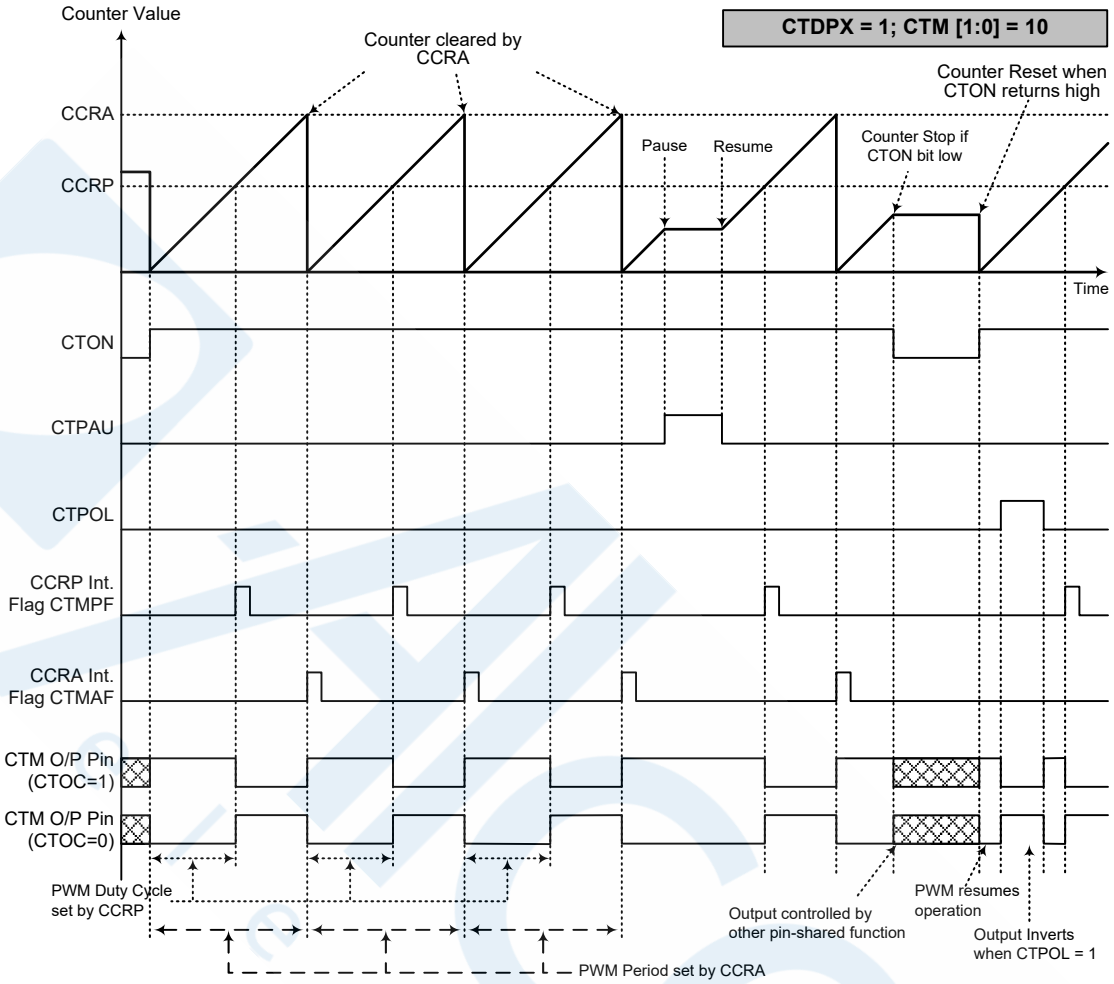
CCRP	1~255	0
Period	CCRA	
Duty	CCRP×256	65536

The PWM output period is determined by the CCRA register value together with the CTM clock while the PWM duty cycle is defined by the CCRP register value.



PWM Output Mode – CTDPX=0

Note: 1. Here CTDPX=0 – Counter cleared by CCRP
2. A counter clear sets the PWM Period
3. The internal PWM function continues even when CTIO[1:0]=00 or 01
4. The CTCCLR bit has no influence on PWM operation

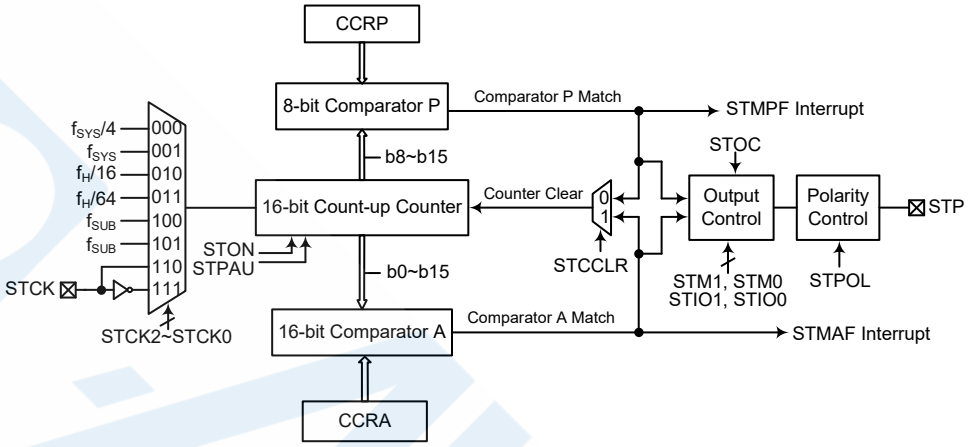


PWM Output Mode – CTDPX=1

Note: 1. Here CTDPX=1 – Counter cleared by CCRA
2. A counter clear sets the PWM Period
3. The internal PWM function continues even when CTIO[1:0]=00 or 01
4. The CTCCLR bit has no influence on PWM operation

Standard Type TM – STM

The Standard Type TM contains four operating modes, which are Compare Match Output, Timer/ Event Counter, Single Pulse Output and PWM Output modes. The Standard TM can also be controlled with one external input pin and can drive one external output pin.



Note: The STM external pins are pin-shared with other functions, so before using the STM function, ensure that the relevant pin-shared function registers have been set properly to enable the STM pin function. The STCK pin, if used, must also be set as an input by setting the corresponding bits in the port control register.

16-bit Standard Type TM Block Diagram

Standard Type TM Operation

The size of Standard TM is 16-bit wide and its core is a 16-bit count-up counter which is driven by a user selectable internal or external clock source. There are also two internal comparators with the names, Comparator A and Comparator P. These comparators will compare the value in the counter with CCRP and CCRA registers. The CCRP comparator is 8-bit wide whose value is compared with the highest 8 bits in the counter while the CCRA is the sixteen bits and therefore compares all counter bits.

The only way of changing the value of the 16-bit counter using the application program, is to clear the counter by changing the STON bit from low to high. The counter will also be cleared automatically by a counter overflow or a compare match with one of its associated comparators. When these conditions occur, an STM interrupt signal will also usually be generated. The Standard Type TM can operate in a number of different operational modes, can be driven by different clock sources including an input pin and can also control the output pin. All operating setup conditions are selected using relevant internal registers.

Standard Type TM Register Description

Overall operation of the Standard TM is controlled using a series of registers. A read only register pair exists to store the internal counter 16-bit value, while a read/write register pair exists to store the internal 16-bit CCRA value. The STMRP register is used to store the 8-bit CCRP value. The remaining two registers are control registers which set the different operating and control modes.

Register Name	Bit							
	7	6	5	4	3	2	1	0
STMC0	STPAU	STCK2	STCK1	STCK0	STON	—	—	—
STMC1	STM1	STM0	STIO1	STIO0	STOC	STPOL	STDPX	STCCLR
STMDL	D7	D6	D5	D4	D3	D2	D1	D0
STMDH	D15	D14	D13	D12	D11	D10	D9	D8
STMAL	D7	D6	D5	D4	D3	D2	D1	D0
STMAH	D15	D14	D13	D12	D11	D10	D9	D8
STMRP	D7	D6	D5	D4	D3	D2	D1	D0

16-bit Standard TM Register List

• **STMC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	STPAU	STCK2	STCK1	STCK0	STON	—	—	—
R/W	R/W	R/W	R/W	R/W	R/W	—	—	—
POR	0	0	0	0	0	—	—	—

Bit 7 **STPAU**: STM counter pause control
0: Run
1: Pause
The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the STM will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

Bit 6~4 **STCK2~STCK0**: STM counter clock selection
000: $f_{SYS}/4$
001: f_{SYS}
010: $f_H/16$
011: $f_H/64$
100: f_{SUB}
101: f_{SUB}
110: STCK rising edge clock
111: STCK falling edge clock
These three bits are used to select the clock source for the STM. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source f_{SYS} is the system clock, while f_H and f_{SUB} are other internal clocks, the details of which can be found in the oscillator section.

Bit 3 **STON**: STM counter on/off control
0: Off
1: On
This bit controls the overall on/off function of the STM. Setting the bit high enables the counter to run while clearing the bit disables the STM. Clearing this bit to zero will stop the counter from counting and turn off the STM which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again.
If the STM is in the Compare Match Output Mode, PWM Output Mode or Single Pulse Output Mode, then the STM output pin will be reset to its initial condition, as specified by the STOC bit, when the STON bit changes from low to high.

Bit 2~0 Unimplemented, read as “0”

• **STMC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	STM1	STM0	STIO1	STIO0	STOC	STPOL	STDPX	STCCLR
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **STM1~STM0**: STM operating mode selection
00: Compare Match Output Mode
01: Undefined
10: PWM Output Mode or Single Pulse Output Mode
11: Timer/Counter Mode

These bits setup the required operating mode for the STM. To ensure reliable operation the STM should be switched off before any changes are made to the STM1 and STM0 bits. In the Timer/Counter Mode, the STM output pin state is undefined.

Bit 5~4 **STIO1~STIO0**: STM external pin function selection

Compare Match Output Mode
00: No change
01: Output low
10: Output high
11: Toggle output

PWM Output Mode/Single Pulse Output Mode
00: PWM output inactive state
01: PWM output active state
10: PWM output
11: Single Pulse Output

Timer/Counter Mode
Unused

These two bits are used to determine how the STM output pin changes state when a certain condition is reached. The function that these bits select depends upon in which mode the STM is running.

In the Compare Match Output Mode, the STIO1 and STIO0 bits determine how the STM output pin changes state when a compare match occurs from the Comparator A. The TM output pin can be setup to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the STM output pin should be setup using the STOC bit in the STMC1 register. Note that the output level requested by the STIO1 and STIO0 bits must be different from the initial value setup using the STOC bit otherwise no change will occur on the STM output pin when a compare match occurs. After the STM output pin changes state, it can be reset to its initial level by changing the level of the STON bit from low to high.

In the PWM Output Mode, the STIO1 and STIO0 bits determine how the STM output pin changes state when a certain compare match condition occurs. The PWM output function is modified by changing these two bits. It is necessary to change the values of the STIO1 and STIO0 bits only after the STM has been switched off. Unpredictable PWM outputs will occur if the STIO1 and STIO0 bits are changed when the STM is running.

Bit 3 **STOC**: STM STP output control

Compare Match Output Mode
0: Initial low
1: Initial high

PWM Output Mode/Single Pulse Output Mode
0: Active low
1: Active high

This is the output control bit for the STM output pin. Its operation depends upon whether STM is being used in the Compare Match Output Mode or in the PWM Output Mode/Single Pulse Output Mode. It has no effect if the STM is in the Timer/

Counter Mode. In the Compare Match Output Mode it determines the logic level of the STM output pin before a compare match occurs. In the PWM Output Mode it determines if the PWM signal is active high or active low. In the Single Pulse Output Mode it determines the logic level of the STM output pin when the STON bit changes from low to high.

Bit 2 **STPOL**: STM STP output polarity control
0: Non-invert
1: Invert
This bit controls the polarity of the STP output pin. When the bit is set high the STM output pin will be inverted and not inverted when the bit is zero. It has no effect if the STM is in the Timer/Counter Mode.

Bit 1 **STDPX**: STM PWM duty/period control
0: CCRP – period; CCRA – duty
1: CCRP – duty; CCRA – period
This bit determines which of the CCRA and CCRP registers are used for period and duty control of the PWM waveform.

Bit 0 **STCCLR**: STM counter clear condition selection
0: Comparator P match
1: Comparator A match
This bit is used to select the method which clears the counter. Remember that the Standard TM contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the STCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The STCCLR bit is not used in the PWM Output Mode or Single Pulse Output Mode.

• STMDL Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: STM Counter Low Byte Register bit 7 ~ bit 0
STM 16-bit Counter bit 7 ~ bit 0

• STMDH Register

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D15~D8**: STM Counter High Byte Register bit 7 ~ bit 0
STM 16-bit Counter bit 15 ~ bit 8

• STMAL Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: STM CCRA Low Byte Register bit 7 ~ bit 0
STM 16-bit CCRA bit 7 ~ bit 0

• **STMAH Register**

Bit	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D15~D8**: STM CCRA High Byte Register bit 7 ~ bit 0
STM 16-bit CCRA bit 15 ~ bit 8

• **STMRP Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: STM CCRP 8-bit register, compared with the STM counter bit 15 ~ bit 8
Comparator P match period=
0: 65536 STM clocks
1~255: (1~255)×256 STM clocks
These eight bits are used to setup the value on the internal CCRP 8-bit register, which are then compared with the internal counter's highest eight bits. The result of this comparison can be selected to clear the internal counter if the STCCLR bit is set to zero. Setting the STCCLR bit to zero ensures that a compare match with the CCRP values will reset the internal counter. As the CCRP bits are only compared with the highest eight counter bits, the compare values exist in 256 clock cycle multiples. Clearing all eight bits to zero is in effect allowing the counter to overflow at its maximum value.

Standard Type TM Operation Modes

The Standard Type TM can operate in one of four operating modes, Compare Match Output Mode, PWM Output Mode, Single Pulse Output Mode or Timer/Counter Mode. The operating mode is selected using the STM1 and STM0 bits in the STMC1 register.

Compare Match Output Mode

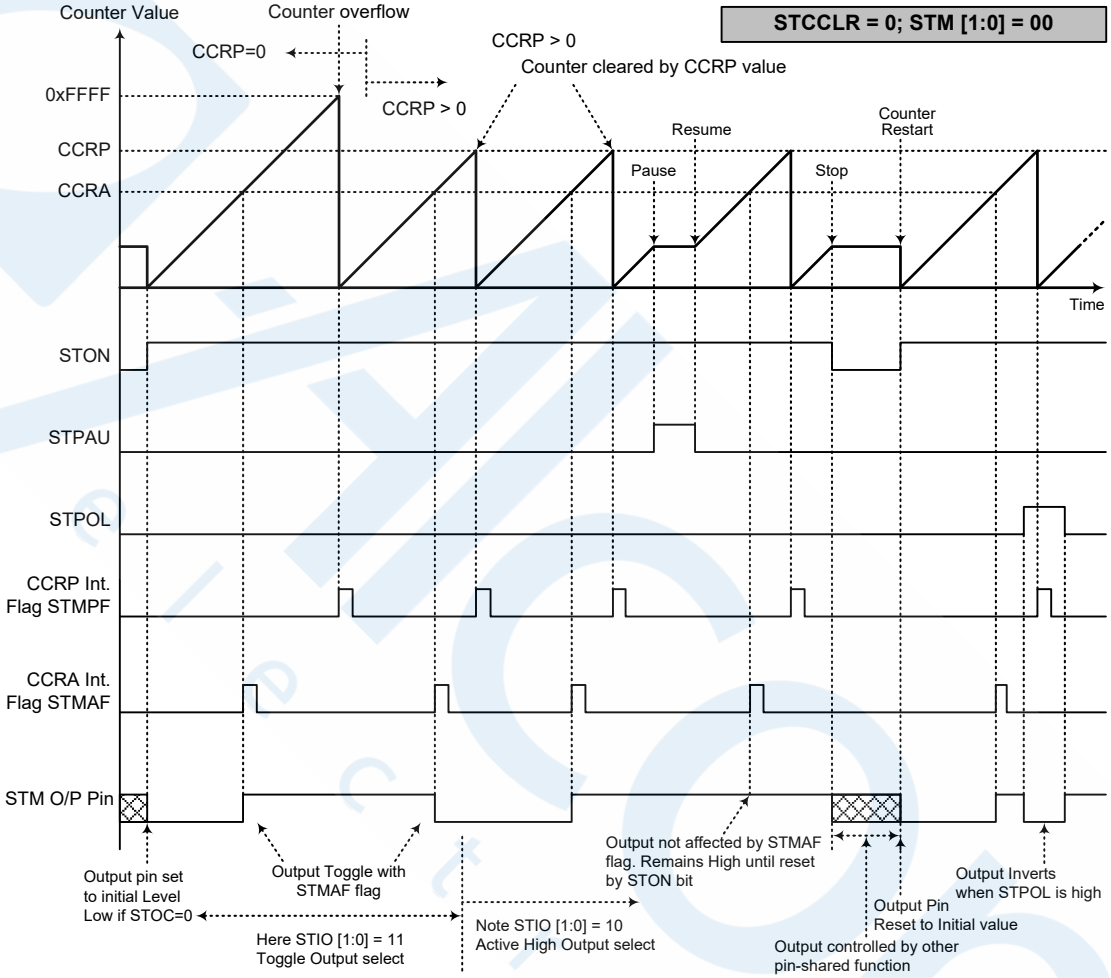
To select this mode, bits STM1 and STM0 in the STMC1 register, should be set to "00" respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the STCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both STMAF and STMPF interrupt request flags for Comparator A and Comparator P respectively, will both be generated.

If the STCCLR bit in the STMC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the STMAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when STCCLR is high no STMPF interrupt request flag will be generated. In the Compare Match Output Mode, the CCRA can not be set to "0".

If the CCRA bits are all zero, the counter will overflow when it reaches its maximum 16-bit, FFFF Hex, value, however here the STMAF interrupt request flag will not be generated.

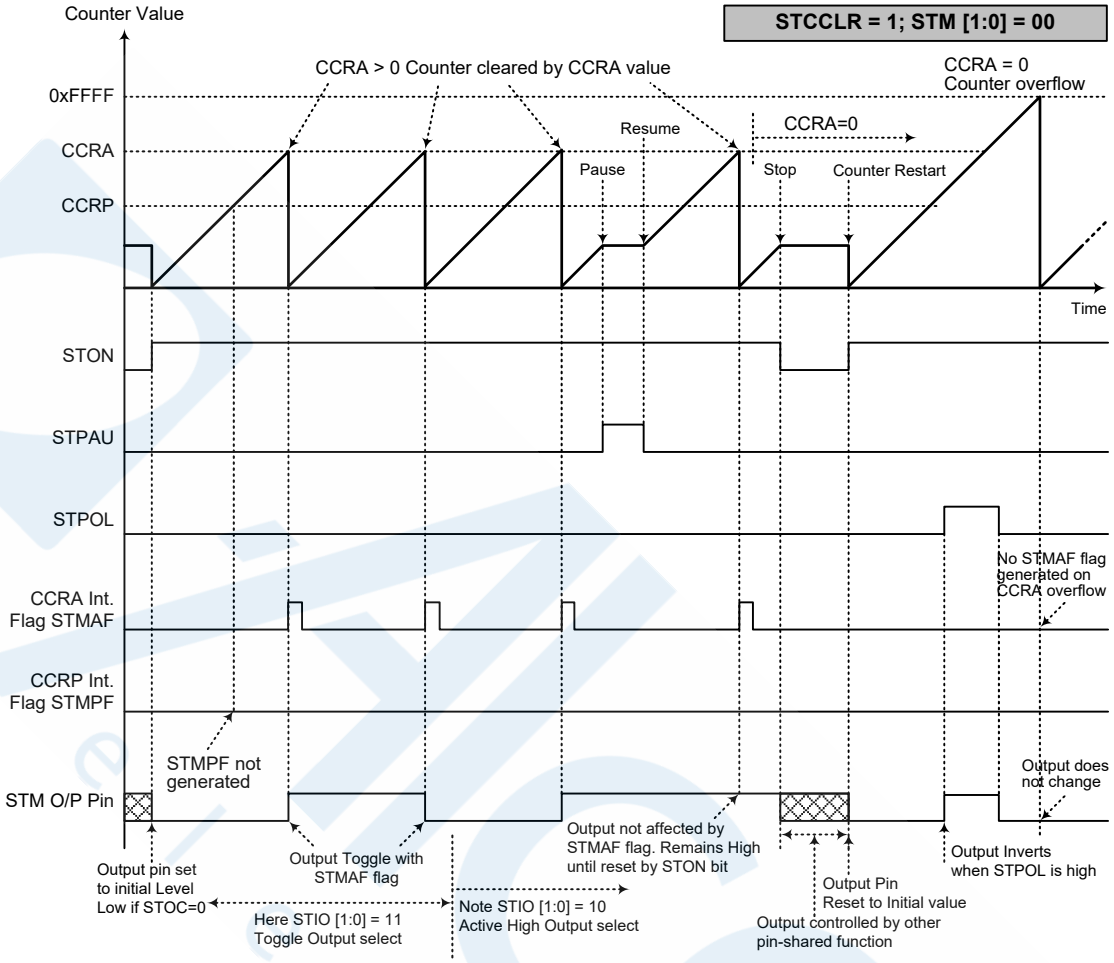
As the name of the mode suggests, after a comparison is made, the STM output pin, will change state. The STM output pin condition however only changes state when an STMAF interrupt request flag is generated after a compare match occurs from Comparator A. The STMPF interrupt request

flag, generated from a compare match occurs from Comparator P, will have no effect on the STM output pin. The way in which the STM output pin changes state are determined by the condition of the STIO1 and STIO0 bits in the STMC1 register. The STM output pin can be selected using the STIO1 and STIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the STM output pin, which is setup after the STON bit changes from low to high, is setup using the STOC bit. Note that if the STIO1 and STIO0 bits are zero then no pin change will take place.



Compare Match Output Mode – STCCLR=0

Note: 1. With STCCLR=0, a Comparator P match will clear the counter
2. The STM output pin is controlled only by the STMAF flag
3. The output pin is reset to its initial state by an STON bit rising edge



Compare Match Output Mode – STCCLR=1

Note: 1. With STCCLR=1, a Comparator A match will clear the counter
2. The STM output pin is controlled only by the STMAF flag
3. The output pin is reset to its initial state by an STON bit rising edge
4. The STMPF flag is not generated when STCCLR=1

Timer/Counter Mode

To select this mode, bits STM1 and STM0 in the STMC1 register should be set to "11" respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the STM output pin is not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the STM output pin is not used in this mode, the pin can be used as a normal I/O pin or other pin-shared functions.

PWM Output Mode

To select this mode, bits STM1 and STM0 in the STMC1 register should be set to "10" respectively and also the STIO1 and STIO0 bits should be set to "10" respectively. The PWM function within the STM is useful for applications which require functions such as motor control, heating control, illumination control etc. By providing a signal of fixed frequency but of varying duty cycle on the STM output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output Mode, the STCCLR bit has no effect on the PWM period. Both of the CCRA and CCRP registers are used to generate the PWM waveform, one register is used to clear the internal counter and thus control the PWM waveform frequency, while the other one is used to control the duty cycle. Which register is used to control either frequency or duty cycle is determined using the STDPX bit in the STMC1 register. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRA and CCRP registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The STOC bit in the STMC1 register is used to select the required polarity of the PWM waveform while the two STIO1 and STIO0 bits are used to enable the PWM output or to force the STM output pin to a fixed high or low level. The STPOL bit is used to reverse the polarity of the PWM output waveform.

- **16-bit STM, PWM Output Mode, Edge-aligned Mode, STDPX=0**

CCRP	1~255	0
Period	CCRP×256	65536
Duty	CCRA	

If f_{SYS}=16MHz, STM clock source is f_{SYS}/4, CCRP=2 and CCRA=128,

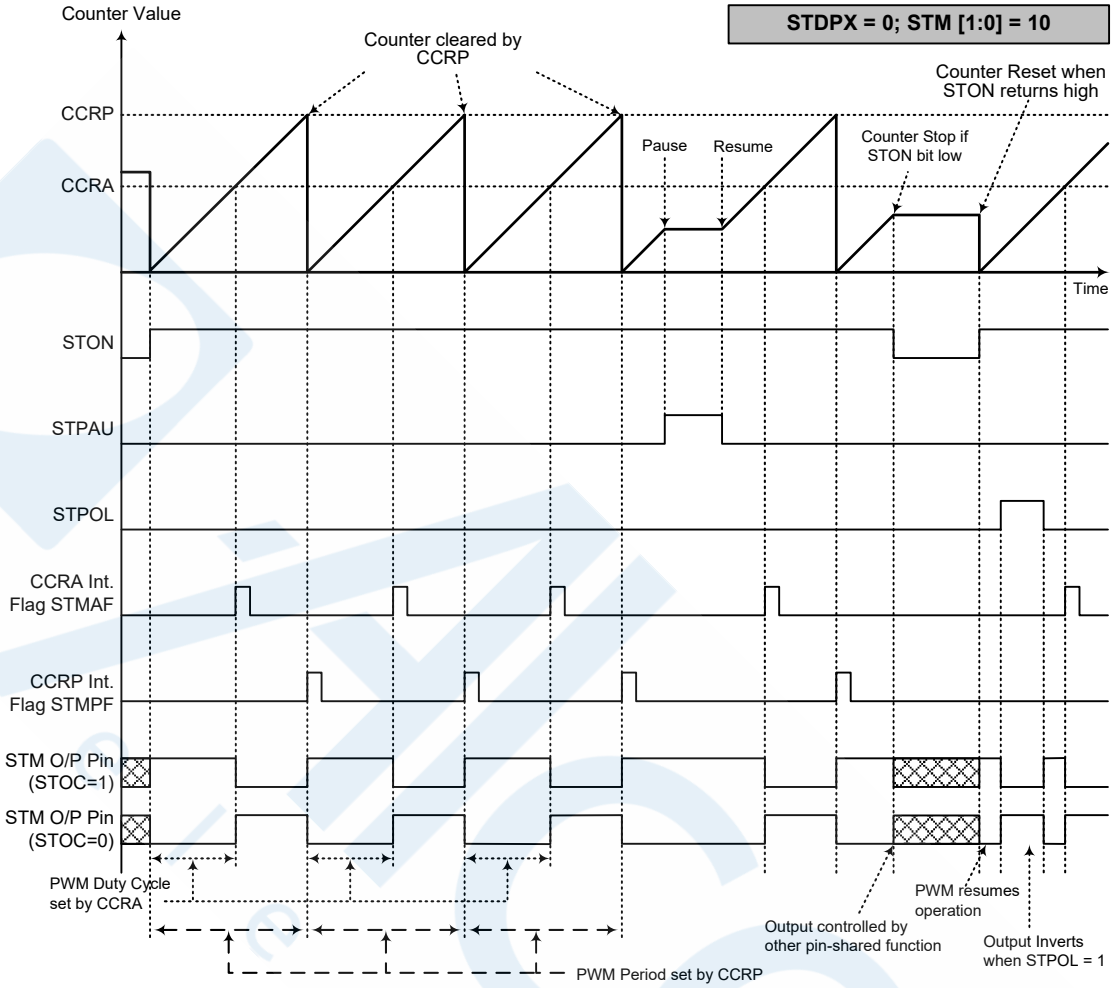
The STM PWM output frequency=(f_{SYS}/4)/(2×256)=f_{SYS}/2048=7.8125kHz, duty=128/(2×256)=25%,

If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.

- **16-bit STM, PWM Output Mode, Edge-aligned Mode, STDPX=1**

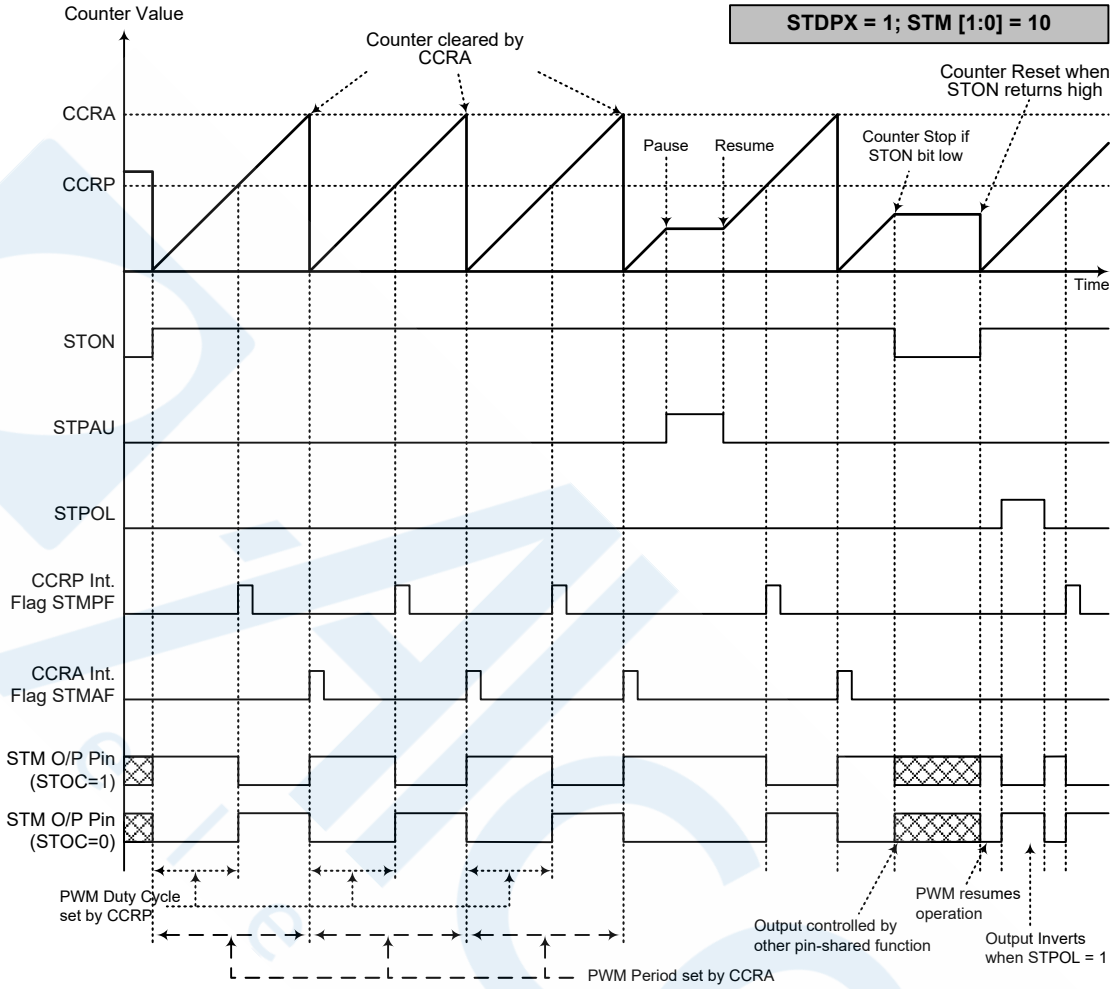
CCRP	1~255	0
Period	CCRA	
Duty	CCRP×256	65536

The PWM output period is determined by the CCRA register value together with the STM clock while the PWM duty cycle is defined by the CCRP register value.



PWM Output Mode – STDPX=0

Note: 1. Here STDPX=0 – Counter cleared by CCRP
2. A counter clear sets the PWM Period
3. The internal PWM function continues even when STIO[1:0]=00 or 01
4. The STCCLR bit has no influence on PWM operation



PWM Output Mode – STDPX=1

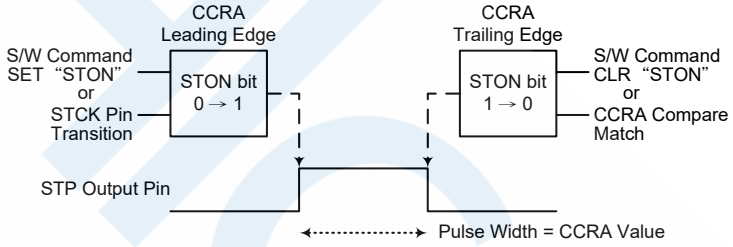
Note: 1. Here STDPX=1 – Counter cleared by CCRA
2. A counter clear sets the PWM Period
3. The internal PWM function continues even when STIO[1:0]=00 or 01
4. The STCCLR bit has no influence on PWM operation

Single Pulse Output Mode

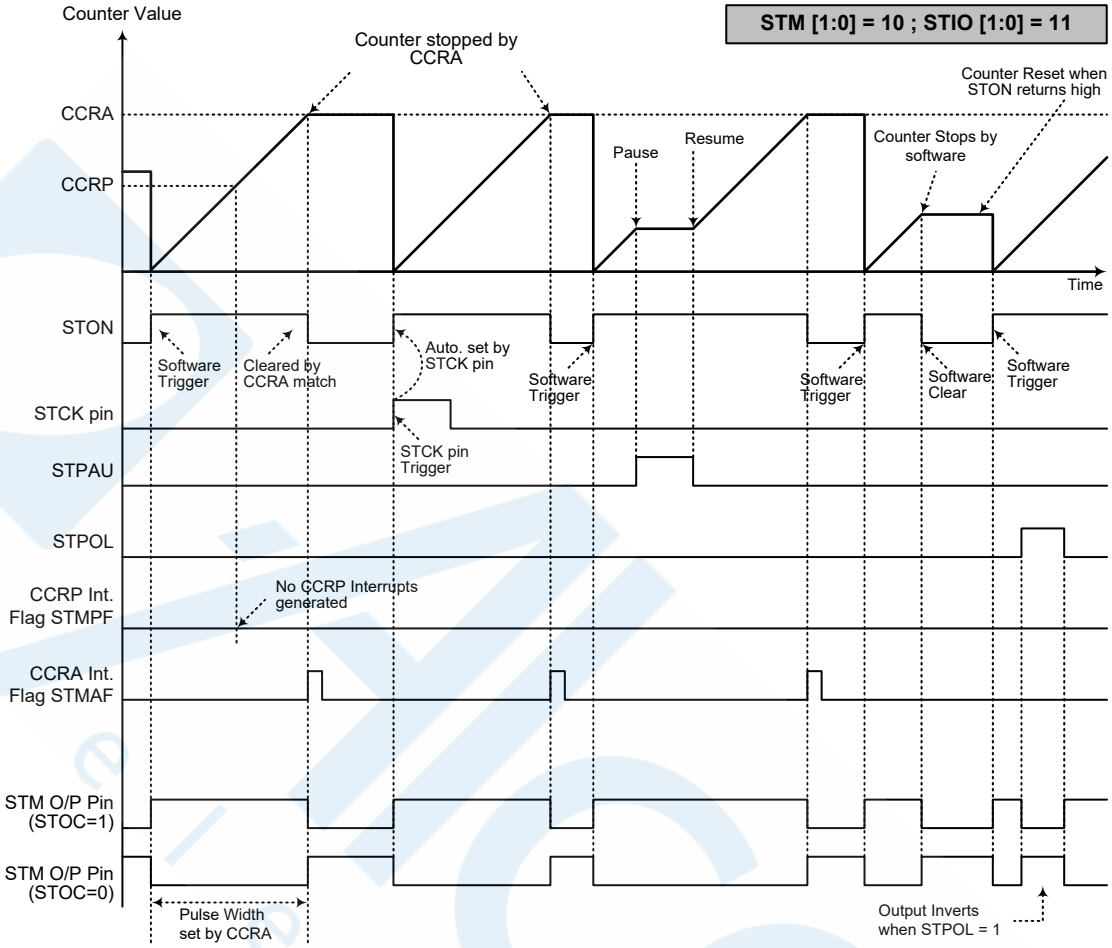
To select this mode, bits STM1 and STM0 in the STMC1 register should be set to "10" respectively and also the STIO1 and STIO0 bits should be set to "11" respectively. The Single Pulse Output Mode, as the name suggests, will generate a single shot pulse on the STM output pin.

The trigger for the pulse output leading edge is a low to high transition of the STON bit, which can be implemented using the application program. However in the Single Pulse Output Mode, the STON bit can also be made to automatically change from low to high using the external STCK pin, which will in turn initiate the Single Pulse output. When the STON bit transitions to a high level, the counter will start running and the pulse leading edge will be generated. The STON bit should remain high when the pulse is in its active state. The generated pulse trailing edge will be generated when the STON bit is cleared to zero, which can be implemented using the application program or when a compare match occurs from Comparator A.

However a compare match from Comparator A will also automatically clear the STON bit and thus generate the Single Pulse output trailing edge. In this way the CCRA value can be used to control the pulse width. A compare match from Comparator A will also generate an STM interrupt. The counter can only be reset back to zero when the STON bit changes from low to high when the counter restarts. In the Single Pulse Output Mode CCRP is not used. The STCCLR and STDPX bits are not used in this Mode.



Single Pulse Generation

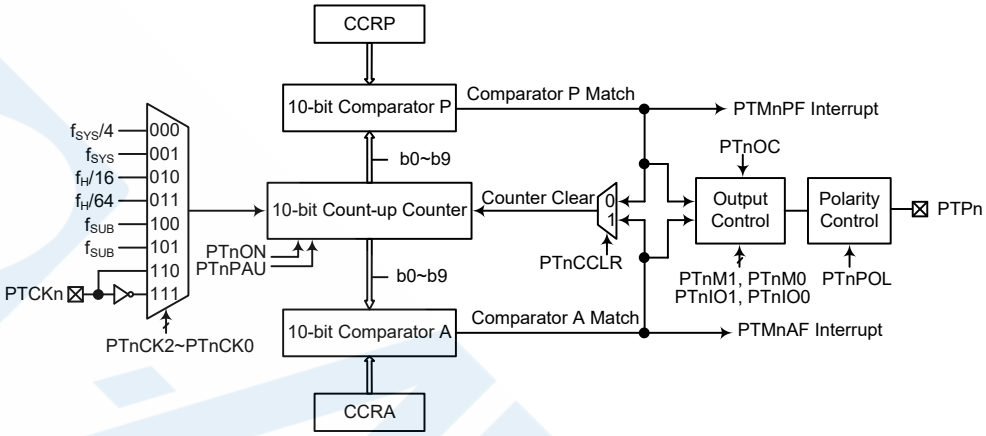


Single Pulse Output Mode

Note: 1. Counter stopped by CCRA
2. CCRP is not used
3. The pulse triggered by the STCK pin or by setting the STON bit high
4. An STCK pin active edge will automatically set the STON bit high.
5. In the Single Pulse Output Mode, STIO[1:0] must be set to “11” and can not be changed

Periodic Type TM – PTM

The Periodic Type TM contains four operating modes, which are Compare Match Output, Timer/ Event Counter, Single Pulse Output and PWM Output modes. The Periodic TM can also be controlled with one external input pin and can drive one external output pin.



Note: 1. The PTMn external pins are pin-shared with other functions, so before using the PTMn function, ensure that the relevant pin-shared function registers have be set properly to enable the PTMn pin function. The PTCKn pin, if used, must also be set as an input by setting the corresponding bits in the port control register.

2. The PTP1 pin is unbonded and the PTM1 Compare Match Output Mode, PWM Output Mode and Single Pulse Output Mode cannot be used.

10-bit Periodic Type TM Block Diagram (n=0~1)

Periodic Type TM Operation

The size of Periodic TM is 10-bit wide and its core is a 10-bit count-up counter which is driven by a user selectable internal or external clock source. There are also two internal comparators with the names, Comparator A and Comparator P. These comparators will compare the value in the counter with CCRP and CCRA registers. The CCRP and CCRA comparators are 10-bit wide whose value is compared with all counter bits.

The only way of changing the value of the 10-bit counter using the application program is to clear the counter by changing the PTnON bit from low to high. The counter will also be cleared automatically by a counter overflow or a compare match with one of its associated comparators. When these conditions occur, a PTMn interrupt signal will also usually be generated. The Periodic Type TM can operate in a number of different operational modes, can be driven by different clock sources including an input pin and can also control an output pin. All operating setup conditions are selected using relevant internal registers.

Periodic Type TM Register Description

Overall operation of the Periodic TM is controlled using a series of registers. A read only register pair exists to store the internal counter 10-bit value, while two read/write register pairs exist to store the internal 10-bit CCRA and CCRP value. The remaining two registers are control registers which setup the different operating and control modes.

Register Name	Bit							
	7	6	5	4	3	2	1	0
PTMnC0	PTnPAU	PTnCK2	PTnCK1	PTnCK0	PTnON	—	—	—
PTMnC1	PTnM1	PTnM0	PTnIO1	PTnIO0	PTnOC	PTnPOL	D1	PTnCCLR
PTMnDL	D7	D6	D5	D4	D3	D2	D1	D0
PTMnDH	—	—	—	—	—	—	D9	D8
PTMnAL	D7	D6	D5	D4	D3	D2	D1	D0
PTMnAH	—	—	—	—	—	—	D9	D8
PTMnRPL	PTnRP7	PTnRP6	PTnRP5	PTnRP4	PTnRP3	PTnRP2	PTnRP1	PTnRP0
PTMnRPH	—	—	—	—	—	—	PTnRP9	PTnRP8

10-bit Periodic TM Register List (n=0~1)

- **PTMnC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PTnPAU	PTnCK2	PTnCK1	PTnCK0	PTnON	—	—	—
R/W	R/W	R/W	R/W	R/W	R/W	—	—	—
POR	0	0	0	0	0	—	—	—

Bit 7 **PTnPAU**: PTMn counter pause control
0: Run
1: Pause
The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the PTMn will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

Bit 6~4 **PTnCK2~PTnCK0**: PTMn counter clock selection
000: $f_{SYS}/4$
001: f_{SYS}
010: $f_H/16$
011: $f_H/64$
100: f_{SUB}
101: f_{SUB}
110: PTCKn rising edge clock
111: PTCKn falling edge clock
These three bits are used to select the clock source for the PTMn. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source f_{SYS} is the system clock, while f_H and f_{SUB} are other internal clocks, the details of which can be found in the oscillator section.

Bit 3 **PTnON**: PTMn counter on/off control
0: Off
1: On
This bit controls the overall on/off function of the PTMn. Setting the bit high enables the counter to run while clearing the bit disables the PTMn. Clearing this bit to zero will stop the counter from counting and turn off the PTMn which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again.
If the PTMn is in the Compare Match Output Mode or PWM output Mode or Single Pulse Output Mode, then the PTMn output pin will be reset to its initial condition, as specified by the PTnOC bit, when the PTnON bit changes from low to high.

Bit 2~0 Unimplemented, read as “0”

• **PTMnC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	PTnM1	PTnM0	PTnIO1	PTnIO0	PTnOC	PTnPOL	D1	PTnCCLR
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PTnM1~PTnM0**: PTMn operating mode selection
00: Compare Match Output Mode
01: Undefined
10: PWM Output Mode or Single Pulse Output Mode
11: Timer/Counter Mode

These bits setup the required operating mode for the PTMn. To ensure reliable operation the PTMn should be switched off before any changes are made to the PTnM1 and PTnM0 bits. In the Timer/Counter Mode, the PTMn output pin state is undefined.

Bit 5~4 **PTnIO1~PTnIO0**: PTMn external pin function selection

Compare Match Output Mode
00: No change
01: Output low
10: Output high
11: Toggle output

PWM Output Mode/Single Pulse Output Mode
00: PWM output inactive state
01: PWM output active state
10: PWM output
11: Single Pulse Output

Timer/Counter Mode
Unused

These two bits are used to determine how the PTMn output pin changes state when a certain condition is reached. The function that these bits select depends upon in which mode the PTMn is running.

In the Compare Match Output Mode, the PTnIO1 and PTnIO0 bits determine how the PTMn output pin changes state when a compare match occurs from the Comparator A. The PTM output pin can be setup to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the PTMn output pin should be setup using the PTnOC bit in the PTMnC1 register. Note that the output level requested by the PTnIO1 and PTnIO0 bits must be different from the initial value setup using the PTnOC bit otherwise no change will occur on the PTMn output pin when a compare match occurs. After the PTMn output pin changes state, it can be reset to its initial level by changing the level of the PTnON bit from low to high.

In the PWM Output Mode, the PTnIO1 and PTnIO0 bits determine how the PTMn output pin changes state when a certain compare match condition occurs. The PTM output function is modified by changing these two bits. It is necessary to change the values of the PTnIO1 and PTnIO0 bits only after the PTMn has been switched off. Unpredictable PWM outputs will occur if the PTnIO1 and PTnIO0 bits are changed when the PTMn is running.

Bit 3 **PTnOC**: PTMn PTPn output control

Compare Match Output Mode
0: Initial low
1: Initial high

PWM Output Mode/Single Pulse Output Mode
0: Active low
1: Active high

This is the output control bit for the PTMn output pin. Its operation depends upon whether PTMn is being used in the Compare Match Output Mode or in the PWM

Output Mode/Single Pulse Output Mode. It has no effect if the PTMn is in the Timer/ Counter Mode. In the Compare Match Output Mode it determines the logic level of the PTMn output pin before a compare match occurs. In the PWM Output Mode it determines if the PWM signal is active high or active low. In the Single Pulse Output Mode it determines the logic level of the PTMn output pin when the PTnON bit changes from low to high.

Bit 2 **PTnPOL**: PTMn PTPn output polarity control
0: Non-invert
1: Invert
This bit controls the polarity of the PTPn output pin. When the bit is set high the PTMn output pin will be inverted and not inverted when the bit is zero. It has no effect if the PTM is in the Timer/Counter Mode.

Bit 1 **D1**: Reserved, must be fixed at "0"

Bit 0 **PTnCCLR**: PTMn counter clear condition selection
0: Comparator P match
1: Comparator A match
This bit is used to select the method which clears the counter. Remember that the Periodic TM contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the PTnCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The PTnCCLR bit is not used in the PWM Output or Single Pulse Output Mode.

• **PTMnDL Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: PTMn Counter Low Byte Register bit 7 ~ bit 0
PTMn 10-bit Counter bit 7 ~ bit 0

• **PTMnDH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	—	—	—	—	—	R	R
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 **D9~D8**: PTMn Counter High Byte Register bit 1 ~ bit 0
PTMn 10-bit Counter bit 9 ~ bit 8

• **PTMnAL Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: PTMn CCRA Low Byte Register bit 7 ~ bit 0
PTMn 10-bit CCRA bit 7 ~ bit 0

• **PTMnAH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as "0"
Bit 1~0 **D9~D8**: PTMn CCRA High Byte Register bit 1 ~ bit 0
PTMn 10-bit CCRA bit 9 ~ bit 8

• **PTMnRPL Register**

Bit	7	6	5	4	3	2	1	0
Name	PTnRP7	PTnRP6	PTnRP5	PTnRP4	PTnRP3	PTnRP2	PTnRP1	PTnRP0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **PTnRP7~PTnRP0**: PTMn CCRP Low Byte Register bit 7 ~ bit 0
PTMn 10-bit CCRP bit 7 ~ bit 0

• **PTMnRPH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	PTnRP9	PTnRP8
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as "0"
Bit 1~0 **PTnRP9~PTnRP8**: PTMn CCRP High Byte Register bit 1 ~ bit 0
PTMn 10-bit CCRP bit 9 ~ bit 8

Periodic Type TM Operation Modes

The Periodic Type TM can operate in one of four operating modes, Compare Match Output Mode, PWM Output Mode, Single Pulse Output Mode or Timer/Counter Mode. The operating mode is selected using the PTnM1 and PTnM0 bits in the PTMnC1 register.

Compare Match Output Mode

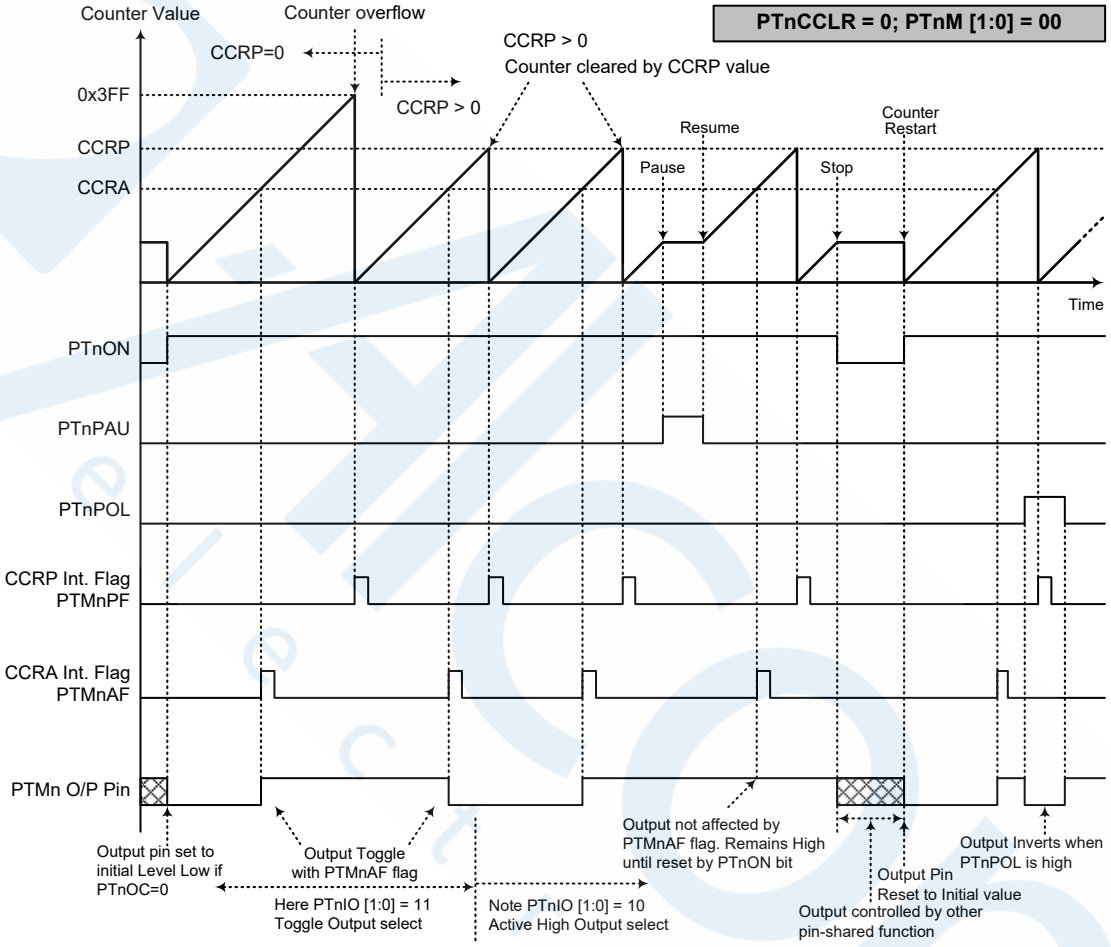
To select this mode, bits PTnM1 and PTnM0 in the PTMnC1 register, should be set to "00" respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the PTnCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both PTMnAF and PTMnPF interrupt request flags for Comparator A and Comparator P respectively, will both be generated.

If the PTnCCLR bit in the PTMnC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the PTMnAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when PTnCCLR is high no PTMnPF interrupt request flag will be generated. In the Compare Match Output Mode, the CCRA can not be set to "0".

If the CCRA bits are all zero, the counter will overflow when its reaches its maximum 10-bit, 3FF Hex, value, however here the PTMnAF interrupt request flag will not be generated.

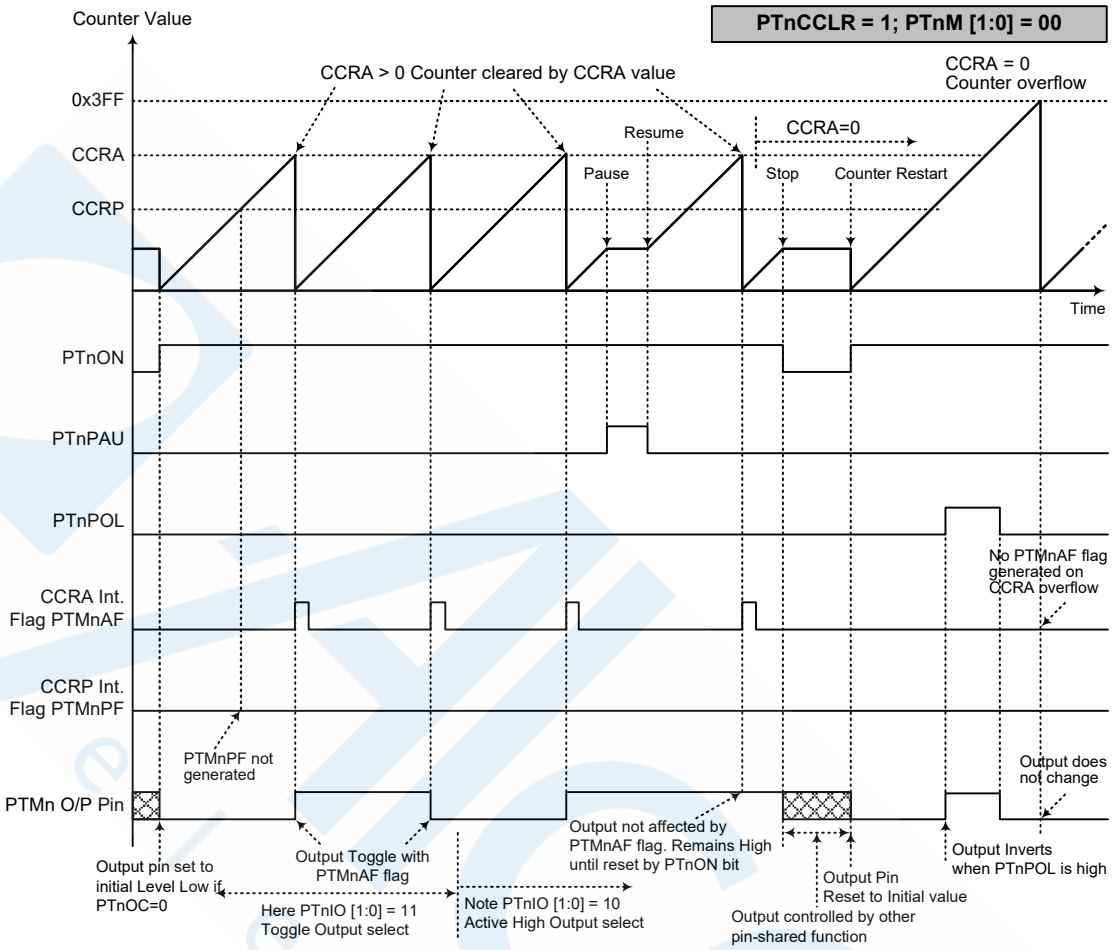
As the name of the mode suggests, after a comparison is made, the PTMn output pin will change state. The PTMn output pin condition however only changes state when a PTMnAF interrupt request

flag is generated after a compare match occurs from Comparator A. The PTMnPF interrupt request flag, generated from a compare match occurs from Comparator P, will have no effect on the PTMn output pin. The way in which the PTMn output pin changes state are determined by the condition of the PTnIO1 and PTnIO0 bits in the PTMnC1 register. The PTMn output pin can be selected using the PTnIO1 and PTnIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the PTMn output pin, which is setup after the PTnON bit changes from low to high, is setup using the PTnOC bit. Note that if the PTnIO1 and PTnIO0 bits are zero then no pin change will take place.



Compare Match Output Mode – PTnCCLR=0 (n=0~1)

Note: 1. With PTnCCLR=0, a Comparator P match will clear the counter
2. The PTMn output pin is controlled only by the PTMnAF flag
3. The output pin is reset to its initial state by a PTnON bit rising edge



Compare Match Output Mode – PTnCCLR=1 (n=0~1)

Note: 1. With PTnCCLR=1, a Comparator A match will clear the counter
2. The PTMn output pin is controlled only by the PTMnAF flag
3. The output pin is reset to its initial state by a PTnON bit rising edge
4. A PTMnPF flag is not generated when PTnCCLR=1

Timer/Counter Mode

To select this mode, bits PTnM1 and PTnM0 in the PTMnC1 register should be set to "11" respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the PTMn output pin is not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the PTMn output pin is not used in this mode, the pin can be used as a normal I/O pin or other pin-shared functions.

PWM Output Mode

To select this mode, bits PTnM1 and PTnM0 in the PTMnC1 register should be set to "10" respectively and also the PTnIO1 and PTnIO0 bits should be set to "10" respectively. The PWM function within the PTMn is useful for applications which require functions such as motor control, heating control, illumination control, etc. By providing a signal of fixed frequency but of varying duty cycle on the PTMn output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output Mode, the PTnCCLR bit has no effect as the PWM period. Both of the CCRP and CCRA registers are used to generate the PWM waveform, one register is used to clear the internal counter and thus control the PWM waveform frequency, while the other one is used to control the duty cycle. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRA and CCRP registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The PTnOC bit in the PTMnC1 register is used to select the required polarity of the PWM waveform while the two PTnIO1 and PTnIO0 bits are used to enable the PWM output or to force the PTMn output pin to a fixed high or low level. The PTnPOL bit is used to reverse the polarity of the PWM output waveform.

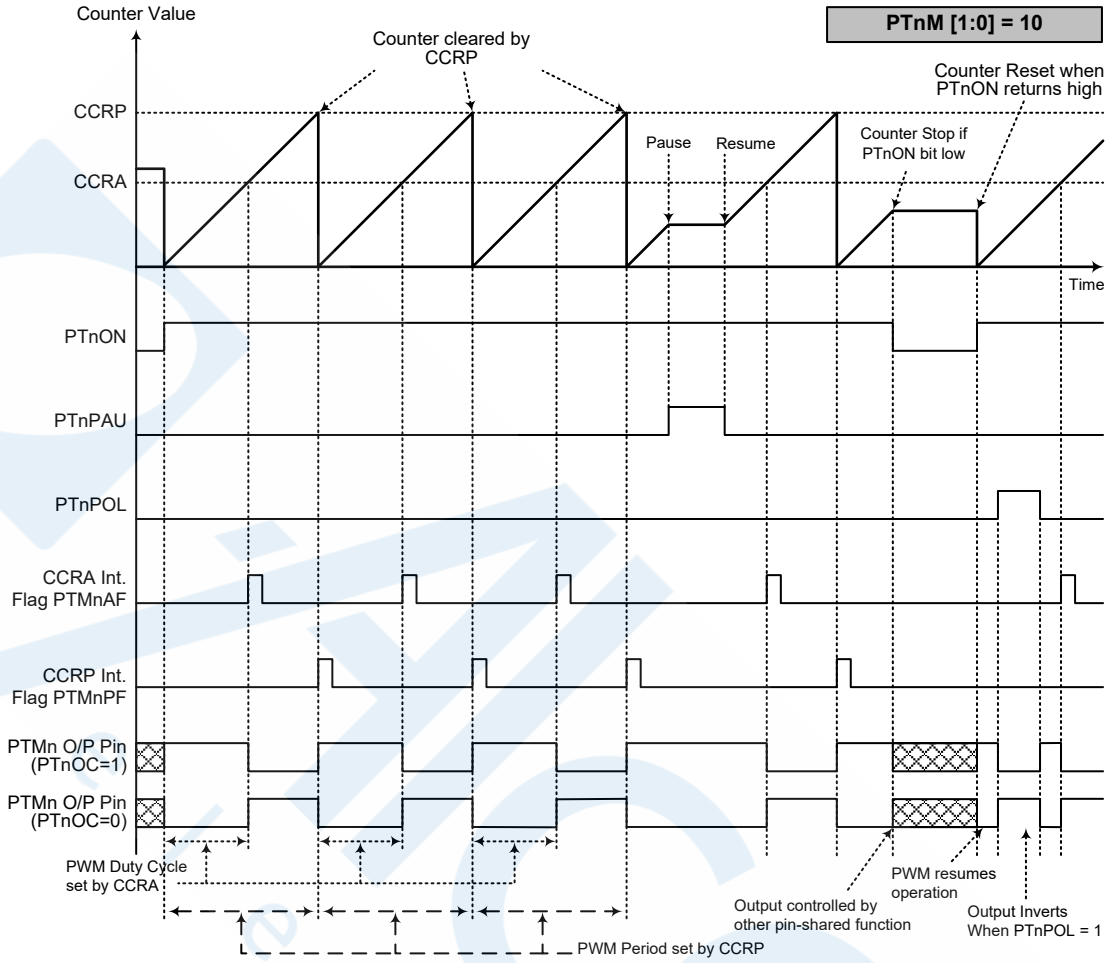
- **10-bit PWM Output Mode, Edge-aligned Mode**

CCRP	1~1023	0
Period	1~1023	1024
Duty	CCRA	

If f_{SYS}=16MHz, PTMn clock source select f_{SYS}/4, CCRP=512 and CCRA=128,

The PTMn PWM output frequency=(f_{SYS}/4)/512=f_{SYS}/2048=7.8125kHz, duty=128/512=25%,

If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.



PWM Output Mode (n=0~1)

Note: 1. The counter is cleared by CCRP

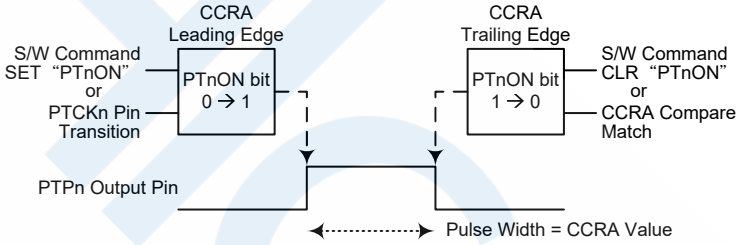
2. A counter clear sets the PWM Period
3. The internal PWM function continues running even when PTnIO[1:0]=00 or 01
4. The PTnCCLR bit has no influence on PWM operation

Single Pulse Output Mode

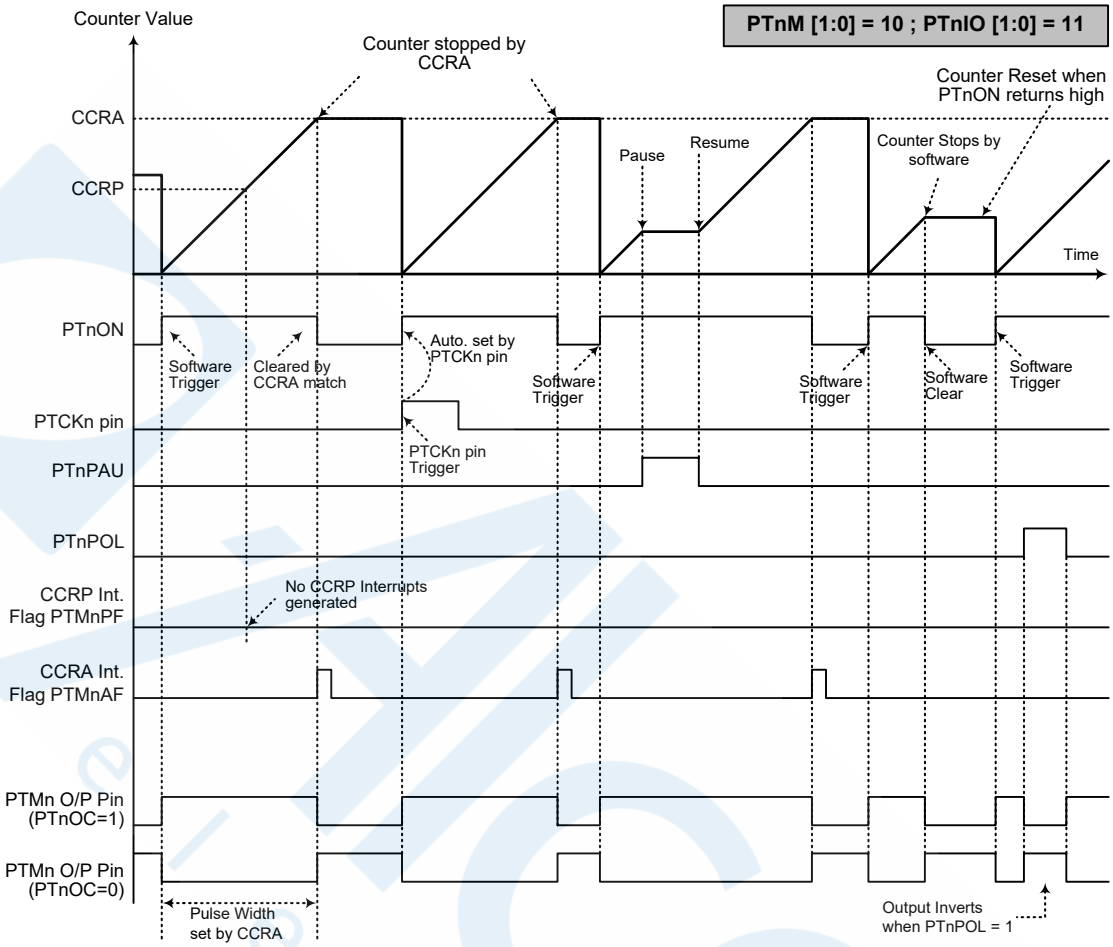
To select this mode, bits PTnM1 and PTnM0 in the PTMnC1 register should be set to "10" respectively and also the PTnIO1 and PTnIO0 bits should be set to "11" respectively. The Single Pulse Output Mode, as the name suggests, will generate a single shot pulse on the PTMn output pin.

The trigger for the pulse output leading edge is a low to high transition of the PTnON bit, which can be implemented using the application program. However in the Single Pulse Output Mode, the PTnON bit can also be made to automatically change from low to high using the external PTCKn pin, which will in turn initiate the Single Pulse output. When the PTnON bit transitions to a high level, the counter will start running and the pulse leading edge will be generated. The PTnON bit should remain high when the pulse is in its active state. The generated pulse trailing edge will be generated when the PTnON bit is cleared to zero, which can be implemented using the application program or when a compare match occurs from Comparator A.

However a compare match from Comparator A will also automatically clear the PTnON bit and thus generate the Single Pulse output trailing edge. In this way the CCRA value can be used to control the pulse width. A compare match from Comparator A will also generate a PTMn interrupt. The counter can only be reset back to zero when the PTnON bit changes from low to high when the counter restarts. In the Single Pulse Output Mode CCRP is not used. The PTnCCLR is not used in this Mode.



Single Pulse Generation (n=0~1)



Single Pulse Output Mode (n=0~1)

Note: 1. Counter stopped by CCRA
2. CCRP is not used
3. The pulse triggered by the PTCKn pin or by setting the PTnON bit high
4. A PTCKn pin active edge will automatically set the PTnON bit high.
5. In the Single Pulse Output Mode, PTnIO[1:0] must be set to “11” and can not be changed

Analog to Digital Converter

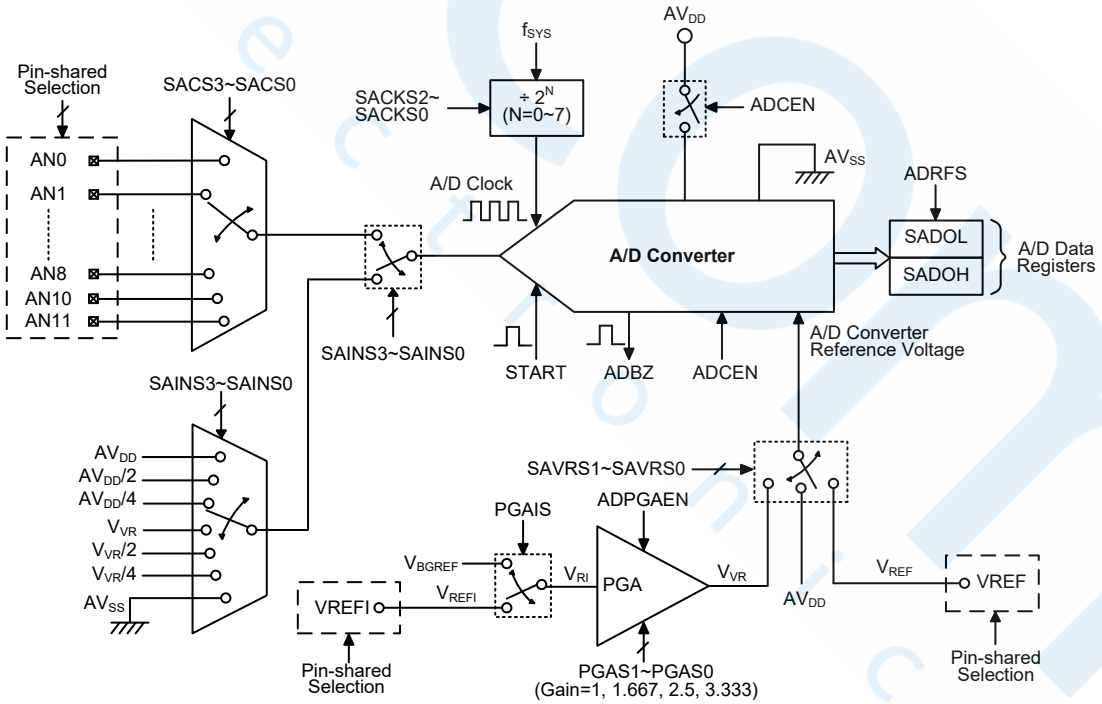
The need to interface to real world analog signals is a common requirement for many electronic systems. However, to properly process these signals by a microcontroller, they must first be converted into digital signals by A/D converters. By integrating the A/D conversion electronic circuitry into the microcontroller, the need for external components is reduced significantly with the corresponding follow-on benefits of lower costs and reduced component space requirements.

A/D Converter Overview

The device contains a multi-channel analog to digital converter which can directly interface to external analog signals, such as that from sensors or other control signals and convert these signals directly into a 12-bit digital value. It also can convert the internal signals, such as the internal A/D converter power supply and the internal PGA output voltage, and convert these signals directly into a 12-bit digital value. The external or internal analog signal to be converted is determined by the SAINS3~SAINS0 bits together with the SACS3~SACS0 bits. When the external analog signal is to be converted, the corresponding pin-shared control bits should first be properly configured and then desired external channel input should be selected using the SAINS3~SAINS0 and SACS3~SACS0 bits. Note that when the internal analog signal is to be converted using the SAINS3~SAINS0 bits, the external channel analog input will be automatically be switched off. More detailed information about the A/D converter input signal is described in the "A/D Converter Control Registers" and "A/D Converter Input Signals" sections respectively.

External Input Channels	Internal Analog Signals	A/D Signal Select
AN0~AN8, AN10~AN11	AV_{DD}, $AV_{DD}/2$, $AV_{DD}/4$, V_{VR}, $V_{VR}/2$, $V_{VR}/4$, AV_{SS}	SAINS3~SAINS0, SACS3~SACS0

The accompanying block diagram shows the overall internal structure of the A/D converter, together with its associated registers.



A/D Converter Structure

A/D Converter Register Description

Overall operation of the A/D converter is controlled using a series of registers. A read only register pair exists to store the A/D converter data 12-bit value. Three registers, SADC0, SADC1 and SADC2, are the control registers which setup the operating conditions and control function of the A/D converter. The VBGRC register contains the VBGREN bit to control the bandgap reference voltage.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SADOL (ADRFS=0)	D3	D2	D1	D0	—	—	—	—
SADOL (ADRFS=1)	D7	D6	D5	D4	D3	D2	D1	D0
SADOH (ADRFS=0)	D11	D10	D9	D8	D7	D6	D5	D4
SADOH (ADRFS=1)	—	—	—	—	D11	D10	D9	D8
SADC0	START	ADBZ	ADCEN	ADRFS	SACS3	SACS2	SACS1	SACS0
SADC1	SAINS3	SAINS2	SAINS1	SAINS0	—	SACKS2	SACKS1	SACKS0
SADC2	ADPGAEN	—	—	PGAIS	SAVRS1	SAVRS0	PGAGS1	PGAGS0
VBGRC	—	—	—	—	—	—	—	VBGREN

A/D Converter Register List

A/D Converter Data Registers – SADOL, SADOH

As the device contains an internal 12-bit A/D converter, it requires two data registers to store the converted value. These are a high byte register, known as SADOH, and a low byte register, known as SADOL. After the conversion process takes place, these registers can be directly read by the microcontroller to obtain the digitised conversion value. As only 12 bits of the 16-bit register space is utilised, the format in which the data is stored is controlled by the ADRFS bit in the SADC0 register as shown in the accompanying table. D0~D11 are the A/D conversion result data bits. Any unused bits will be read as zero. Note that the A/D converter data register contents will keep unchanged if the A/D converter is disabled.

ADRFS	SADOH								SADOL							
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
0	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	0	0	0	0
1	0	0	0	0	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0

A/D Converter Data Registers

A/D Converter Control Registers – SADC0, SADC1, SADC2

To control the function and operation of the A/D converter, several control registers known as SADC0, SADC1, SADC2 are provided. These 8-bit registers define functions such as the selection of which analog channel is connected to the internal A/D converter, the digitised data format, the A/D converter clock source as well as controlling the start function and monitoring the A/D converter busy status. As the device contains only one actual analog to digital converter hardware circuit, each of the external and internal analog signals must be routed to the converter. The SACS3~SACS0 bits in the SADC0 register are used to determine which external channel input is selected to be converted. The SAINS3~SAINS0 bits in the SADC1 register are used to determine that the analog signal to be converted comes from the internal analog signal or external analog channel input. The A/D converter also contains a programmable gain amplifier, PGA, to generate the A/D converter internal reference voltage. The overall operation of the PGA is controlled using the SADC2 register.

The relevant pin-shared function selection bits determine which pins on I/O ports are used as analog inputs for the A/D converter input and which pins are not to be used as the A/D converter input.

When the pin is selected to be an A/D converter input, its original function whether it is an I/O or other pin-shared function will be removed. In addition, any internal pull-high resistor connected to the pin will be automatically removed if the pin is selected to be an A/D converter input.

• **SADC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	START	ADBZ	ADCEN	ADRFS	SACS3	SACS2	SACS1	SACS0
R/W	R/W	R	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **START**: Start the A/D conversion
0→1→0: Start A/D conversion
This bit is used to initiate an A/D conversion process. The bit is normally low but if set high and then cleared low again, the A/D converter will initiate a conversion process.

Bit 6 **ADBZ**: A/D Converter busy flag
0: No A/D conversion is in progress
1: A/D conversion is in progress
This read only flag is used to indicate whether the A/D conversion is in progress or not. When the START bit is set from low to high and then to low again, the ADBZ flag will be set high to indicate that the A/D conversion is initiated. The ADBZ flag will be cleared to zero after the A/D conversion is complete.

Bit 5 **ADCEN**: A/D Converter function enable control
0: Disable
1: Enable
This bit controls the A/D converter internal function. This bit should be set high to enable the A/D converter. If the bit is cleared to zero, then the A/D converter will be switched off reducing the device power consumption. When the A/D converter function is disabled, the contents of the A/D converter data register pair, SADOH and SADOL, will keep unchanged.

Bit 4 **ADRFS**: A/D Converter data format control
0: ADC output data format → SADOH=D[11:4]; SADOL=D[3:0]
1: ADC output data format → SADOH=D[11:8]; SADOL=D[7:0]
This bit controls the format of the 12-bit converted A/D converter value in the two A/D converter data registers. Details are provided in the A/D converter data register section.

Bit 3~0 **SACS3~SACS0**: A/D converter external analog input channel selection
0000: AN0
0001: AN1
0010: AN2
0011: AN3
0100: AN4
0101: AN5
0110: AN6
0111: AN7
1000: AN8
1001: Reserved
1010: AN10
1011: AN11
1100~1111: Non-existed channel, the input will be floating

• **SADC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	SAINS3	SAINS2	SAINS1	SAINS0	—	SACKS2	SACKS1	SACKS0
R/W	R/W	R/W	R/W	R/W	—	R/W	R/W	R/W
POR	0	0	0	0	—	0	0	0

Bit 7~4 **SAINS3~SAINS0**: A/D converter input signal selection
0000: External signal – External analog channel input , ANn
0001: Internal signal – Internal A/D converter power supply voltage AV_{DD}
0010: Internal signal – Internal A/D converter power supply voltage $AV_{DD}/2$
0011: Internal signal – Internal A/D converter power supply voltage $AV_{DD}/4$
0100: External signal – External analog channel input , ANn
0101: Internal signal – Internal signal derived from PGA output V_{VR}
0110: Internal signal – Internal signal derived from PGA output $V_{VR}/2$
0111: Internal signal – Internal signal derived from PGA output $V_{VR}/4$
10xx: Internal signal – connected to ground, AV_{SS}
1100~1111: External signal – External analog channel input , ANn

Care must be taken if the SAINS3~SAINS0 bits are set to "0001"~"0011", "0101"~"0111" to select the internal analog signal to be converted. When the internal analog signal is selected to be converted, the external channel input signal will automatically be switched off regardless of the SACS3~SACS0 bits value. It will prevent the external channel input from being connected together with the internal analog signal.

Bit 3 Unimplemented, read as "0"

Bit 2~0 **SACKS2~SACKS0**: A/D conversion clock source selection
000: f_{SYS}
001: $f_{SYS}/2$
010: $f_{SYS}/4$
011: $f_{SYS}/8$
100: $f_{SYS}/16$
101: $f_{SYS}/32$
110: $f_{SYS}/64$
111: $f_{SYS}/128$

These three bits are used to select the clock source for the A/D converter.

• **SADC2 Register**

Bit	7	6	5	4	3	2	1	0
Name	ADPGAEN	—	—	PGAIS	SAVRS1	SAVRS0	PGAGS1	PGAGS0
R/W	R/W	—	—	R/W	R/W	R/W	R/W	R/W
POR	0	—	—	0	0	0	0	0

Bit 7 **ADPGAEN**: A/D converter PGA enable/disable control
0: Disable
1: Enable

This bit is used to control the A/D converter internal PGA function. When the PGA output voltage is selected as A/D input or A/D reference voltage, the PGA needs to be enabled by setting this bit high. Otherwise the PGA needs to be disabled by clearing the ADPGAEN bit to zero to conserve power.

Bit 6~5 Unimplemented, read as "0"

Bit 4 **PGAIS**: PGA input voltage (V_{RI}) selection
0: From VREFI pin
1: From internal reference voltage V_{BGREF}

This bit is used to select the PGA input voltage source. When the internal reference voltage V_{BGREF} is selected as the PGA input voltage, the external reference voltage on the VREFI pin will be automatically switched off. When this bit is set high to select

V_{BGREF} as PGA input, the internal bandgap reference V_{BGREF} should be enabled by setting the VBGREN bit in the VBGRC register to "1".

Bit 3~2 **SAVRS1~SAVRS0**: A/D converter reference voltage selection
00: Internal A/D converter power, AV_{DD}
01: External VREF pin
1x: Internal PGA output voltage, V_{VR}

These bits are used to select the A/D converter reference voltage source. When the internal A/D converter power supply or PGA output is set as the reference voltage, the reference voltage derived from the external VREF pin will be automatically switched off.

Bit 1~0 **PGAGS1~PGAGS0**: PGA gain select
00: Gain=1
01: Gain=1.667 – V_{VR}=2V as V_{RI}=1.2V
10: Gain=2.5 – V_{VR}=3V as V_{RI}=1.2V
11: Gain=3.333 – V_{VR}=4V as V_{RI}=1.2V

These bits are used to select the PGA gain. Note that here the gain is guaranteed only when the PGA input voltage is equal to 1.2V.

Bandgap Referenc Voltage Control Register – VBGRC

A high performance bandgap voltage reference is included in the device. It has an accurate voltage reference output, V_{BGREF}, when input supply voltage changes or temperature variates. The VBGRC register is used to control the bandgap reference voltage circuit enable or disable.

- **VBGRC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	VBGREN
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

Bit 7~1 Unimplemented, read as "0"

Bit 0 **VBGREN**: Bandgap reference voltage control
0: Disable
1: Enable

This bit is used to enable the internal Bandgap reference circuit. The internal Bandgap reference circuit should first be enabled before the V_{BGREF} voltage is selected to be used. A specific start-up time is necessary for the Bandgap circuit to become stable and accurate. When this bit is cleared to 0, the Bandgap voltage output V_{BGREF} is in a low state.

A/D Converter Reference Voltage

The actual reference voltage supply to the A/D Converter can be supplied from the internal A/D converter power, AV_{DD}, an external reference source supplied on pin VREF or an internal reference voltage V_{VR} determined by the SAVRS1~SAVRS0 bits in the SADC2 register. The internal reference voltage V_{VR} is derived from a programmable gain amplifier, PGA, which is controlled by the ADPGAEN bit in the SADC2 register. The PGA gain can be equal to 1, 1.667, 2.5 or 3.333 and selected using the PGAGS1~PGAGS0 bits in the SADC2 register. The PGA input can come from the external reference input pin, VREFI, or an internal Bandgap reference voltage, V_{BGREF}, selected by the PGAIS bit in the SADC2 register. Note that the internal Bandgap reference circuit should first be enabled before the V_{BGREF} is selected to be used.

As the VREFI and VREF pin both are pin-shared with other functions, when the VREFI or VREF pin is selected as the reference voltage pin, the VREFI or VREF pin-shared function selection bits should first be properly configured to disable other pin-shared functions. However, if the internal

reference signal is selected as the reference source, the external reference voltage input from the VREF or VREFI pin will automatically be switched off by hardware.

The analog input values must not be allowed to exceed the value of the selected A/D reference voltage.

SAVRS[1:0]	Reference	Description
00	AV_{DD}	Internal A/D converter power supply voltage AV_{DD}
01	VREF pin	External A/D converter reference pin VREF
10 or 11	V_{VR}	Internal A/D converter PGA output voltage

A/D Converter Reference Voltage Selection

A/D Converter Input Signals

All of the external A/D analog input pins are pin-shared with the I/O pins as well as other functions. The corresponding pin-shared function selection bits in the PxS1 and PxS0 registers, determine whether the external input pins are set as A/D converter analog channel inputs or whether they have other functions. If the corresponding pin is setup to be an A/D converter analog channel input, the original pin functions will be disabled. In this way, pins can be changed under program control to change their function between A/D inputs and other functions. All pull-high resistors, which are setup through register programming, will be automatically disconnected if the pins are setup as A/D converter inputs. Note that it is not necessary to first setup the A/D pin as an input in the port control register to enable the A/D converter input as when the relevant A/D converter input function selection bits enable an A/D converter input, the status of the port control register will be overridden.

If the SAINS3~SAINS0 bits are set to "0000", "0100" or "1100~1111", the external analog channel input is selected to be converted and the SACS3~SACS0 bits can determine which actual external channel is selected to be converted. If the SAINS3~SAINS0 bits are set to other values, the internal analog signal will be selected. If the internal analog signal is selected to be converted, the external input channel will automatically be switched off regardless of the SACS3~SACS0 bits value. It will prevent the external channel input from being connected together with the internal analog signal.

SAINS[3:0]	SACS[3:0]	Input Signals	Description
0000, 0100, 1100~1111	0000~1011	AN0~AN8, AN10~AN11	External channel analog input ANn
	1100~1111	—	Floating, no external channel is selected
0001	xxxx	AV_{DD}	Internal A/D converter power supply voltage AV_{DD}
0010	xxxx	$AV_{DD}/2$	Internal A/D converter power supply voltage $AV_{DD}/2$
0011	xxxx	$AV_{DD}/4$	Internal A/D converter power supply voltage $AV_{DD}/4$
0101	xxxx	V_{VR}	Internal A/D converter PGA output V_{VR}
0110	xxxx	$V_{VR}/2$	Internal A/D converter PGA output $V_{VR}/2$
0111	xxxx	$V_{VR}/4$	Internal A/D converter PGA output $V_{VR}/4$
10xx	xxxx	AV_{SS}	Connected to the ground

"x": Don't care

A/D Converter Input Signal Selection

A/D Converter Operation

The START bit in the SADC0 register is used to start the A/D conversion. When the microcontroller sets this bit from low to high and then low again, an analog to digital conversion cycle will be initiated.

The ADBZ bit in the SADC0 register is used to indicate whether the analog to digital conversion process is in process or not. This bit will be automatically set to "1" by the microcontroller after an A/D conversion is successfully initiated. When the A/D conversion is complete, the ADBZ will be

cleared to "0". In addition, the corresponding A/D converter interrupt request flag will be set in the interrupt control register, and if the interrupts are enabled, an appropriate internal interrupt signal will be generated. This A/D converter internal interrupt signal will direct the program flow to the associated A/D converter internal interrupt address for processing. If the A/D converter internal interrupt is disabled, the microcontroller can be used to poll the ADBZ bit in the SADC0 register to check whether it has been cleared as an alternative method of detecting the end of an A/D conversion cycle.

The clock source for the A/D converter, which originates from the system clock f_{SYS}, can be chosen to be either f_{SYS} or a subdivided version of f_{SYS}. The division ratio value is determined by the SACKS2~SACKS0 bits in the SADC1 register. Although the A/D conversion clock source is determined by the system clock f_{SYS}, and by bits SACKS2~SACKS0, there are some limitations on the A/D conversion clock source speed that can be selected. As the recommended value of permissible A/D conversion clock period, t_{ADCK}, is from 0.5μs to 10μs, care must be taken for system clock frequencies. For example, if the system clock operates at a frequency of 8MHz, the SACKS2~SACKS0 bits should not be set to "000", "001" or "111". Doing so will give A/D conversion clock periods that are less than the minimum A/D conversion clock period or greater than the maximum A/D conversion clock period which may result in inaccurate A/D conversion values. Refer to the following table for examples, where values marked with an asterisk * special care must be taken, as the values may be less or larger than the specified A/D clock period range.

f_{SYS}	A/D Conversion Clock Period (t_{ADCK})							
	SACKS[2:0] =000 (f_{SYS})	SACKS[2:0] =001 (f_{SYS}/2)	SACKS[2:0] =010 (f_{SYS}/4)	SACKS[2:0] =011 (f_{SYS}/8)	SACKS[2:0] =100 (f_{SYS}/16)	SACKS[2:0] =101 (f_{SYS}/32)	SACKS[2:0] =110 (f_{SYS}/64)	SACKS[2:0] =111 (f_{SYS}/128)
1MHz	1μs	2μs	4μs	8μs	16μs*	32μs*	64μs*	128μs*
2MHz	500ns	1μs	2μs	4μs	8μs	16μs*	32μs*	64μs*
4MHz	250ns*	500ns	1μs	2μs	4μs	8μs	16μs*	32μs*
8MHz	125ns*	250ns*	500ns	1μs	2μs	4μs	8μs	16μs*
12MHz	83ns*	167ns*	333ns*	667ns	1.33μs	2.67μs	5.33μs	10.67μs*
16MHz	62.5ns*	125ns*	250ns*	500ns	1μs	2μs	4μs	8μs

A/D Conversion Clock Period Examples

Controlling the power on/off function of the A/D conversion circuitry is implemented using the ADCEN bit in the SADC0 register. This bit must be set high to power on the A/D converter. When the ADCEN bit is set high to power on the A/D conversion internal circuitry, a certain delay as indicated in the timing diagram must be allowed before an A/D conversion is initiated. Even if no pins are selected for use as A/D converter inputs by configuring the corresponding pin control bits, if the ADCEN bit is high then some power will still be consumed. In power conscious applications it is therefore recommended that the ADCEN is set low to reduce power consumption when the A/D converter function is not being used.

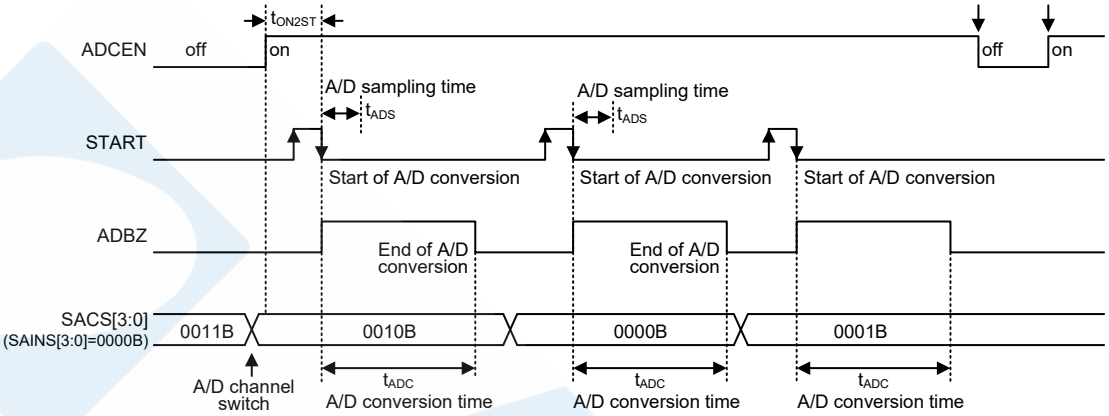
A/D Conversion Rate and Timing Diagram

A complete A/D conversion contains two parts, data sampling and data conversion. The data sampling which is defined as t_{ADS} takes 4 A/D clock periods and the data conversion takes 12 A/D converter clock periods. Therefore a total of 16 A/D conversion clock periods for an A/D conversion which is defined as t_{ADC} are necessary.

Maximum single A/D conversion rate = 1/(A/D conversion clock period × 16)

The accompanying diagram shows graphically the various stages involved in an analog to digital conversion process and its associated timing. After an A/D conversion process has been initiated by the application program, the microcontroller internal hardware will begin to carry out the

conversion, during which time the program can continue with other functions. The time taken for the A/D conversion is $16t_{ADCK}$ where t_{ADCK} is equal to the A/D conversion clock period.



A/D Conversion Timing – External Channel Input

Summary of A/D Conversion Steps

The following summarises the individual steps that should be executed in order to implement an A/D conversion process.

- Step 1

  Select the required A/D conversion clock by properly programming the SACKS2~SACKS0 bits in the SADC1 register.

- Step 2

  Enable the A/D converter by setting the ADCEN bit in the SADC0 register to "1".

- Step 3

  Select which signal is to be connected to the internal A/D converter by correctly configuring the SAINS3~SAINS0 bits.

  Select the external channel input to be converted, go to Step 4.

  Select the internal analog signal to be converted, go to Step 5.

- Step 4

  If the A/D input signal comes from the external channel input selected by configuring the SAINS3~SAINS0, the corresponding pin should be configured as an A/D input function by configuring the relevant pin-shared function control bits. The desired external channel then should be selected by configuring the SACS3~SACS0. After this step, go to Step 6.

- Step 5

  If the A/D input signal is selected to come from the internal analog signal by configuring the SAINS3~SAINS0 and the external channel analog signal input will be automatically switched off regardless of the SACS3~SACS0 bits value. After this step, go to Step 6.

- Step 6

  Select the reference voltage source by configuring the SAVRS1~SAVRS0 bits in the SADC2 register. Select the PGA input signal and the desired PGA gain if the PGA output voltage, V_{VR}, is selected as the A/D converter reference voltage.

- Step 7

  Select A/D converter output data format by configuring the ADRFS bit in the SADC0 register.

- Step 8

  If A/D converter interrupt is used, the interrupt control registers must be correctly configured to ensure the A/D converter interrupt function is active. The master interrupt control bit, EMI, and the A/D converter interrupt bit, ADE, must both set high in advance.

- Step 9

  The A/D conversion procedure can now be initialised by setting the START bit from low to high and then low again.

- Step 10

  If A/D conversion is in progress, the ADBZ flag will be set high. After the A/D conversion process is completed, the ADBZ flag will go low and then output data can be read from the SADOH and SADOL registers.

Note: When checking for the end of the conversion process, if the method of polling the ADBZ bit in the SADC0 register is used, the interrupt enable step above can be omitted.

Programming Considerations

During microcontroller operations where the A/D converter is not being used, the A/D conversion internal circuitry can be switched off to reduce power consumption by setting the ADCEN bit low in the SADC0 register. When this happens, the internal A/D conversion circuits will not consume power irrespective of what analog voltage is applied to their input lines. If the A/D converter input lines are used as normal I/Os, then care must be taken as if the input voltage is not at a valid logic level, then this may lead to some increase in power consumption.

A/D Conversion Function

As the device contains a 12-bit A/D converter, its full-scale converted digitised value is equal to FFFH. Since the full-scale analog input value is equal to the actual A/D converter reference voltage, V_{REF}, this gives a single bit analog input value of V_{REF} divided by 4096.

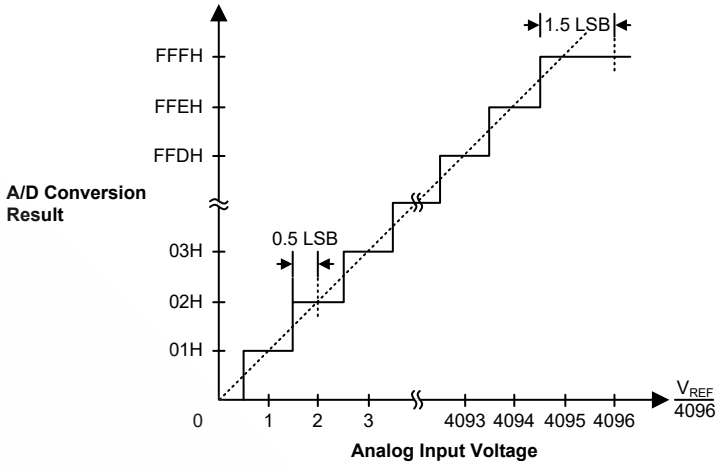
$$1\ LSB=V_{REF} \div 4096$$

The A/D Converter input voltage value can be calculated using the following equation:

$$\text{A/D converter input voltage}=\text{A/D converter output digital value} \times V_{REF} \div 4096$$

The diagram shows the ideal transfer function between the analog input value and the digitised output value for the A/D converter. Except for the digitised zero value, the subsequent digitised values will change at a point 0.5 LSB below where they would change without the offset, and the last full scale digitised value will change at a point 1.5 LSB below the V_{REF} level.

Note that here the V_{REF} voltage is the actual A/D converter reference voltage determined by the SAVRS bit field.



Ideal A/D Conversion Function

A/D Converter Programming Examples

The following two programming examples illustrate how to setup and implement an A/D conversion. In the first example, the method of polling the ADBZ bit in the SADC0 register is used to detect when the conversion cycle is complete, whereas in the second example, the A/D converter interrupt is used to determine when the conversion is complete.

Example: using an ADBZ polling method to detect the end of conversion

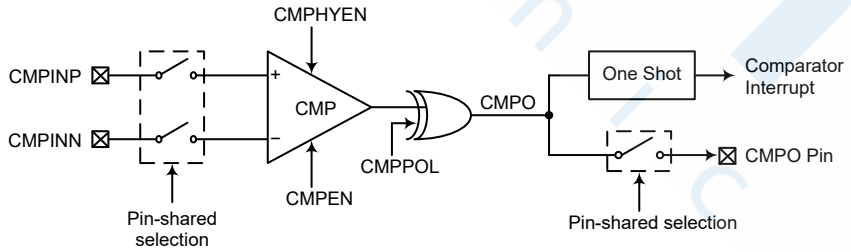
```
clr  ADE                ; disable ADC interrupt
mov  a,03H
mov  SADC1,a            ; select input signal from external channel input, fSYS/8 as
                        ; A/D clock
mov  a,00H
mov  SADC2,a            ; select reference voltage from AVDD
mov  a,02h              ; setup PBS0 to configure pin AN0
mov  PBS0,a
mov  a,20h
mov  SADC0,a            ; enable A/D converter and connect AN0 channel to A/D
                        ; converter
:
start_conversion:
clr  START              ; high pulse on start bit to initiate conversion
set  START              ; reset A/D converter
clr  START              ; start A/D conversion
polling_EOC:
sz   ADBZ               ; poll the SADC0 register ADBZ bit to detect end of A/D
                        ; conversion
jmp  polling_EOC        ; continue polling
mov  a,SADOL            ; read low byte conversion result value
mov  SADOL_buffer,a     ; save result to user defined register
mov  a,SADOH            ; read high byte conversion result value
mov  SADOH_buffer,a     ; save result to user defined register
:
jmp  start_conversion   ; start next A/D conversion
```

Example: using the interrupt method to detect the end of conversion

```
clr  ADE                ; disable ADC interrupt
mov  a,03H
mov  SADC1,a            ; select input signal from external channel input, fSYS/8 as
                        ; A/D clock
mov  a,00H
mov  SADC2,a            ; select reference voltage from AVDD
mov  a,02h              ; setup PBS0 to configure pin AN0
mov  PBS0,a
mov  a,20h
mov  SADC0,a            ; enable A/D converter and connect AN0 channel to A/D
                        ; converter
Start_conversion:
clr  START              ; high pulse on START bit to initiate conversion
set  START              ; reset A/D converter
clr  START              ; start A/D conversion
clr  ADF                ; clear ADC interrupt request flag
set  ADE                ; enable ADC interrupt
set  EMI                ; enable global interrupt
:
:
; ADC interrupt service routine
ADC_ISR:
mov  acc_stack,a        ; save ACC to user defined memory
mov  a,STATUS
mov  status_stack,a     ; save STATUS to user defined memory
:
:
mov  a,SADOL            ; read low byte conversion result value
mov  SADOL_buffer,a     ; save result to user defined register
mov  a,SADOH            ; read high byte conversion result value
mov  SADOH_buffer,a     ; save result to user defined register
:
:
EXIT_INT_ISR:
mov  a,status_stack
mov  STATUS,a           ; restore STATUS from user defined memory
mov  a,acc_stack        ; restore ACC from user defined memory
reti
```

Comparator

An analog comparator is contained within the device. The comparator function offers flexibility via their register controlled features such as power-down, polarity select, hysteresis etc. In sharing their pins with normal I/O pins the comparators do not waste precious I/O pins if there functions are otherwise unused.



Comparator

Comparator Operation

The device contains a comparator function which is used to compare two analog voltages and provide an output based on their input difference.

Any pull-high resistors connected to the shared comparator input pins will be automatically disconnected when the corresponding comparator functional pins are selected. As the comparator inputs approach their switching level, some spurious output signals may be generated on the comparator output due to the slow rising or falling nature of the input signals. This can be minimised by the hysteresis function which will apply a small amount of positive feedback to the comparator. Ideally the comparator should switch at the point where the positive and negative inputs signals are at the same voltage level. However, unavoidable input offsets introduce some uncertainties here. The hysteresis function will also increases the switching offset value. The hysteresis window will be changed for different comparator response time selections. The comparator response time is shown in the comparator electrical characteristics.

Comparator Registers

Full control over each internal comparator is provided via the control register, CMPC. The comparator output is recorded via a bit in the respective control register, but can also be transferred out onto a shared I/O pin. Additional comparator functions include output polarity, response tine and power down control.

- **CMPC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	CMPEN	CMPPOL	CMPO	—	—	—	CMPHYEN
R/W	—	R/W	R/W	R	—	—	—	R/W
POR	—	0	0	0	—	—	—	1

Bit 7 Unimplemented, read as "0"

Bit 6 **CMPEN**: Comparator Enable Control
0: Disable
1: Enable
This is the Comparator on/off control bit. If the bit is zero the comparator will be switched off and no power consumed even if analog voltages are applied to its inputs. For power sensitive applications this bit should be cleared to zero if the comparator is not used or before the device enters the SLEEP or IDLE mode. Note that the comparator output will be set low when this bit is cleared to zero.

Bit 5 **CMPPOL**: Comparator output polarity Control
0: Output is not inverted
1: Output is inverted
This is the Comparator polarity control bit. If the bit is zero then the comparator output bit, CMPO, will reflect the non-inverted output condition of the comparator. If the bit is high the comparator output bit will be inverted.

Bit 4 **CMPO**: Comparator output bit
If CMPPOL=0,
0: CMPINP < CMPINN
1: CMPINP > CMPINN
If CMPPOL=1,
0: CMPINP > CMPINN
1: CMPINP < CMPINN
This bit stores the Comparator output bit. The polarity of the bit is determined by the voltages on the comparator inputs and by the condition of the CMPPOL bit.

Bit 3~1 Unimplemented, read as "0"

Bit 0 **CMPHYEN**: Comparator hysteresis voltage control
0: Comparator hysteresis disable
1: Comparator hysteresis enable
Refer to Comparator Characteristics section.

Comparator Interrupt

The comparator has its own interrupt function. When the comparator output bit changes state, its relevant interrupt flag will be set, and if the corresponding interrupt enable bit is set, then a jump to its relevant interrupt vector will be executed. If the microcontroller is in the SLEEP or IDLE Mode and the comparator is enabled, then if the external input lines cause the comparator output bit to change state, the resulting generated interrupt flag will also generate a wake-up. If it is required to disable a wake-up from occurring, then the interrupt flag should be first set high before entering the SLEEP or IDLE Mode.

Programming Considerations

If the comparator is enabled, it will remain active when the microcontroller enters the SLEEP or IDLE Mode, however as it will consume a certain amount of power, the user may wish to consider disabling it before the SLEEP or IDLE Mode is entered.

As comparator pins are shared with normal I/O pins the I/O registers for these pins will be read as zero (port control register is "1") or read as port data register value (port control register is "0") if the comparator function is enabled.

Serial Interface Module – SIM

The device contains a Serial Interface Module, which includes the four-line SPI interface or two-line I^2C interface types, to allow an easy method of communication with external peripheral hardware. Having relatively simple communication protocols, these serial interface types allow the microcontroller to interface to external SPI or I^2C based hardware such as sensors, Flash or EEPROM memory, etc. The SIM interface pins are pin-shared with other I/O pins and therefore the SIM interface functional pins must first be selected using the corresponding pin-shared function selection bits. As both interface types share the same pins and registers, the choice of whether the SPI or I^2C type is used is made using the SIM operating mode control bits, named SIM2~SIM0, in the SIMC0 register. These pull-high resistors of the SIM pin-shared I/O pins are selected using pull-high control registers when the SIM function is enabled and the corresponding pins are used as SIM input pins.

SPI Interface

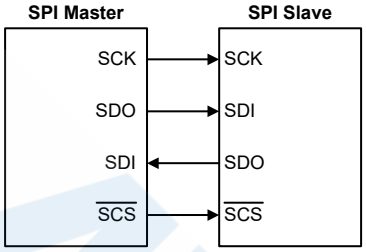
The SPI interface is often used to communicate with external peripheral devices such as sensors, Flash or EEPROM memory devices, etc. Originally developed by Motorola, the four-line SPI interface is a synchronous serial data interface that has a relatively simple communication protocol simplifying the programming requirements when communicating with external hardware devices.

The communication is full duplex and operates as a slave/master type, where the device can be either master or slave. Although the SPI interface specification can control multiple slave devices from a single master, the device provides only one $\overline{SCS}$ pin. If the master needs to control multiple slave devices from a single master, the master can use I/O pin to select the slave devices.

SPI Interface Operation

The SPI interface is a full duplex synchronous serial data link. It is a four-line interface with pin names SDI, SDO, SCK and $\overline{SCS}$. Pins SDI and SDO are the Serial Data Input and Serial Data

Output lines, the SCK pin is the Serial Clock line and $\overline{SCS}$ is the Slave Select line. As the SPI interface pins are pin-shared with normal I/O pins and with the I^2C function pins, the SPI interface pins must first be selected by setting correct bits in the SIMC0 and SIMC2 registers. After the desired SPI configuration has been set it can be disabled or enabled using the SIMEN bit in the SIMC0 register. Communication between devices connected to the SPI interface is carried out in a slave/master mode with all data transfer initiations being implemented by the master. The Master also controls the clock signal. As the device only contains a single $\overline{SCS}$ pin only one slave device can be utilized. The $\overline{SCS}$ pin is controlled by software, set CSEN bit to 1 to enable $\overline{SCS}$ pin function, set CSEN bit to 0 the $\overline{SCS}$ pin will be floating state.

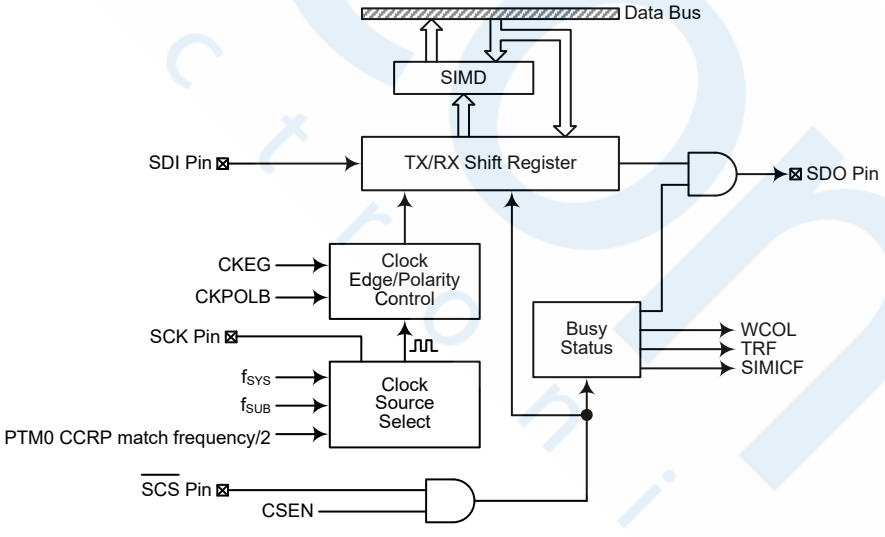


SPI Master/Slave Connection

The SPI function in this device offers the following features:

- Full duplex synchronous data transfer
- Both Master and Slave modes
- LSB first or MSB first data transmission modes
- Transmission complete flag
- Rising or falling active clock edge

The status of the SPI interface pins is determined by a number of factors such as whether the device is in the master or slave mode and upon the condition of certain control bits such as CSEN and SIMEN.



SPI Block Diagram

SPI Registers

There are three internal registers which control the overall operation of the SPI interface. These are

the SIMD data register and two registers SIMC0 and SIMC2. The SIMC1 register is only used by the I²C interface.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SIMC0	SIM2	SIM1	SIM0	—	SIMDEB1	SIMDEB0	SIMEN	SIMICF
SIMC2	D7	D6	CKPOLB	CKEG	MLS	CSEN	WCOL	TRF
SIMD	D7	D6	D5	D4	D3	D2	D1	D0

SPI Register List

SPI Data Register

The SIMD register is used to store the data being transmitted and received. The same register is used by both the SPI and I²C functions. Before the device writes data to the SPI bus, the actual data to be transmitted must be placed in the SIMD register. After the data is received from the SPI bus, the device can read it from the SIMD register. Any transmission or reception of data from the SPI bus must be made via the SIMD register.

- **SIMD Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

"x": unknown

Bit 7~0 **D7~D0**: SIM data register bit 7 ~ bit 0

SPI Control Registers

There are also two control registers for the SPI interface, SIMC0 and SIMC2. Note that the SIMC2 register also has the name SIMA which is used by the I²C function. The SIMC1 register is not used by the SPI function, only by the I²C function. The SIMC0 register is used to control the enable/disable function and to set the data transmission clock frequency. The SIMC2 register is used for other control functions such as LSB/MSB selection, write collision flag etc.

- **SIMC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	SIM2	SIM1	SIM0	—	SIMDEB1	SIMDEB0	SIMEN	SIMICF
R/W	R/W	R/W	R/W	—	R/W	R/W	R/W	R/W
POR	1	1	1	—	0	0	0	0

Bit 7~5 **SIM2~SIM0**: SIM operating mode control
000: SPI master mode; SPI clock is $f_{SYS}/4$
001: SPI master mode; SPI clock is $f_{SYS}/16$
010: SPI master mode; SPI clock is $f_{SYS}/64$
011: SPI master mode; SPI clock is f_{SUB}
100: SPI master mode; SPI clock is PTM0 CCRP match frequency/2
101: SPI slave mode
110: I²C slave mode
111: Unused mode

These bits setup the overall operating mode of the SIM function. As well as selecting if the I²C or SPI function, they are used to control the SPI Master/Slave selection and the SPI Master clock frequency. The SPI clock is a function of the system clock but can also be chosen to be sourced from PTM0 and f_{SUB}. If the SPI Slave Mode is selected then the clock will be supplied by an external Master device.

Bit 4 Unimplemented, read as "0"

Bit 3~2 **SIMDEB1~SIMDEB0**: I²C debounce time selection

These bits are only used in the I²C mode and the detailed definition is described in the I²C function.

Bit 1 **SIMEN**: SIM enable control

0: Disable

1: Enable

The bit is the overall on/off control for the SIM interface. When the SIMEN bit is cleared to zero to disable the SIM interface, the SDI, SDO, SCK and $\overline{SCS}$, or SDA and SCL lines will lose their SPI or I²C function and the SIM operating current will be reduced to a minimum value. When the bit is high the SIM interface is enabled. If the SIM is configured to operate as an SPI interface via the SIM2~SIM0 bits, the contents of the SPI control registers will remain at the previous settings when the SIMEN bit changes from low to high and should therefore be first initialised by the application program. If the SIM is configured to operate as an I²C interface via the SIM2~SIM0 bits and the SIMEN bit changes from low to high, the contents of the I²C control bits such as HTX and TXAK will remain at the previous settings and should therefore be first initialised by the application program while the relevant I²C flags such as HCF, HAAS, HBB, SRW and RXAK will be set to their default states.

Bit 0 **SIMICF**: SIM SPI slave mode incomplete transfer flag

0: SIM SPI slave mode incomplete condition is not occurred

1: SIM SPI slave mode incomplete condition is occurred

This bit is only available when the SIM is configured to operate in an SPI slave mode. If the SPI operates in the slave mode with the SIMEN and CSEN bits both being set to 1 but the $\overline{SCS}$ line is pulled high by the external master device before the SPI data transfer is completely finished, the SIMICF bit will be set to 1 together with the TRF bit set high. When this condition occurs, the corresponding interrupt will occur if the interrupt function is enabled. However, the TRF bit will not be set to 1 if the SIMICF bit is set to 1 by software application program.

- **SIMC2 Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	CKPOLB	CKEG	MLS	CSEN	WCOL	TRF
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **D7~D6**: Undefined bits

These bits can be read or written by application program.

Bit 5 **CKPOLB**: SPI clock line base condition selection

0: The SCK line will be high when the clock is inactive

1: The SCK line will be low when the clock is inactive

The CKPOLB bit determines the base condition of the clock line, if the bit is high, then the SCK line will be low when the clock is inactive. When the CKPOLB bit is low, then the SCK line will be high when the clock is inactive.

Bit 4 **CKEG**: SPI SCK clock active edge type selection

CKPOLB=0

0: SCK is high base level when the clock is inactive and data capture at SCK rising edge

1: SCK is high base level when the clock is inactive and data capture at SCK falling edge

CKPOLB=1

0: SCK is low base level when the clock is inactive and data capture at SCK falling edge

1: SCK is low base level when the clock is inactive and data capture at SCK rising edge

The CKEG and CKPOLB bits are used to setup the way that the clock signal outputs

and inputs data on the SPI bus. These two bits must be configured before data transfer is executed otherwise an erroneous clock edge may be generated. The CKPOLB bit determines the base condition of the clock line, if the bit is high, then the SCK line will be low when the clock is inactive. When the CKPOLB bit is low, then the SCK line will be high when the clock is inactive. The CKEG bit determines active clock edge type which depends upon the condition of CKPOLB bit.

Bit 3 **MLS**: SPI data shift order
0: LSB first
1: MSB first
This is the data shift select bit and is used to select how the data is transferred, either MSB or LSB first. Setting the bit high will select MSB first and low for LSB first.

Bit 2 **CSEN**: SPI $\overline{SCS}$ pin control
0: Disable
1: Enable
The CSEN bit is used as an enable/disable for the $\overline{SCS}$ pin. If this bit is low, then the $\overline{SCS}$ pin will be disabled and placed into a floating condition. If the bit is high the $\overline{SCS}$ pin will be enabled and used as a select pin.

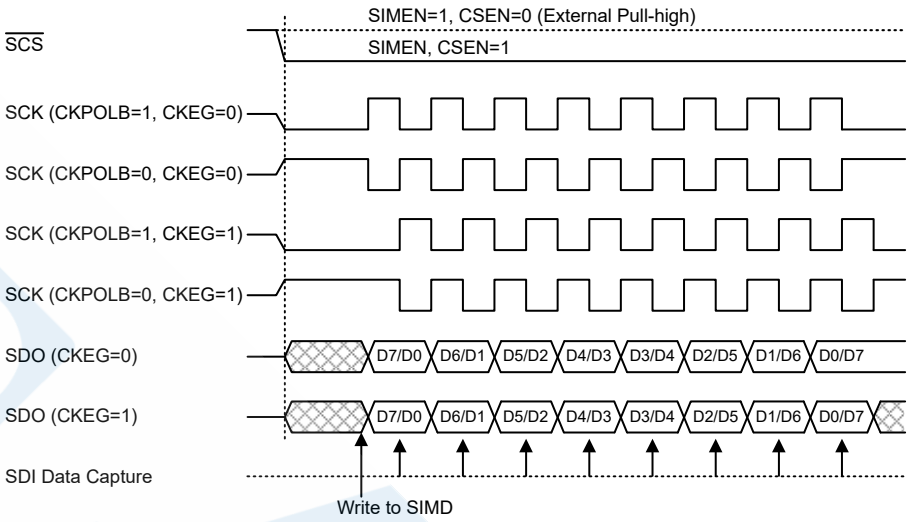
Bit 1 **WCOL**: SPI write collision flag
0: No collision
1: Collision
The WCOL flag is used to detect if a data collision has occurred. If this bit is high it means that data has been attempted to be written to the SIMD register during a data transfer operation. This writing operation will be ignored if data is being transferred. The bit can be cleared by the application program.

Bit 0 **TRF**: SPI Transmit/Receive complete flag
0: SPI data is being transferred
1: SPI data transmission is completed
The TRF bit is the Transmit/Receive Complete flag and is set "1" automatically when an SPI data transmission is completed, but must set to "0" by the application program. It can be used to generate an interrupt.

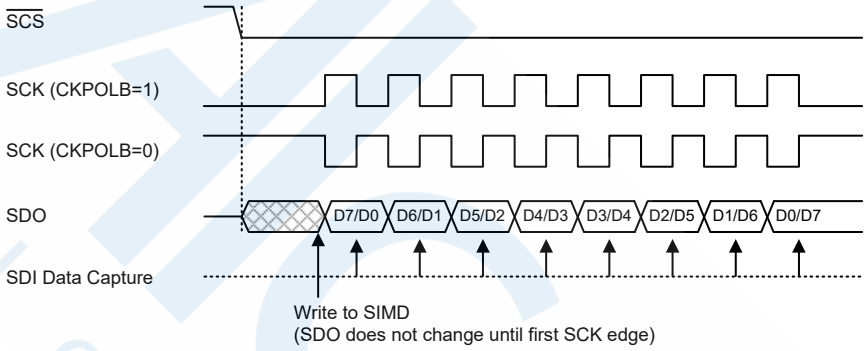
SPI Communication

After the SPI interface is enabled by setting the SIMEN bit high, then in the Master Mode, when data is written to the SIMD register, transmission/reception will begin simultaneously. When the data transfer is complete, the TRF flag will be set automatically, but must be cleared using the application program. In the Slave Mode, when the clock signal from the master has been received, any data in the SIMD register will be transmitted and any data on the SDI pin will be shifted into the SIMD register. The master should output an $\overline{SCS}$ signal to enable the slave devices before a clock signal is provided. The slave data to be transferred should be well prepared at the appropriate moment relative to the SCK signal depending upon the configurations of the CKPOLB bit and CKEG bit. The accompanying timing diagrams show the relationship between the slave data and SCK signal for various configurations of the CKPOLB and CKEG bits.

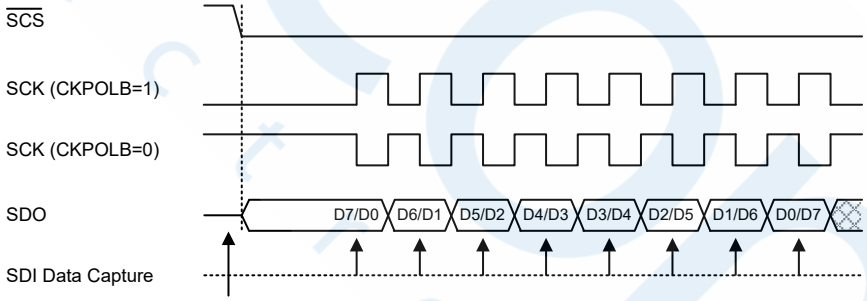
The SPI Master mode will continue to function if the SPI clock is running.



SPI Master Mode Timing

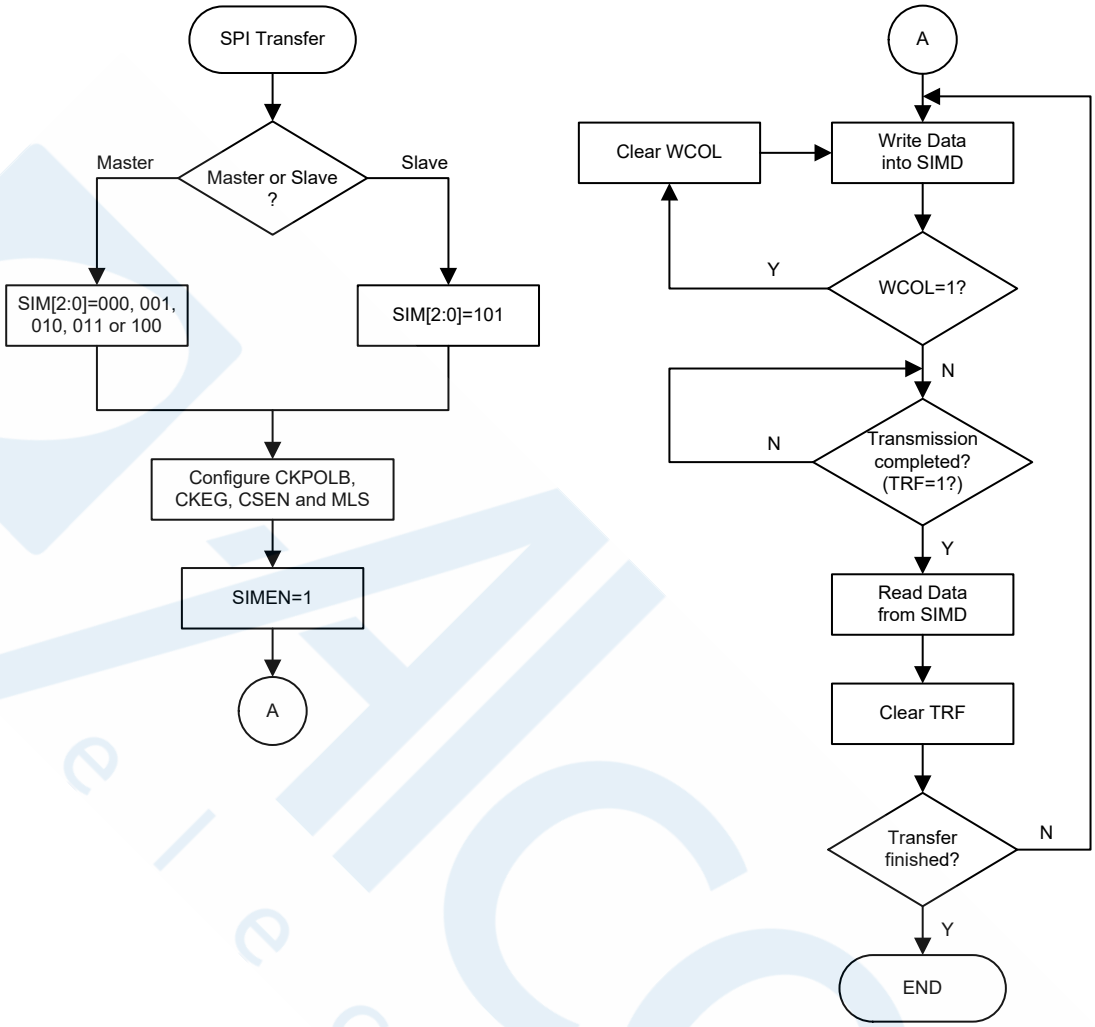


SPI Slave Mode Timing – CKEG=0



Note: For SPI slave mode, if SIMEN=1 and CSEN=0, SPI is always enabled and ignores the $\overline{SCS}$ level.

SPI Slave Mode Timing – CKEG=1



SPI Transfer Control Flow Chart

SPI Bus Enable/Disable

To enable the SPI bus, set CSEN=1 and $\overline{SCS}$=0, then wait for data to be written into the SIMD (TXRX buffer) register. For the Master Mode, after data has been written to the SIMD (TXRX buffer) register, then transmission or reception will start automatically. When all the data has been transferred, the TRF bit should be set. For the Slave Mode, when clock pulses are received on SCK, data in the TXRX buffer will be shifted out or data on SDI will be shifted in.

When the SPI bus is disabled, SCK, SDI, SDO and $\overline{SCS}$ can become I/O pins or other pin-shared functions using the corresponding control bits.

SPI Operation Steps

All communication is carried out using the 4-line interface for either Master or Slave Mode.

The CSEN bit in the SIMC2 register controls the overall function of the SPI interface. Setting this bit high will enable the SPI interface by allowing the $\overline{SCS}$ line to be active, which can then be used to control the SPI interface. If the CSEN bit is low, the SPI interface will be disabled and the $\overline{SCS}$ line will be in a floating condition and can therefore not be used for control of the SPI interface. If the CSEN bit and the SIMEN bit in the SIMC0 are set high, this will place the SDI line in a

floating condition and the SDO line high. If in Master Mode the SCK line will be either high or low depending upon the clock polarity selection bit CKPOLB in the SIMC2 register. If in Slave Mode the SCK line will be in a floating condition. If the SIMEN bit is low, then the bus will be disabled and $\overline{SCS}$, SDI, SDO and SCK will all become I/O pins or the other functions using the corresponding pin-shared control bits. In the Master Mode the Master will always generate the clock signal. The clock and data transmission will be initiated after data has been written into the SIMD register. In the Slave Mode, the clock signal will be received from an external master device for both data transmission and reception. The following sequences show the order to be followed for data transfer in both Master and Slave Mode.

Master Mode

- Step 1

  Select the SPI Master mode and clock source using the SIM2~SIM0 bits in the SIMC0 control register.

- Step 2

  Setup the CSEN bit and setup the MLS bit to choose if the data is MSB or LSB first, this setting must be the same with the Slave devices.

- Step 3

  Setup the SIMEN bit in the SIMC0 control register to enable the SPI interface.

- Step 4

  For write operations: write the data to the SIMD register, which will actually place the data into the TXRX buffer. Then use the SCK and SDO lines to output the data. After this, go to step5.

  For read operations: the data transferred in on the SDI line will be stored in the TXRX buffer until all the data has been received at which point it will be latched into the SIMD register.

- Step 5

  Check the WCOL bit if set high then a collision error has occurred so return to step 4. If equal to zero then go to the following step.

- Step 6

  Check the TRF bit or wait for an SIM SPI serial bus interrupt.

- Step 7

  Read data from the SIMD register.

- Step 8

  Clear TRF.

- Step 9

  Go to step 4.

Slave Mode

- Step 1

  Select the SPI Slave mode using the SIM2~SIM0 bits in the SIMC0 control register

- Step 2

  Setup the CSEN bit and setup the MLS bit to choose if the data is MSB or LSB first, this setting must be the same with the Master devices.

- Step 3

  Setup the SIMEN bit in the SIMC0 control register to enable the SPI interface.

- Step 4

  For write operations: write the data to the SIMD register, which will actually place the data into the TXRX buffer. Then wait for the master clock SCK and $\overline{SCS}$ signal. After this, go to step5.

  For read operations: the data transferred in on the SDI line will be stored in the TXRX buffer until all the data has been received at which point it will be latched into the SIMD register.

- Step 5

  Check the WCOL bit if set high then a collision error has occurred so return to step 4. If equal to zero then go to the following step.

- Step 6

  Check the TRF bit or wait for an SIM SPI serial bus interrupt.

- Step 7

  Read data from the SIMD register.

- Step 8

  Clear TRF.

- Step 9

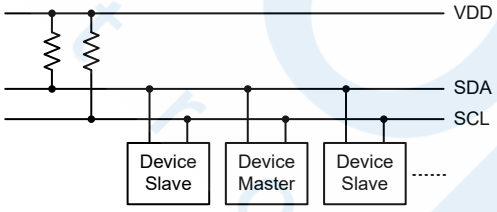
  Go to step 4.

Error Detection

The WCOL bit in the SIMC2 register is provided to indicate errors during data transfer. The bit is set by the SPI serial Interface but must be cleared by the application program. This bit indicates a data collision has occurred which happens if a write to the SIMD register takes place during a data transfer operation and will prevent the write operation from continuing.

I²C Interface

The I²C interface is used to communicate with external peripheral devices such as sensors, EEPROM memory etc. Originally developed by Philips, it is a two-line low speed serial interface for synchronous serial data transfer. The advantage of only two lines for communication, relatively simple communication protocol and the ability to accommodate multiple devices on the same bus has made it an extremely popular interface type for many applications.



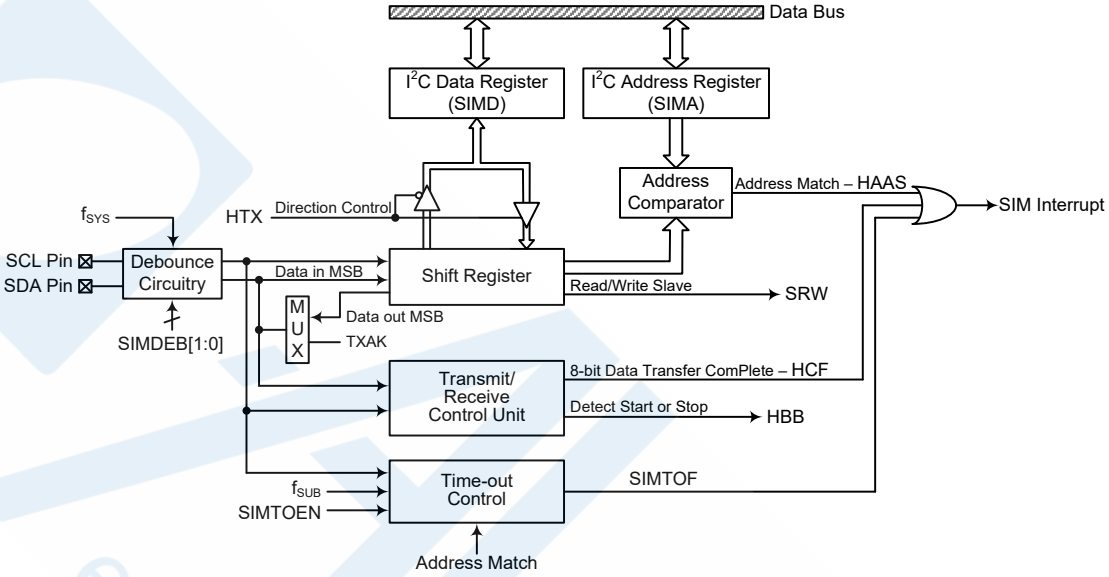
I²C Master Slave Bus Connection

I²C interface Operation

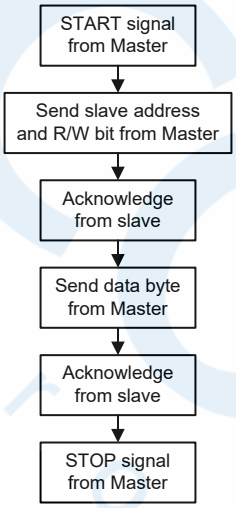
The I²C serial interface is a two-line interface, a serial data line, SDA, and serial clock line, SCL. As many devices may be connected together on the same bus, their outputs are both open drain types. For this reason it is necessary that external pull-high resistors are connected to these outputs. Note that no chip select line exists, as each device on the I²C bus is identified by a unique address which will be transmitted and received on the I²C bus.

When two devices communicate with each other on the bidirectional I²C bus, one is known as the master device and one as the slave device. Both master and slave can transmit and receive data,

however, it is the master device that has overall control of the bus. For the device, which only operates in slave mode, there are two methods of transferring data on the I²C bus, the slave transmit mode and the slave receive mode. The pull-high control function pin-shared with SCL/SDA pin is still applicable even if I²C device is activated and the related internal pull-high function could be controlled by its corresponding pull-high control register.



I²C Block Diagram



I²C Interface Operation

The SIMDEB1 and SIMDEB0 bits determine the debounce time of the I²C interface. This uses the system clock to in effect add a debounce time to the external clock to reduce the possibility of glitches on the clock line causing erroneous operation. The debounce time, if selected, can be chosen to be either 2 or 4 system clocks. To achieve the required I²C data transfer speed, there exists a relationship between the system clock, f_{SYS}, and the I²C debounce time. For either the I²C Standard or Fast mode operation, users must take care of the selected system clock frequency and the configured debounce time to match the criterion shown in the following table.

I²C Debounce Time Selection	I²C Standard Mode (100kHz)	I²C Fast Mode (400kHz)
No Debounce	f_{SYS}>2MHz	f_{SYS}>4MHz
2 system clock debounce	f_{SYS}>4MHz	f_{SYS}>8MHz
4 system clock debounce	f_{SYS}>4MHz	f_{SYS}>8MHz

I²C Minimum f_{SYS} Frequency Requirements

I²C Registers

There are three control registers associated with the I²C bus, SIMC0, SIMC1 and SIMTOC, one slave address register, SIMA, and one data register, SIMD.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SIMC0	SIM2	SIM1	SIM0	—	SIMDEB1	SIMDEB0	SIMEN	SIMICF
SIMC1	HCF	HAAS	HBB	HTX	TXAK	SRW	IAMWU	RXAK
SIMD	D7	D6	D5	D4	D3	D2	D1	D0
SIMA	SIMA6	SIMA5	SIMA4	SIMA3	SIMA2	SIMA1	SIMA0	D0
SIMTOC	SIMTOEN	SIMTOF	SIMTOS5	SIMTOS4	SIMTOS3	SIMTOS2	SIMTOS1	SIMTOS0

I²C Register List

I²C Data Register

The SIMD register is used to store the data being transmitted and received. The same register is used by both the SPI and I²C functions. Before the device writes data to the I²C bus, the actual data to be transmitted must be placed in the SIMD register. After the data is received from the I²C bus, the device can read it from the SIMD register. Any transmission or reception of data from the I²C bus must be made via the SIMD register.

- **SIMD Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

"x": unknown

Bit 7~0 **D7~D0**: SIM data register bit 7 ~ bit 0

I²C Address Register

The SIMA register is also used by the SPI interface but has the name of SIMC2. The SIMA register is the location where the 7-bit slave address of the slave device is stored. Bit 7~bit 1 of the SIMA register define the device slave address. Bit 0 is not defined. When a master device, which is connected to the I²C bus, sends out an address, which matches the slave address in the SIMA register, the slave device will be selected. Note that the SIMA register locates at the same register address as SIMC2 which is used by the SPI interface.

- **SIMA Register**

Bit	7	6	5	4	3	2	1	0
Name	SIMA6	SIMA5	SIMA4	SIMA3	SIMA2	SIMA1	SIMA0	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~1 **SIMA6~SIMA0**: I²C slave address
SIMA6~SIMA0 is the I²C slave address bit 6 ~ bit 0.

Bit 0 **D0**: Reserved bit, can be read or written

I²C Control Registers

There are also three control registers for the I²C interface, SIMC0, SIMC1 and SIMTOC. The register SIMC0 is used to control the enable/disable function and to select the I²C slave mode and debounce time. The SIMC1 register contains the relevant flags which are used to indicate the I²C communication status. The SIMTOC register is used to control the I²C bus time-out function which is described in the corresponding section.

- **SIMC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	SIM2	SIM1	SIM0	—	SIMDEB1	SIMDEB0	SIMEN	SIMICF
R/W	R/W	R/W	R/W	—	R/W	R/W	R/W	R/W
POR	1	1	1	—	0	0	0	0

Bit 7~5 **SIM2~SIM0**: SIM operating mode control
000: SPI master mode; SPI clock is $f_{SYS}/4$
001: SPI master mode; SPI clock is $f_{SYS}/16$
010: SPI master mode; SPI clock is $f_{SYS}/64$
011: SPI master mode; SPI clock is f_{SUB}
100: SPI master mode; SPI clock is PTM0 CCRP match frequency/2
101: SPI slave mode
110: I²C slave mode
111: Unused mode

These bits setup the overall operating mode of the SIM function. As well as selecting if the I²C or SPI function, they are used to control the SPI Master/Slave selection and the SPI Master clock frequency. The SPI clock is a function of the system clock but can also be chosen to be sourced from PTM0 and f_{SUB}. If the SPI Slave Mode is selected then the clock will be supplied by an external Master device.

Bit 4 Unimplemented, read as "0"

Bit 3~2 **SIMDEB1~SIMDEB0**: I²C debounce time selection
00: No debounce
01: 2 system clock debounce
1x: 4 system clock debounce

These bits are used to select the I²C debounce time when the SIM is configured as the I²C interface function by setting the SIM2~SIM0 bits to "110".

Bit 1 **SIMEN**: SIM enable control
0: Disable
1: Enable

The bit is the overall on/off control for the SIM interface. When the SIMEN bit is cleared to zero to disable the SIM interface, the SDI, SDO, SCK and $\overline{SCS}$, or SDA and SCL lines will lose their SPI or I²C function and the SIM operating current will be reduced to a minimum value. When the bit is high the SIM interface is enabled. The SIM configuration option must have first enabled the SIM interface for this bit to be effective. If the SIM is configured to operate as an SPI interface via the SIM2~SIM0 bits, the contents of the SPI control registers will remain at the previous settings when the SIMEN bit changes from low to high and should therefore be first initialised by the application program. If the SIM is configured to operate as an I²C interface via the SIM2~SIM0 bits and the SIMEN bit changes from low to high, the contents of the I²C control bits such as HTX and TXAK will remain at the previous settings and should therefore be first initialised by the application program while the relevant I²C flags such as HCF, HAAS, HBB, SRW and RXAK will be set to their default states.

Bit 0 **SIMICF**: SIM SPI incomplete flag
This bit is only available when the SIM is configured to operate in an SPI slave mode. Refer to the SPI register section.

• **SIMC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	HCF	HAAS	HBB	HTX	TXAK	SRW	IAMWU	RXAK
R/W	R	R	R	R/W	R/W	R/W	R/W	R
POR	1	0	0	0	0	0	0	1

Bit 7 **HCF**: I²C bus data transfer completion flag
0: Data is being transferred
1: Completion of an 8-bit data transfer

The HCF flag is the data transfer flag. This flag will be zero when data is being transferred. Upon completion of an 8-bit data transfer the flag will go high and an interrupt will be generated.

Bit 6 **HAAS**: I²C bus address match flag
0: Not address match
1: Address match

The HAAS flag is the address match flag. This flag is used to determine if the slave device address is the same as the master transmit address. If the addresses match then this bit will be high, if there is no match then the flag will be low.

Bit 5 **HBB**: I²C bus busy flag
0: I²C Bus is not busy
1: I²C Bus is busy

The HBB flag is the I²C busy flag. This flag will be "1" when the I²C bus is busy which will occur when a START signal is detected. The flag will be set to "0" when the bus is free which will occur when a STOP signal is detected.

Bit 4 **HTX**: I²C slave device transmitter/receiver selection
0: Slave device is the receiver
1: Slave device is the transmitter

Bit 3 **TXAK**: I²C bus transmit acknowledge flag
0: Slave send acknowledge flag
1: Slave does not send acknowledge flag

The TXAK bit is the transmit acknowledge flag. After the slave device receipt of 8 bits of data, this bit will be transmitted to the bus on the 9th clock from the slave device. The slave device must always clear the TXAK bit to "0" before further data is received.

Bit 2 **SRW**: I²C slave read/write flag
0: Slave device should be in receive mode
1: Slave device should be in transmit mode

The SRW flag is the I²C Slave Read/Write flag. This flag determines whether the master device wishes to transmit or receive data from the I²C bus. When the transmitted address and slave address match, that is when the HAAS flag is set high, the slave device will check the SRW flag to determine whether it should be in transmit mode or receive mode. If the SRW flag is high, the master is requesting to read data from the bus, so the slave device should be in transmit mode. When the SRW flag is zero, the master will write data to the bus, therefore the slave device should be in receive mode to read this data.

Bit 1 **IAMWU**: I²C address match wake-up control
0: Disable
1: Enable

This bit should be set to 1 to enable the I²C address match wake-up from the SLEEP or IDLE Mode. If the IAMWU bit has been set before entering either the SLEEP or IDLE mode to enable the I²C address match wake-up, then this bit must be cleared to zero by the application program after wake-up to ensure correction device operation.

Bit 0 **RXAK**: I²C bus receive acknowledge flag
0: Slave receives acknowledge flag
1: Slave does not receive acknowledge flag

The RXAK flag is the receiver acknowledge flag. When the RXAK flag is "0", it means that a acknowledge signal has been received at the 9th clock, after 8 bits of data have been transmitted. When the slave device in the transmit mode, the slave device checks the RXAK flag to determine if the master receiver wishes to receive the next byte. The slave transmitter will therefore continue sending out data until the RXAK flag is "1". When this occurs, the slave transmitter will release the SDA line to allow the master to send a STOP signal to release the I²C Bus.

I²C Bus Communication

Communication on the I²C bus requires four separate steps, a START signal, a slave device address transmission, a data transmission and finally a STOP signal. When a START signal is placed on the I²C bus, all devices on the bus will receive this signal and be notified of the imminent arrival of data on the bus. The first seven bits of the data will be the slave address with the first bit being the MSB. If the address of the slave device matches that of the transmitted address, the HAAS bit in the SIMC1 register will be set and an SIM I²C interrupt will be generated. After entering the interrupt service routine, the slave device must first check the condition of the HAAS and SIMTOF bits to determine whether the interrupt source originates from an address match or from the completionof an 8-bit data transfer completion or I²C bus time-out occurrence. During a data transfer, note that after the 7-bit slave address has been transmitted, the following bit, which is the 8th bit, is the read/write bit whose value will be placed in the SRW bit. This bit will be checked by the slave device to determine whether to go into transmit or receive mode. Before any transfer of data to or from the I²C bus, the microcontroller must initialise the bus, the following are steps to achieve this:

- Step 1

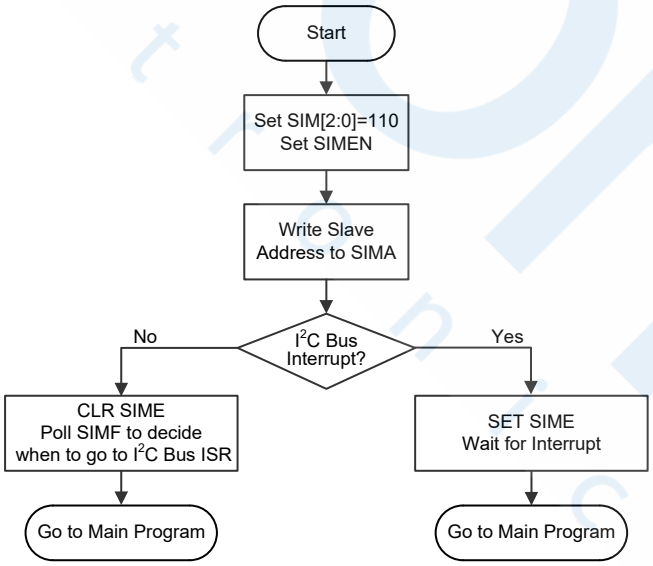
  Set the SIM2~SIM0 and SIMEN bits in the SIMC0 register to "110" and "1" respectively to enable the I²C bus.

- Step 2

  Write the slave address of the device to the I²C bus address register SIMA.

- Step 3

  Set the SIME interrupt enable bit of the interrupt control register to enable the SIM interrupt.



I²C Bus Initialisation Flow Chart

I²C Bus Start Signal

The START signal can only be generated by the master device connected to the I²C bus and not by the slave device. This START signal will be detected by all devices connected to the I²C bus. When detected, this indicates that the I²C bus is busy and therefore the HBB bit will be set. A START condition occurs when a high to low transition on the SDA line takes place when the SCL line remains high.

I²C Slave Address

The transmission of a START signal by the master will be detected by all devices on the I²C bus. To determine which slave device the master wishes to communicate with, the address of the slave device will be sent out immediately following the START signal. All slave devices, after receiving this 7-bit address data, will compare it with their own 7-bit slave address. If the address sent out by the master matches the internal address of the microcontroller slave device, then an internal SIM I²C bus interrupt signal will be generated. The next bit following the address, which is the 8th bit, defines the read/write status and will be saved to the SRW bit of the SIMC1 register. The slave device will then transmit an acknowledge bit, which is a low level, as the 9th bit. The slave device will also set the status flag HAAS when the addresses match.

As an SIM I²C bus interrupt signal can come from three sources, when the program enters the interrupt subroutine, the HAAS and SIMTOF bits should be examined to see whether the interrupt source has come from a matching slave address, the completion of a data byte transfer or from the I²C bus time-out occurrence. When a slave address is matched, the device must be placed in either the transmit mode and then write data to the SIMD register, or in the receive mode where it must implement a dummy read from the SIMD register to release the SCL line.

I²C Bus Read/Write Signal

The SRW bit in the SIMC1 register defines whether the master device wishes to read data from the I²C bus or write data to the I²C bus. The slave device should examine this bit to determine if it is to be a transmitter or a receiver. If the SRW flag is “1” then this indicates that the master device wishes to read data from the I²C bus, therefore the slave device must be setup to send data to the I²C bus as a transmitter. If the SRW flag is “0” then this indicates that the master wishes to send data to the I²C bus, therefore the slave device must be setup to read data from the I²C bus as a receiver.

I²C Bus Slave Address Acknowledge Signal

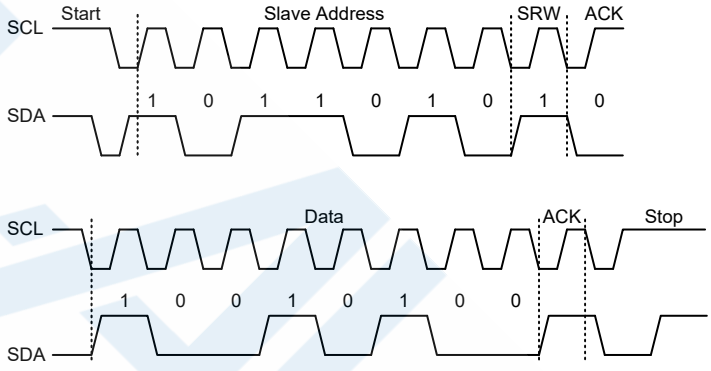
After the master has transmitted a calling address, any slave device on the I²C bus, whose own internal address matches the calling address, must generate an acknowledge signal. The acknowledge signal will inform the master that a slave device has accepted its calling address. If no acknowledge signal is received by the master then a STOP signal must be transmitted by the master to end the communication. When the HAAS flag is high, the addresses have matched and the slave device must check the SRW flag to determine if it is to be a transmitter or a receiver. If the SRW flag is high, the slave device should be setup to be a transmitter so the HTX bit in the SIMC1 register should be set to “1”. If the SRW flag is low, then the microcontroller slave device should be setup as a receiver and the HTX bit in the SIMC1 register should be set to “0”.

I²C Bus Data and Acknowledge Signal

The transmitted data is 8-bits wide and is transmitted after the slave device has acknowledged receipt of its slave address. The order of serial bit transmission is the MSB first and the LSB last. After receipt of 8-bits of data, the receiver must transmit an acknowledge signal, level “0”, before it can receive the next data byte. If the slave transmitter does not receive an acknowledge bit signal

from the master receiver, then the slave transmitter will release the SDA line to allow the master to send a STOP signal to release the I²C Bus. The corresponding data will be stored in the SIMD register. If setup as a transmitter, the slave device must first write the data to be transmitted into the SIMD register. If setup as a receiver, the slave device must read the transmitted data from the SIMD register.

When the slave receiver receives the data byte, it must generate an acknowledge bit, known as TXAK, on the 9th clock. The slave device, which is setup as a transmitter will check the RXAK bit in the SIMC1 register to determine if it is to send another data byte, if not then it will release the SDA line and await the receipt of a STOP signal from the master.

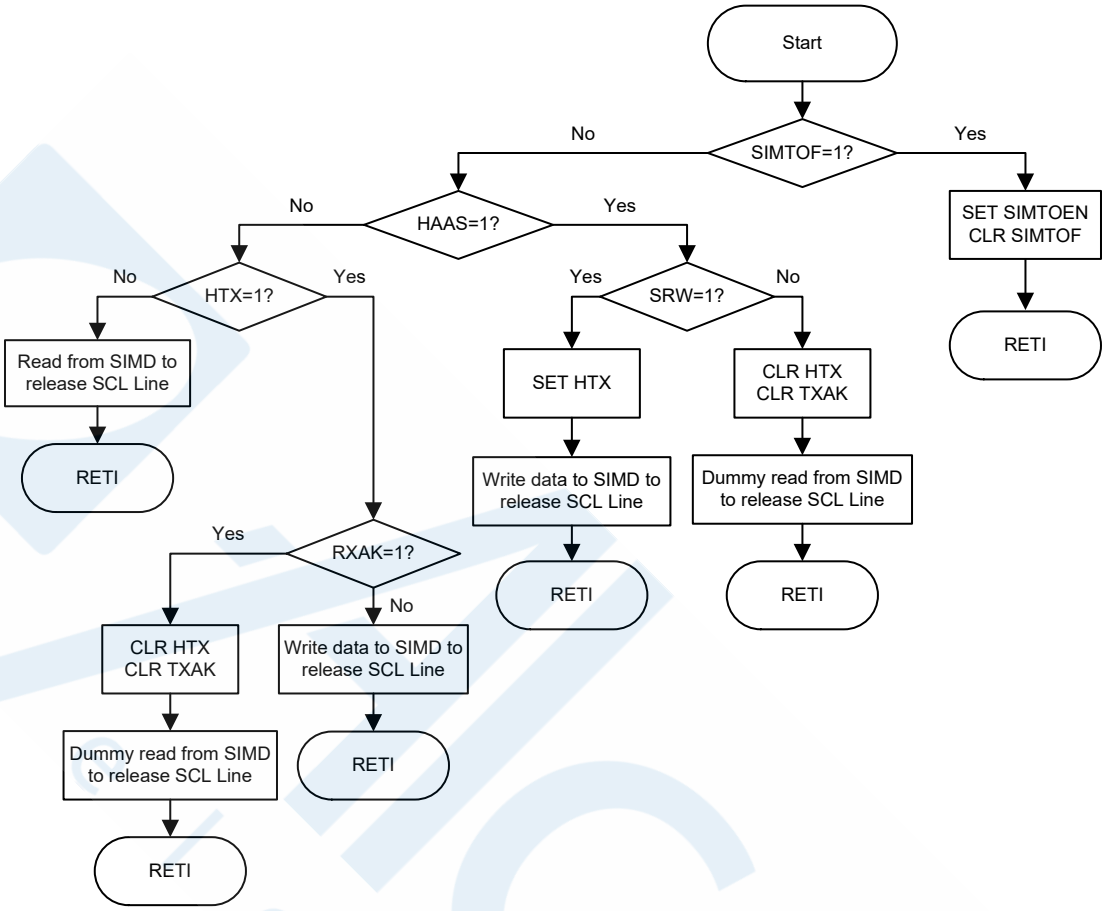


S=Start (1 bit)
SA=Slave Address (7 bits)
SR=SRW bit (1 bit)
M=Slave device send acknowledge bit (1 bit)
D=Data (8 bits)
A=ACK (RXAK bit for transmitter, TXAK bit for receiver, 1 bit)
P=Stop (1 bit)

S	SA	SR	M	D	A	D	A	·····	S	SA	SR	M	D	A	D	A	·····	P

I²C Communication Timing Diagram

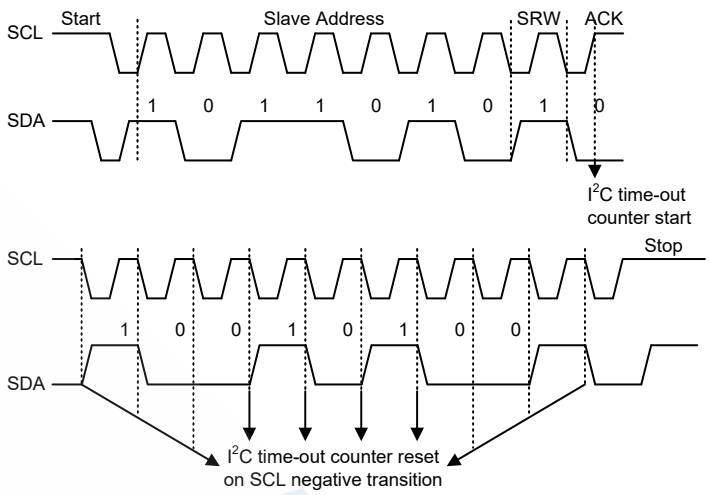
Note: When a slave address is matched, the device must be placed in either the transmit mode and then write data to the SIMD register, or in the receive mode where it must implement a dummy read from the SIMD register to release the SCL line.



I²C Bus ISR Flow Chart

I²C Time-out Control

In order to reduce the problem of I²C lockup due to reception of erroneous clock sources, a time-out function is provided. If the clock source to the I²C bus is not received for a while, then the I²C circuitry and registers will be reset after a certain time-out period. The time-out counter starts counting on an I²C bus "START" & "address match" condition, and is cleared by an SCL falling edge. Before the next SCL falling edge arrives, if the time elapsed is greater than the time-out setup by the SIMTOC register, then a time-out condition will occur. The time-out function will stop when an I²C "STOP" condition occurs.



I²C Time-out

When an I²C time-out counter overflow occurs, the counter will stop and the SIMTOEN bit will be cleared to zero and the SIMTOF bit will be set high to indicate that a time-out condition has occurred. The time-out condition will also generate an interrupt which uses the SIM interrupt vector. When an I²C time-out occurs, the I²C internal circuitry will be reset and the registers will be reset into the following condition:

Registers	After I²C Time-out
SIMD, SIMA, SIMC0	No change
SIMC1	Reset to POR condition

I²C Register after Time-out

The SIMTOF flag can be cleared by the application program. There are 64 time-out period selections which can be selected using the SIMTOS bit field in the SIMTOC register. The time-out duration is calculated by the formula: ((1~64)×32)/f_{SUB}. This gives a time-out period which ranges from about 1ms to 64ms.

- **SIMTOC Register**

Bit	7	6	5	4	3	2	1	0
Name	SIMTOEN	SIMTOF	SIMTOS5	SIMTOS4	SIMTOS3	SIMTOS2	SIMTOS1	SIMTOS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **SIMTOEN**: SIM I²C time-out function control
0: Disable
1: Enable

Bit 6 **SIMTOF**: SIM I²C time-out flag
0: No time-out occurred
1: Time-out occurred
This bit is set high when time-out occurs and can only be cleared to zero by application program.

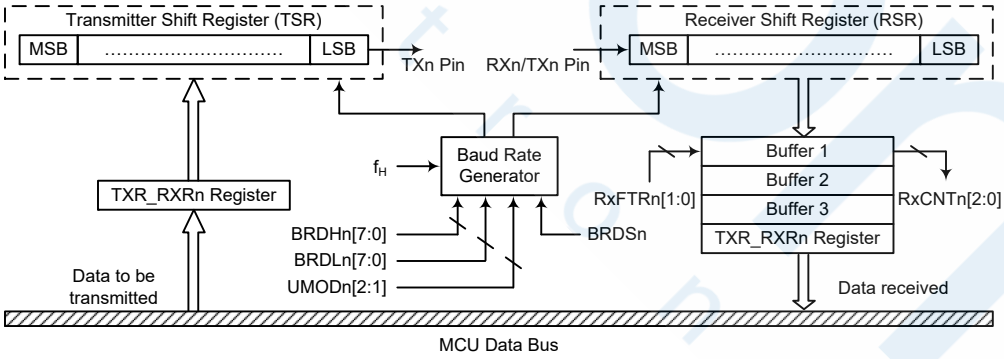
Bit 5~0 **SIMTOS5~SIMTOS0**: SIM I²C time-out period selection
I²C time-out clock source is f_{SUB}/32.
I²C time-out time is equal to (SIMTOS[5:0]+1)×(32/f_{SUB}).

UART Interfaces

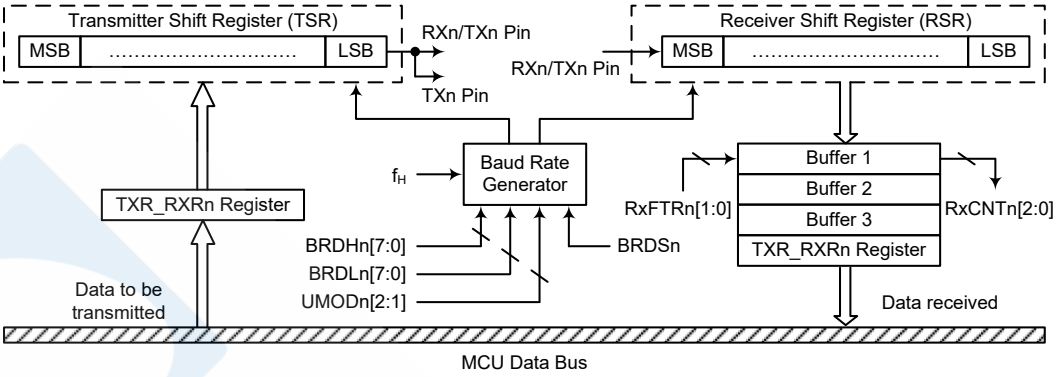
The device contains two integrated full-duplex or half-duplex asynchronous serial communications UART interfaces that enable communication with external devices that contain a serial interface. Note that the UART1 only supports the single wire mode.Each UART function has many features and can transmit and receive data serially by transferring a frame of data with eight or nine data bits per transmission as well as being able to detect errors when the data is overwritten or incorrectly framed. Each UART function possesses its own internal interrupt which can be used to indicate when a reception occurs or when a transmission terminates.

The integrated UARTn function contains the following features:

- Full-duplex or half-duplex (single wire mode), asynchronous communication
- 8 or 9 bits character length
- Even, odd, mark, space or no parity options
- One or two stop bits
- Baud rate generator with 16-bit prescaler
- Parity, framing, noise and overrun error detection
- Support for interrupt on address detect (last character bit=1)
- Separately enabled transmitter and receiver
- 4-byte Deep FIFO Receive Data Buffer
- 1-byte Deep FIFO Transmit Data Buffer
- RXn/TXn pin wake-up function
- Transmit and receive interrupts
- Interrupts can be triggered by the following conditions:
  - Transmitter Empty
  - Transmitter Idle
  - Receiver reaching FIFO trigger level
  - Receiver Overrun
  - Address Mode Detect



UARTn Data Transfer Block Diagram – SWMn=0 (n=0~1)



Note that the UART1 has not the independent TX1 pin.

UARTn Data Transfer Block Diagram – SWMn=1 (n=0~1)

UART External Pins

To communicate with an external serial interface, the internal UARTn has two external pins known as TXn and RXn/TXn, which are pin-shared with I/O or other pin functions. The TXn and RXn/TXn pin function should first be selected by the pin-shared function selection register before the UARTn function is used. Along with the UARTENn bit, the TXENn and RXENn bits, if set, will setup these pins to transmitter output and receiver input conditions. At this time the internal pull-high resistor related to the transmitter output pin will be disabled, while the internal pull-high resistor related to the receiver input pin is controlled by the corresponding I/O pull-high function control bit. When the TXn or RXn/TXn pin function is disabled by clearing the UARTENn, TXENn or RXENn bit, the TXn or RXn/TXn pin will be set to a floating state. At this time whether the internal pull-high resistor is connected to the TXn or RXn/TXn pin or not is determined by the corresponding I/O pull-high function control bit.

UART Single Wire Mode

The UARTn function also supports the Single Wire Mode communication which is selected using the SWMn bit in the UnCR3 register. When the SWMn bit is set high, the UARTn function will be in the single wire mode. In the single wire mode, a single RXn/TXn pin can be used to transmit and receive data depending upon the corresponding control bits. When the RXENn bit is set high, the RXn/TXn pin is used as a receiver pin. When the RXENn bit is cleared to zero and the TXENn bit is set high, the RXn/TXn pin will act as a transmitter pin.

It is recommended not to set both the RXENn and TXENn bits high in the single wire mode. If both the RXENn and TXENn bits are set high, the RXENn bit will have the priority and the UARTn will act as a receiver.

It is important to note that the functional description in this UART Interfaces chapter, which is described from the full-duplex communication standpoint, also applies to the half-duplex (single wire mode) communication except the pin usage. In the single wire mode, the TXn pin mentioned in this chapter should be replaced by the RXn/TXn pin to understand the whole UARTn single wire mode function.

In the single wire mode, the data can also be transmitted on the TXn pin in a transmission operation with proper software configurations. Therefore, the data will be output on the RXn/TXn and TXn pins.

UART Data Transfer Scheme

The UARTn Data Transfer Block Diagram shows the overall data transfer structure arrangement for the UARTn interface. The actual data to be transmitted from the MCU is first transferred to the

TXR_RXRn register by the application program. The data will then be transferred to the Transmit Shift Register from where it will be shifted out, LSB first, onto the TXn pin at a rate controlled by the Baud Rate Generator. Only the TXR_RXRn register is mapped onto the MCU Data Memory, the Transmit Shift Register is not mapped and is therefore inaccessible to the application program.

Data to be received by the UARTn is accepted on the external RXn/TXn pin, from where it is shifted in, LSB first, to the Receiver Shift Register at a rate controlled by the Baud Rate Generator. When the shift register is full, the data will then be transferred from the shift register to the internal TXR_RXRn register, where it is buffered and can be manipulated by the application program. Only the TXR_RXRn register is mapped onto the MCU Data Memory, the Receiver Shift Register is not mapped and is therefore inaccessible to the application program.

It should be noted that the actual register for data transmission and reception only exists as a single shared register in the Data Memory. This shared register known as the TXR_RXRn register is used for both data transmission and data reception.

UART Status and Control Registers

There are nine control registers associated with the UARTn function. The SWMn bit in the UnCR3 register is used to enable/disable the UARTn Single Wire Mode. The UnSR, UnCR1, UnCR2, UFCRn and RxCNTn registers control the overall function of the UARTn, while the BRDHn and BRDLn registers control the Baud rate. The actual data to be transmitted and received on the serial interface is managed through the TXR_RXRn data register.

Register Name	Bit							
	7	6	5	4	3	2	1	0
UnSR	PERRn	NFn	FERRn	OERRn	RIDLEn	RXIFn	TIDLEn	TXIFn
UnCR1	UARTENn	BNOn	PRENn	PRTn1	PRTn0	TXBRKn	RX8n	TX8n
UnCR2	TXENn	RXENn	STOPSn	ADDENn	WAKEn	RIEn	TIIEn	TEIEn
UnCR3	—	—	—	—	—	—	—	SWMn
TXR_RXRn	D7	D6	D5	D4	D3	D2	D1	D0
BRDHn	D7	D6	D5	D4	D3	D2	D1	D0
BRDLn	D7	D6	D5	D4	D3	D2	D1	D0
UFCRn	—	—	UMODn2	UMODn1	UMODn0	BRDSn	RxFTRn1	RxFTRn0
RxCNTn	—	—	—	—	—	D2	D1	D0

UARTn Register List (n=0~1)

• UnSR Register

The UnSR register is the status register for the UARTn, which can be read by the program to determine the present status of the UARTn. All flags within the UnSR register are read only. Further explanation on each of the flags is given below:

Bit	7	6	5	4	3	2	1	0
Name	PERRn	NFn	FERRn	OERRn	RIDLEn	RXIFn	TIDLEn	TXIFn
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	1	0	1	1

Bit 7 **PERRn**: Parity error flag
0: No parity error is detected
1: Parity error is detected
The PERRn flag is the parity error flag. When this read only flag is “0”, it indicates a parity error has not been detected. When the flag is “1”, it indicates that the parity of the received word is incorrect. This error flag is applicable only if the parity is enabled and the parity type (odd, even, mark or space) is selected. The flag can also be cleared

to zero by a software sequence which involves a read to the status register UnSR followed by an access to the TXR_RXRn data register.

Bit 6 **NFn**: Noise flag
0: No noise is detected
1: Noise is detected

The NFn flag is the noise flag. When this read only flag is "0", it indicates no noise condition. When the flag is "1", it indicates that the UARTn has detected noise on the receiver input. The NFn flag is set during the same cycle as the RXIFn flag but will not be set in the case of as overrun. The NFn flag can be cleared to zero by a software sequence which will involve a read to the status register UnSR followed by an access to the TXR_RXRn data register.

Bit 5 **FERRn**: Framing error flag
0: No framing error is detected
1: Framing error is detected

The FERRn flag is the framing error flag. When this read only flag is "0", it indicates that there is no framing error. When the flag is "1", it indicates that a framing error has been detected for the current character. The flag can also be cleared to zero by a software sequence which will involve a read to the status register UnSR followed by an access to the TXR_RXRn data register.

Bit 4 **OERRn**: Overrun error flag
0: No overrun error is detected
1: Overrun error is detected

The OERRn flag is the overrun error flag which indicates when the receiver buffer has overflowed. When this read only flag is "0", it indicates that there is no overrun error. When the flag is "1", it indicates that an overrun error occurs which will inhibit further transfers to the TXR_RXRn receive data register. The flag is cleared to zero by a software sequence, which is a read to the status register UnSR followed by an access to the TXR_RXRn data register.

Bit 3 **RIDLEn**: Receiver status
0: Data reception is in progress (Data being received)
1: No data reception is in progress (Receiver is idle)

The RIDLEn flag is the receiver status flag. When this read only flag is "0", it indicates that the receiver is between the initial detection of the start bit and the completion of the stop bit. When the flag is "1", it indicates that the receiver is idle. Between the completion of the stop bit and the detection of the next start bit, the RIDLEn bit is "1" indicating that the UARTn receiver is idle and the RXn/TXn pin stays in logic high condition.

Bit 2 **RXIFn**: Receive TXR_RXRn data register status
0: TXR_RXRn data register is empty
1: TXR_RXRn data register has available data and Receiver FIFO trigger level is reached

The RXIFn flag is the receive data register status flag. When this read only flag is "0", it indicates that the TXR_RXRn read data register is empty. When the flag is "1", it indicates that the TXR_RXRn read data register contains new data and Receiver FIFO trigger level is reached. When the contents of the shift register are transferred to the TXR_RXRn register and Receiver FIFO trigger level is reached, an interrupt is generated if RIEn=1 in the UnCR2 register. If one or more errors are detected in the received word, the appropriate receive-related flags NFn, FERRn, and/or PERRn are set within the same clock cycle. The RXIFn flag will eventually be cleared to zero when the UnSR register is read with RXIFn set, followed by a read from the TXR_RXRn register, and if the TXR_RXRn register has no more new data available.

Bit 1 **TIDLEn**: Transmission idle
0: Data transmission is in progress (Data being transmitted)
1: No data transmission is in progress (Transmitter is idle)

The TIDLEn flag is known as the transmission complete flag. When this read only flag is "0", it indicates that a transmission is in progress. This flag will be set high when

the TXIFn flag is "1" and when there is no transmit data or break character being transmitted. When TIDLEn is equal to "1", the TXn pin becomes idle with the pin state in logic high condition. The TIDLEn flag is cleared to zero by reading the UnSR register with TIDLEn set and then writing to the TXR_RXRn register. The flag is not generated when a data character or a break is queued and ready to be sent.

Bit 0 **TXIFn**: Transmit TXR_RXRn data register status
0: Character is not transferred to the transmit shift register
1: Character has transferred to the transmit shift register (TXR_RXRn data register is empty)

The TXIFn flag is the transmit data register empty flag. When this read only flag is "0", it indicates that the character is not transferred to the transmitter shift register. When the flag is "1", it indicates that the transmitter shift register has received a character from the TXR_RXRn data register. The TXIFn flag is cleared to zero by reading the UARTn status register (UnSR) with TXIFn set and then writing to the TXR_RXRn data register. Note that when the TXENn bit is set, the TXIFn flag bit will also be set since the transmit data register is not yet full.

• UnCR1 Register

The UnCR1 register together with the UnCR2 and UnCR3 registers are the three UARTn control registers that are used to set the various options for the UARTn function, such as overall on/off control, parity control, data transfer bit length, single wire mode communication etc. Further explanation on each of the bits is given below:

Bit	7	6	5	4	3	2	1	0
Name	UARTENn	BNOn	PRENn	PRTn1	PRTn0	TXBRKn	RX8n	TX8n
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R	W
POR	0	0	0	0	0	0	x	0

"x": unknown

Bit 7 **UARTENn**: UARTn function enable control
0: Disable UARTn. TXn and RXn/TXn pins are in a floating state
1: Enable UARTn. TXn and RXn/TXn pins can function as UARTn pins

The UARTENn bit is the UARTn enable bit. When this bit is equal to "0", the UARTn will be disabled and the RXn/TXn pin as well as the TXn pin will be set in a floating state. When the bit is equal to "1", the UARTn will be enabled and the TXn and RXn/TXn pins will function as defined by the SWMn mode selection bit together with the TXENn and RXENn enable control bits.

When the UARTn is disabled, it will empty the buffer so any character remaining in the buffer will be discarded. In addition, the value of the baud rate counter will be reset. If the UARTn is disabled, all error and status flags will be reset. Also the TXENn, RXENn, TXBRKn, RXIFn, OERRn, FERRn, PERRn and NFn bits as well as the RxCNTn register will be cleared to zero, while the TIDLEn, TXIFn and RIDLEn bits will be set high. Other control bits in UnCR1, UnCR2, UnCR3, UFCRn, BRDHn and BRDLn registers will remain unaffected. If the UARTn is active and the UARTENn bit is cleared to zero, all pending transmissions and receptions will be terminated and the module will be reset as defined above. When the UARTn is re-enabled, it will restart in the same configuration.

Bit 6 **BNOn**: Number of data transfer bits selection
0: 8-bit data transfer
1: 9-bit data transfer

This bit is used to select the data length format, which can have a choice of either 8-bit or 9-bit format. When this bit is equal to "1", a 9-bit data length format will be selected. If the bit is equal to "0", then an 8-bit data length format will be selected. If 9-bit data length format is selected, then bits RX8n and TX8n will be used to store the 9th bit of the received and transmitted data respectively.

Note that the 9th bit of data if BNOn=1, or the 8th bit of data if BNOn=0, which is used as the parity bit, does not transfer to RX8n or TXRX7 respectively when the parity function is enabled.

Bit 5 **PRENn**: Parity function enable control
0: Parity function is disabled
1: Parity function is enabled
This is the parity enable bit. When this bit is equal to "1", the parity function will be enabled. If the bit is equal to "0", then the parity function will be disabled.

Bit 4~3 **PRTn1~PRTn0**: Parity type selection bits
00: Even parity for parity generator
01: Odd parity for parity generator
10: Mark parity for parity generator
11: Space parity for parity generator
These bits are the parity type selection bits. When these bits are equal to 00b, even parity type will be selected. If these bits are equal to 01b, then odd parity type will be selected. If these bits are equal to 10b, then a 1 (Mark) in the parity bit location will be selected. If these bits are equal to 11b, then a 0 (Space) in the parity bit location will be selected.

Bit 2 **TXBRKn**: Transmit break character
0: No break character is transmitted
1: Break characters transmit
The TXBRKn bit is the Transmit Break Character bit. When this bit is "0", there are no break characters and the TXn pin operates normally. When the bit is "1", there are transmit break characters and the transmitter will send logic zeros. When this bit is equal to "1", after the buffered data has been transmitted, the transmitter output is held low for a minimum of a 13-bit length and until the TXBRKn bit is reset.

Bit 1 **RX8n**: Receive data bit 8 for 9-bit data transfer format (read only)
This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the received data known as RX8n. The BNOn bit is used to determine whether data transfers are in 8-bit or 9-bit format.

Bit 0 **TX8n**: Transmit data bit 8 for 9-bit data transfer format (write only)
This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the transmitted data known as TX8n. The BNOn bit is used to determine whether data transfers are in 8-bit or 9-bit format.

• UnCR2 Register

The UnCR2 register is the second of the UARTn control registers and serves several purposes. One of its main functions is to control the basic enable/disable operation of the UARTn Transmitter and Receiver as well as enabling the various UARTn interrupts sources. The register also serves to control the STOP bit number selection, receiver wake-up enable and the address detect enable. Further explanation on each of the bits is given below:

Bit	7	6	5	4	3	2	1	0
Name	TXENn	RXENn	STOPSn	ADDENn	WAKEn	RIEn	TIIEn	TEIEn
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **TXENn**: UARTn transmitter enabled control
0: UARTn transmitter is disabled
1: UARTn transmitter is enabled
The bit named TXENn is the Transmitter Enable Bit. When this bit is equal to "0", the transmitter will be disabled with any pending data transmissions being aborted. In addition the buffers will be reset. In this situation the TXn pin will be set in a floating state.
If the TXENn bit is equal to "1" and the UARTENn bit are also equal to "1", the transmitter will be enabled and the TXn pin will be controlled by the UARTn. Clearing the

TXENn bit during a transmission will cause the data transmission to be aborted and will reset the transmitter. If this situation occurs, the TXn pin will be set in a floating state.

Bit 6 **RXENn**: UARTn Receiver enabled control
0: UARTn receiver is disabled
1: UARTn receiver is enabled

The bit named RXENn is the Receiver Enable Bit. When this bit is equal to "0", the receiver will be disabled with any pending data receptions being aborted. In addition the receive buffers will be reset. In this situation the RXn/TXn pin will be set in a floating state. If the RXENn bit is equal to "1" and the UARTENn bit is also equal to "1", the receiver will be enabled and the RXn/TXn pin will be controlled by the UARTn. Clearing the RXENn bit during a reception will cause the data reception to be aborted and will reset the receiver. If this situation occurs, the RXn/TXn pin will be set in a floating state.

Bit 5 **STOPSn**: Number of stop bits selection for transmitter
0: One stop bit format is used
1: Two stop bits format is used

This bit determines if one or two stop bits are to be used for transmitter. When this bit is equal to "1", two stop bits are used. If this bit is equal to "0", then only one stop bit is used.

Bit 4 **ADDENn**: Address detect function enable control
0: Address detect function is disabled
1: Address detect function is enabled

The bit named ADDENn is the address detect function enable control bit. When this bit is equal to "1", the address detect function is enabled. When it occurs, if the 8th bit, which corresponds to TXRX7 if BNOn=0 or the 9th bit, which corresponds to RX8n if BNOn=1, has a value of "1", then the received word will be identified as an address, rather than data. If the corresponding interrupt is enabled, an interrupt request will be generated each time the received word has the address bit set, which is the 8th or 9th bit depending on the value of BNOn. If the address bit known as the 8th or 9th bit of the received word is "0" with the address detect function being enabled, an interrupt will not be generated and the received data will be discarded.

Bit 3 **WAKEn**: RXn/TXn pin wake-up UARTn function enable control
0: RXn/TXn pin wake-up UARTn function is disabled
1: RXn/TXn pin wake-up UARTn function is enabled

This bit is used to control the wake-up UARTn function when a falling edge on the RXn/TXn pin occurs. Note that this bit is only available when the UARTn clock (f_H) is switched off. There will be no RXn/TXn pin wake-up UARTn function if the UARTn clock (f_H) exists. If the WAKEn bit is set to 1 as the UARTn clock (f_H) is switched off, a UARTn wake-up request will be initiated when a falling edge on the RXn/TXn pin occurs. When this request happens and the corresponding interrupt is enabled, an RXn/TXn pin wake-up UARTn interrupt will be generated to inform the MCU to wake up the UARTn function by switching on the UARTn clock (f_H) via the application program. Otherwise, the UARTn function cannot resume even if there is a falling edge on the RXn/TXn pin when the WAKEn bit is cleared to 0.

Bit 2 **RIEn**: Receiver interrupt enable control
0: Receiver related interrupt is disabled
1: Receiver related interrupt is enabled

This bit enables or disables the receiver interrupt. If this bit is equal to "1" and when the receiver overrun flag OERRn or receive data available flag RXIFn is set, the UARTn interrupt request flag will be set. If this bit is equal to "0", the UARTn interrupt request flag will not be influenced by the condition of the OERRn or RXIFn flags.

Bit 1 **TIIEn**: Transmitter Idle interrupt enable control
0: Transmitter idle interrupt is disabled
1: Transmitter idle interrupt is enabled

This bit enables or disables the transmitter idle interrupt. If this bit is equal to "1" and when the transmitter idle flag TIDLEn is set, due to a transmitter idle condition, the UARTn interrupt request flag will be set. If this bit is equal to "0", the UARTn interrupt request flag will not be influenced by the condition of the TIDLEn flag.

Bit 0 **TEIEn**: Transmitter Empty interrupt enable control
0: Transmitter empty interrupt is disabled
1: Transmitter empty interrupt is enabled
This bit enables or disables the transmitter empty interrupt. If this bit is equal to "1" and when the transmitter empty flag TXIFn is set, due to a transmitter empty condition, the UARTn interrupt request flag will be set. If this bit is equal to "0", the UARTn interrupt request flag will not be influenced by the condition of the TXIFn flag.

• UnCR3 Register

The UnCR3 register is used to enable the UARTn Single Wire Mode communication. As the name suggests in the single wire mode the UARTn communication can be implemented in one single line, RXn/TXn, together with the control of the RXENn and TXENn bits in the UnCR2 register.

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	—	SWMn
R/W	—	—	—	—	—	—	—	R/W
POR	—	—	—	—	—	—	—	0

Bit 7~1 Unimplemented, read as "0"

Bit 0 **SWMn**: Single Wire Mode enable control
0: Disable, the RXn/TXn pin is used as UARTn receiver function only
1: Enable, the RXn/TXn pin can be used as UARTn receiver or transmitter function controlled by the RXENn and TXENn bits
Note that when the Single Wire Mode is enabled, if both the RXENn and TXENn bits are high, the RXn/TXn pin will just be used as UARTn receiver input. And the UART1 only supports the single wire mode.

• TXR_RXRn Register

The TXR_RXRn register is the data register which is used to store the data to be transmitted on the TXn pin or being received from the RXn/TXn pin.

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

"x": unknown

Bit 7~0 **D7~D0**: UARTn Transmit/Receive Data bit 7 ~ bit 0

• BRDHn Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: Baud Rate divider high byte
The baud rate divider BRDn (BRDHn/BRDLn) defines the UARTn clock divider ratio.
Baud Rate=f_H/(BRDn+UMODn/8)
BRDn=16~65535 or 8~65535 depending on BRDSn
Note: 1. BRDn value should not be set to less than 16 when BRDSn=0 or less than 8 when BRDSn=1, otherwise errors may occur.

2. The BRDLn must be written first and then BRDHn, otherwise errors may occur.
3. The BRDHn register should not be modified during data transmission process.

• **BRDLn Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: Baud Rate divider low byte

The baud rate divider BRDn (BRDHn/BRDLn) defines the UARTn clock divider ratio.

Baud Rate=f_H/(BRDn+UMODn/8)

BRDn=16~65535 or 8~65535 depending on BRDSn

Note: 1. BRDn value should not be set to less than 16 when BRDSn=0 or less than 8 when BRDSn=1, otherwise errors may occur.
2. The BRDLn must be written first and then BRDHn, otherwise errors may occur.
3. The BRDLn register should not be modified during data transmission process.

• **UFCRn Register**

The UFCRn register is the FIFO control register which is used for UARTn modulation control, BRDn range selection and trigger level selection for RXIFn and interrupt.

Bit	7	6	5	4	3	2	1	0
Name	—	—	UMODn2	UMODn1	UMODn0	BRDSn	RxFTRn1	RxFTRn0
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as "0"

Bit 5~3 **UMODn2~UMODn0**: UARTn Modulation Control bits

The modulation control bits are used to correct the baud rate of the received or transmitted UARTn signal. These bits determine if the extra UARTn clock cycle should be added in a UARTn bit time. The UMODn2~UMODn0 will be added to internal accumulator for every UARTn bit time. Until a carry to bit 3, the corresponding UARTn bit time increases a UARTn clock cycle.

Bit 2 **BRDSn**: BRDn range selection

0: BRDn range is from 16 to 65535
1: BRDn range is from 8 to 65535

The BRDSn is used to control the sampling point in a UARTn bit time. If the BRDSn is cleared to zero, the sampling point will be BRDn/2, BRDn/2+1×f_H, and BRDn/2+2×f_H in a UARTn bit time. If the BRDSn is set high, the sampling point will be BRDn/2-1×f_H, BRDn/2, and BRDn/2+2×f_H in a UARTn bit time.

Note that the BRDSn bit should not be modified during data transmission process.

Bit 1~0 **RxFTRn1~RxFTRn0**: Receiver FIFO trigger level (bytes)

00: 4 bytes in Receiver FIFO
01: 1 or more bytes in Receiver FIFO
10: 2 or more bytes in Receiver FIFO
11: 3 or more bytes in Receiver FIFO

For the receiver these bits define the number of received data bytes in the Receiver FIFO that will trigger the RXIFn bit being set high, an interrupt will also be generated if the RIEn bit is enabled. After the reset the receiver FIFO is empty.

• **RxCNTn Register**

The RxCNTn register is the counter used to indicate the number of received data bytes in the Receiver FIFO which have not been read by the MCU. This register is read only.

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	D2	D1	D0
R/W	—	—	—	—	—	R	R	R
POR	—	—	—	—	—	0	0	0

Bit 7~3 Unimplemented, read as "0"

Bit 2~0 **D2~D0**: Receiver FIFO counter

The RxCNTn register is the counter used to indicate the number of receiver data bytes in Receiver FIFO which is not read by MCU. When Receiver FIFO receives one byte data, the RxCNTn will increase by one; when the MCU reads one byte data from Receiver FIFO, the RxCNTn will decrease by one. If there are 4 bytes of data in the Receiver FIFO, the 5th data will be saved in the shift register. If there is 6th data, the 6th data will be saved in the shift register. But the RxCNTn remains the value of 4. The RxCNTn will be cleared when reset occurs or UARTENn=1. This register is read only.

Baud Rate Generator

To setup the speed of the serial data communication, the UARTn function contains its own dedicated baud rate generator. The baud rate is controlled by its own internal free running 16-bit timer, the period of which is determined by two factors. The first of these is the value placed in BRDHn/BRDLn register and the second is the UARTn modulation control bits (UMODn2~UMODn0). If a baud rate BR is required with UARTn clock f_H.

$$f_H/BR=\text{Integer Part}+\text{Fractional Part}$$

The integer part is loaded into BRDn (BRDHn/BRDLn). The fractional part is multiplied by 8 and rounded, then loaded into UMODn bit field as following:

$$BRDn=TRUNC\ (f_H/BR)$$

$$UMODn=ROUND\ [MOD\ (f_H/BR)\times 8]$$

Therefore, the actual baud rate is as following:

$$\text{Baud rate}=f_H/[BRDn+(UMODn/8)]$$

Calculating the Baud Rate and Error Values

For a clock frequency of 4MHz, determine the BRDHn/BRDLn register value, the actual baud rate and the error value for a desired baud rate of 230400.

From the above formula, the $BRDn=TRUNC\ (f_H/BR)=TRUNC(17.36111)=17$

The $UMODn=ROUND[MOD(f_H/BR)\times 8]=ROUND(0.36111\times 8)=ROUND(2.88888)=3$

The actual Baud Rate$=f_H/[BRDn+(UMODn/8)]=230215.83$

Therefore the error is equal to (230215.83-230400)/230400=-0.08%

Modulation Control Example

To get the best-fitting bit sequence for UARTn modulation control bits UMODn2~UMODn0, the following algorithm can be used: Firstly, the fractional part of the theoretical division factor is multiplied by 8. Then the product will be rounded and UMODn2~UMODn0 bits will be filled with the rounded value. The UMODn2~UMODn0 bits will be added to internal accumulator for every UARTn bit time. Until a carry to bit 3, the corresponding UARTn bit time increases a UARTn clock cycle. The following is an example using the fraction 0.36111 previously calculated: UMODn[2:0]= ROUND(0.36111×8)=011b.

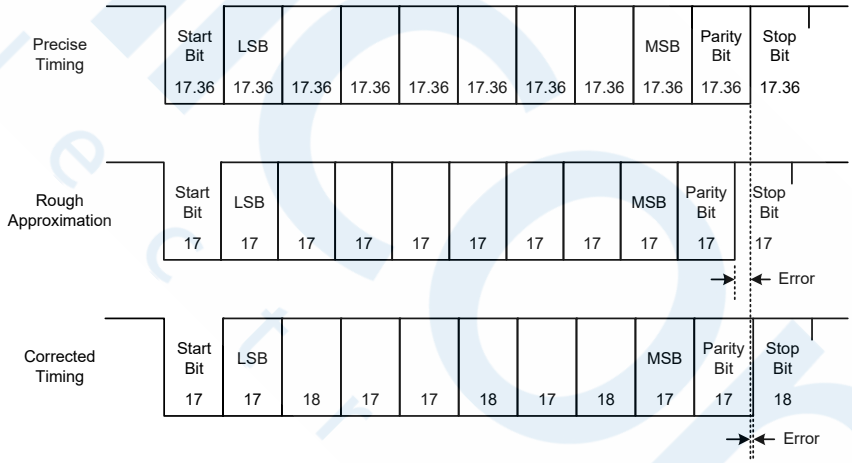
Fraction Addition	Carry to Bit 3	UARTn Bit Time Sequence	Extra UARTn Clock Cycle
0000b+0011b=0011b	No	Start bit	No
0011b+0011b=0110b	No	D0	No
0110b+0011b=1001b	Yes	D1	Yes
1001b+0011b=1100b	No	D2	No
1100b+0011b=1111b	No	D3	No
1111b+0011b=0010b	Yes	D4	Yes
0010b+0011b=0101b	No	D5	No
0101b+0011b=1000b	Yes	D6	Yes
1000b+0011b=1011b	No	D7	No
1011b+0011b=1110b	No	Parity bit	No
1110b+0011b=0001b	Yes	Stop bit	Yes

Baud Rate Correction Example

The following figure presents an example using a baud rate of 230400 generated with UARTn clock f_H. The data format for the following figure is: eight data bits, parity enabled, no address bit, two stop bits.

The following figure shows three different frames:

- The upper frame is the correct one, with a bit-length of 17.36 f_H cycles (4000000/230400=17.36).
- The middle frame uses a rough estimate, with 17 f_H cycles for the bit length.
- The lower frame shows a corrected frame using the best fit for the UARTn modulation control bits UMODn2~UMODn0.



UART Setup and Control

For data transfer, the UARTn function utilizes a non-return-to-zero, more commonly known as NRZ, format. This is composed of one start bit, eight or nine data bits, and one or two stop bits. Parity is supported by the UARTn hardware, and can be setup to be even, odd, mark, space or no parity. For the most common data format, 8 data bits along with no parity and one stop bit, denoted as 8, N, 1, is used as the default setting, which is the setting at power-on. The number of data bits and stop bits, along with the parity, are setup by programming the corresponding BNOn, PRTn1~PRTn0, PRENn, and STOPSn bits. The baud rate used to transmit and receive data is setup using the internal 16-bit baud rate generator, while the data is transmitted and received LSB first. Although the UARTn transmitter and receiver are functionally independent, they both use the same data format and baud rate. In all cases stop bits will be used for data transmission.

Enabling/Disabling the UARTn Interface

The basic on/off function of the internal UARTn function is controlled using the UARTENn bit in the UnCR1 register. If the UARTENn, TXENn and RXENn bits are set, then these two UARTn pins will act as normal TXn output pin and RXn/TXn input pin respectively. If no data is being transmitted on the TXn pin, then it will default to a logic high value.

Clearing the UARTENn bit will disable the TXn and RXn/TXn pins and allow these two pins to be used as normal I/O or other pin-shared functional pins by configuring the corresponding pin-shared control bits. When the UARTn function is disabled the buffer will be reset to an empty condition, at the same time discarding any remaining residual data. Disabling the UARTn will also reset the error and status flags with bits TXENn, RXENn, TXBRKn, RXIFn, OERRn, FERRn, PERRn and NFn as well as register RxCNTn being cleared while bits TIDLEn, TXIFn and RIDLEn will be set. The remaining control bits in the UnCR1, UnCR2, UnCR3, UFCRn, BRDHn and BRDLn registers will remain unaffected. If the UARTENn bit in the UnCR1 register is cleared while the UARTn is active, then all pending transmissions and receptions will be immediately suspended and the UARTn will be reset to a condition as defined above. If the UARTn is then subsequently re-enabled, it will restart again in the same configuration.

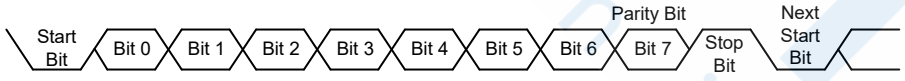
Data, Parity and Stop Bit Selection

The format of the data to be transferred is composed of various factors such as data bit length, parity on/off, parity type, address bits and the number of stop bits. These factors are determined by the setup of various bits within the UnCR1 and UnCR2 registers. The BNOn bit controls the number of data bits which can be set to either 8 or 9, the PRTn1~PRTn0 bits control the choice of odd, even, mark or space parity, the PRENn bit controls the parity on/off function and the STOPSn bit decides whether one or two stop bits are to be used. The following table shows various formats for data transmission. The address bit, which is the MSB of the data byte, identifies the frame as an address character or data if the address detect function is enabled. The number of stop bits, which can be either one or two, is independent of the data length and is only used for the transmitter. There is only one stop bit for the receiver.

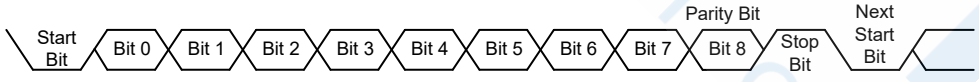
Start Bit	Data Bits	Address Bit	Parity Bit	Stop Bit
Example of 8-bit Data Formats				
1	8	0	0	1
1	7	0	1	1
1	7	1	0	1
Example of 9-bit Data Formats				
1	9	0	0	1
1	8	0	1	1
1	8	1	0	1

Transmitter Receiver Data Format

The following diagram shows the transmit and receive waveforms for both 8-bit and 9-bit data formats.



8-bit data format



9-bit data format

UART Transmitter

Data word lengths of either 8 or 9 bits can be selected by programming the BNOn bit in the UnCR1 register. When BNOn bit is set, the word length will be set to 9 bits. In this case the 9th bit, which is the MSB, needs to be stored in the TX8n bit in the UnCR1 register. At the transmitter core lies the Transmitter Shift Register, more commonly known as the TSRn, whose data is obtained from the transmit data register, which is known as the TXR_RXRn register. The data to be transmitted is loaded into this TXR_RXRn register by the application program. The TSRn register is not written to with new data until the stop bit from the previous transmission has been sent out. As soon as this stop bit has been transmitted, the TSRn can then be loaded with new data from the TXR_RXRn register, if it is available. It should be noted that the TSRn register, unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations. An actual transmission of data will normally be enabled when the TXENn bit is set, but the data will not be transmitted until the TXR_RXRn register has been loaded with data and the baud rate generator has defined a shift clock source. However, the transmission can also be initiated by first loading data into the TXR_RXRn register, after which the TXENn bit can be set. When a transmission of data begins, the TSRn is normally empty, in which case a transfer to the TXR_RXRn register will result in an immediate transfer to the TSRn. If during a transmission the TXENn bit is cleared, the transmission will immediately cease and the transmitter will be reset. The TXn output pin can then be configured as the I/O or other pin-shared function by configuring the corresponding pin-shared control bits.

Transmitting Data

When the UARTn is transmitting data, the data is shifted on the TXn pin from the shift register, with the least significant bit first. In the transmit mode, the TXR_RXRn register forms a buffer between the internal bus and the transmitter shift register. It should be noted that if 9-bit data format has been selected, then the MSB will be taken from the TX8n bit in the UnCR1 register. The steps to initiate a data transfer can be summarized as follows:

- Make the correct selection of the BNOn, PRTn1~PRTn0, PRENn and STOPSn bits to define the required word length, parity type and number of stop bits.
- Setup the BRDHn, BRDLn registers and UMODn2~UMODn0 bits to select the desired baud rate.
- Set the TXENn bit to ensure that the TXn pin is used as a UARTn transmitter pin.
- Access the UnSR register and write the data that is to be transmitted into the TXR_RXRn register. Note that this step will clear the TXIFn bit.

This sequence of events can now be repeated to send additional data. It should be noted that when TXIFn=0, data will be inhibited from being written to the TXR_RXRn register. Clearing the TXIFn flag is always achieved using the following software sequence:

1. A UnSR register access
2. A TXR_RXRn register write execution

The read-only TXIFn flag is set by the UARTn hardware and if set indicates that the TXR_RXRn register is empty and that other data can now be written into the TXR_RXRn register without overwriting the previous data. If the TEIEn bit is set then the TXIFn flag will generate an interrupt.

During a data transmission, a write instruction to the TXR_RXRn register will place the data into the TXR_RXRn register, which will be copied to the shift register at the end of the present transmission. When there is no data transmission in progress, a write instruction to the TXR_RXRn register will place the data directly into the shift register, resulting in the commencement of data transmission, and the TXIFn bit being immediately set. When a frame transmission is complete, which happens

after stop bits are sent or after the break frame, the TIDLEn bit will be set. To clear the TIDLEn bit the following software sequence is used:

1. A UnSR register access
2. A TXR_RXRn register write execution

Note that both the TXIFn and TIDLEn bits are cleared by the same software sequence.

Transmiting Break

If the TXBRKn bit is set high and the state keeps for a time greater than [(BRDn+1)×t_H] while TIDLEn=1, then break characters will be sent on the next transmission. Break character transmission consists of a start bit, followed by 13×N '0' bits and stop bits, where N=1, 2, etc. If a break character is to be transmitted then the TXBRKn bit must be first set by the application program, and then cleared to generate the stop bits. Transmitting a break character will not generate a transmit interrupt. Note that a break condition length is at least 13 bits long. If the TXBRKn bit is continually kept at a logic high level then the transmitter circuitry will transmit continuous break characters. After the application program has cleared the TXBRKn bit, the transmitter will finish transmitting the last break character and subsequently send out one or two stop bits. The automatic logic highs at the end of the last break character will ensure that the start bit of the next frame is recognized.

UART Receiver

The UARTn is capable of receiving word lengths of either 8 or 9 bits. If the BNOn bit is set, the word length will be set to 9 bits with the MSB being stored in the RX8n bit of the UnCR1 register. At the receiver core lies the Receive Serial Shift Register, commonly known as the RSRn. The data which is received on the RXn/TXn external input pin is sent to the data recovery block. The data recovery block operating speed is 16 times that of the baud rate, while the main receive serial shifter operates at the baud rate. After the RXn/TXn pin is sampled for the stop bit, the received data in RSRn is transferred to the receive data register, if the register is empty. The data which is received on the external RXn/TXn input pin is sampled three times by a majority detect circuit to determine the logic level that has been placed onto the RXn/TXn pin. It should be noted that the RSRn register, unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations.

Receiving Data

When the UARTn receiver is receiving data, the data is serially shifted in on the external RXn/TXn input pin, LSB first. In the read mode, the TXR_RXRn register forms a buffer between the internal bus and the receiver shift register. The TXR_RXRn register is a four-byte deep FIFO data buffer, where four bytes can be held in the FIFO while a fifth byte can continue to be received. Note that the application program must ensure that the data is read from TXR_RXRn before the fifth byte has been completely shifted in, otherwise this fifth byte will be discarded and an overrun error OERRn will be subsequently indicated. The steps to initiate a data transfer can be summarized as follows:

- Make the correct selection of BNOn, PRTn1~PRTn0 and PRENn bits to define the word length, parity type.
- Setup the BRDHn, BRDLn registers and the UMODn2~UMODn0 to select the desired baud rate.
- Set the RXENn bit to ensure that the RXn/TXn pin is used as a UARTn receiver pin.

At this point the receiver will be enabled which will begin to look for a start bit.

When a character is received the following sequence of events will occur:

- The RXIFn bit in the UnSR register will be set when the TXR_RXRn register has data available, the number of the available data bytes can be checked by polling the RxCNTn register content.

- When the contents of the shift register have been transferred to the TXR_RXRn register and reach Receiver FIFO trigger level is reached, if the RIEn bit is set, then an interrupt will be generated.
- If during reception, a frame error, noise error, parity error or an overrun error has been detected, then the error flags can be set.

The RXIFn bit can be cleared using the following software sequence:

1. A UnSR register access
2. A TXR_RXRn register read execution

Receiving Break

Any break character received by the UARTn will be managed as a framing error. The receiver will count and expect a certain number of bit times as specified by the values programmed into the BNOn bit plus one stop bit. If the break is much longer than 13 bit times, the reception will be considered as complete after the number of bit times specified by BNOn plus one stop bit. The RXIFn bit is set, FERRn is set, zeros are loaded into the receive data register, interrupts are generated if appropriate and the RIDLEn bit is set. A break is regarded as a character that contains only zeros with the FERRn flag set. If a long break signal has been detected, the receiver will regard it as a data frame including a start bit, data bits and the invalid stop bit and the FERRn flag will be set. The receiver must wait for a valid stop bit before looking for the next start bit. The receiver will not make the assumption that the break condition on the line is the next start bit. The break character will be loaded into the buffer and no further data will be received until stop bits are received. It should be noted that the RIDLEn read only flag will go high when the stop bits have not yet been received. The reception of a break character on the UARTn registers will result in the following:

- The framing error flag, FERRn, will be set.
- The receive data register, TXR_RXRn, will be cleared.
- The OERRn, NFn, PERRn, RIDLEn or RXIFn flags will possibly be set.

Idle Status

When the receiver is reading data, which means it will be in between the detection of a start bit and the reading of a stop bit, the receiver status flag in the UnSR register, otherwise known as the RIDLEn flag, will have a zero value. In between the reception of a stop bit and the detection of the next start bit, the RIDLEn flag will have a high value, which indicates the receiver is in an idle condition.

Receiver Interrupt

The read only receive interrupt flag RXIFn in the UnSR register is set by an edge generated by the receiver. An interrupt is generated if RIEn=1, when a word is transferred from the Receive Shift Register, RSRn, to the Receive Data Register, TXR_RXRn. An overrun error can also generate an interrupt if RIEn=1.

Managing Receiver Errors

Several types of reception errors can occur within the UARTn module, the following section describes the various types and how they are managed by the UARTn.

Overrun Error – OERRn

The TXR_RXRn register is composed of a four-byte deep FIFO data buffer, where four bytes can be held in the FIFO register, while a fifth byte can continue to be received. Before this fifth byte has been entirely shifted in, the data should be read from the TXR_RXRn register. If this is not done, the overrun error flag OERRn will be consequently indicated.

In the event of an overrun error occurring, the following will happen:

- The OERRn flag in the UnSR register will be set.
- The TXR_RXRn contents will not be lost.
- The shift register will be overwritten.
- An interrupt will be generated if the RIEn bit is set.

The OERRn flag can be cleared by an access to the UnSR register followed by a read to the TXR_RXRn register.

Noise Error – NFn

Over-sampling is used for data recovery to identify valid incoming data and noise. If noise is detected within a frame, the following will occur:

- The read only noise flag, NFn, in the UnSR register will be set on the rising edge of the RXIFn bit.
- Data will be transferred from the shift register to the TXR_RXRn register.
- No interrupt will be generated. However this bit rises at the same time as the RXIFn bit which itself generates an interrupt.

Note that the NFn flag is reset by a UnSR register read operation followed by a TXR_RXRn register read operation.

Framing Error – FERRn

The read only framing error flag, FERRn, in the UnSR register, is set if a zero is detected instead of stop bits. If two stop bits are selected, both stop bits must be high; otherwise the FERRn flag will be set. The FERRn flag and the received data will be recorded in the UnSR and TXR_RXRn registers respectively, and the flag is cleared in any reset.

Parity Error – PERRn

The read only parity error flag, PERRn, in the UnSR register, is set if the parity of the received word is incorrect. This error flag is only applicable if the parity is enabled, PRENn=1, and if the parity type, odd or even is selected. The read only PERRn flag and the received data will be recorded in the UnSR and TXR_RXRn registers respectively. It is cleared on any reset, it should be noted that the flags, FERRn and PERRn, in the UnSR register should first be read by the application program before reading the data word.

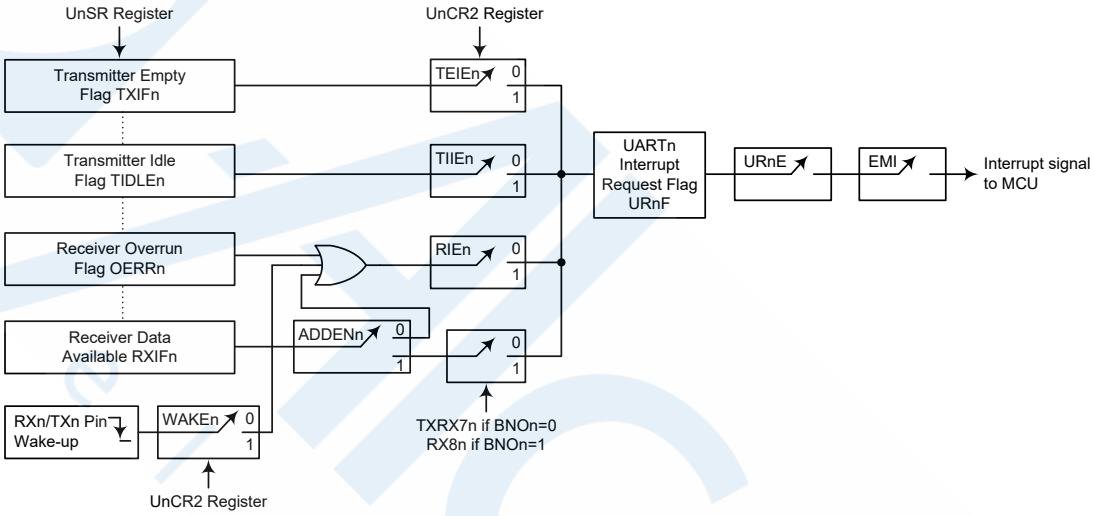
UART Interrupt Structure

Several individual UARTn conditions can generate a UARTn interrupt. When these conditions exist, a low pulse will be generated to get the attention of the microcontroller. These conditions are a transmitter data register empty, transmitter idle, receiver reaching FIFO trigger level, receiver overrun, address detect and an RXn/TXn pin wake-up. When any of these conditions are created, if the global interrupt enable bit and its corresponding interrupt control bit are enabled and the stack is not full, the program will jump to its corresponding interrupt vector where it can be serviced before returning to the main program. Four of these conditions have the corresponding UnSR register flags which will generate a UARTn interrupt if its associated interrupt enable control bit in the UnCR2 register is set. The two transmitter interrupt conditions have their own corresponding enable control bits, while the two receiver interrupt conditions have a shared enable control bit. These enable bits can be used to mask out individual UARTn interrupt sources.

The address detect condition, which is also a UARTn interrupt source, does not have an associated flag, but will generate a UARTn interrupt when an address detect condition occurs if its function is

enabled by setting the ADDENn bit in the UnCR2 register. An RXn/TXn pin wake-up, which is also a UARTn interrupt source, does not have an associated flag, but will generate a UARTn interrupt if the UARTn clock (f_H) source is switched off and the WAKEn and RIEn bits in the UnCR2 register are set when a falling edge on the RXn/TXn pin occurs.

Note that the UnSR register flags are read only and cannot be cleared or set by the application program, neither will they be cleared when the program jumps to the corresponding interrupt servicing routine, as is the case for some of the other interrupts. The flags will be cleared automatically when certain actions are taken by the UARTn, the details of which are given in the UARTn register section. The overall UARTn interrupt can be disabled or enabled by the related interrupt enable control bits in the interrupt control registers of the microcontroller to decide whether the interrupt requested by the UARTn module is masked out or allowed.



UARTn Interrupt Structure (n=0~1)

Address Detect Mode

Setting the Address Detect Mode bit, ADDENn, in the UnCR2 register, enables this special mode. If this bit is enabled then an additional qualifier will be placed on the generation of a Receiver Data Available interrupt, which is requested by the RXIFn flag. If the ADDENn bit is enabled, then when data is available, an interrupt will only be generated if the highest received bit has a high value. Note that the URnE and EMI interrupt enable bits must also be enabled for correct interrupt generation. This highest address bit is the 9th bit if BNOn=1 or the 8th bit if BNOn=0. If this bit is high, then the received word will be defined as an address rather than data. A Data Available interrupt will be generated every time the last bit of the received word is set. If the ADDENn bit is not enabled, then a Receiver Data Available interrupt will be generated each time the RXIFn flag is set, irrespective of the data last bit status. The address detect mode and parity enable are mutually exclusive functions. Therefore if the address detect mode is enabled, then to ensure correct operation, the parity function should be disabled by resetting the parity enable bit PRENn to zero.

ADDENn	9th bit if BNOn=1, 8th bit if BNOn=0	UARTn Interrupt Generated
0	0	√
	1	√
1	0	×
	1	√

ADDENn Bit Function (n=0~1)

UART Power Down and Wake-up

When the UARTn clock, f_H, is switched off, the UARTn will cease to function, all clock sources to the module are shutdown. If the UARTn clock, f_H, is off while a transmission is still in progress, then the transmission will be paused until the UARTn clock source derived from the microcontroller is activated. In a similar way, if the MCU enters the IDLE or SLEEP mode while receiving data, then the reception of data will likewise be paused. When the MCU enters the IDLE or SLEEP mode, note that the UnSR, UnCR1, UnCR2, UnCR3, UFCRn, RxCNTn and TXR_RXRn registers, as well as the BRDHn/BRDLn register will not be affected. It is recommended to make sure first that the UARTn data transmission or reception has been finished before the microcontroller enters the IDLE or SLEEP mode.

The UARTn function contains a receiver RXn/TXn pin wake-up function, which is enabled or disabled by the WAKEn bit in the UnCR2 register. If this bit, along with the UARTn enable bit, UARTENn, the receiver enable bit, RXENn and the receiver interrupt bit, RInE, are all set when the UARTn clock (f_H) is off, then a falling edge on the RXn/TXn pin will trigger an RXn/TXn pin wake-up UARTn interrupt. Note that as it takes certain system clock cycles after a wake-up, before normal microcontroller operation resumes, any data received during this time on the RXn/TXn pin will be ignored.

For a UARTn wake-up interrupt to occur, in addition to the bits for the wake-up being set, the global interrupt enable bit, EMI, and the UARTn interrupt enable bit, URnE, must be set. If the EMI and URnE bits are not set then only a wake-up event will occur and no interrupt will be generated. Note also that as it takes certain system clock cycles after a wake-up before normal microcontroller resumes, the UARTn interrupt will not be generated until after this time has elapsed.

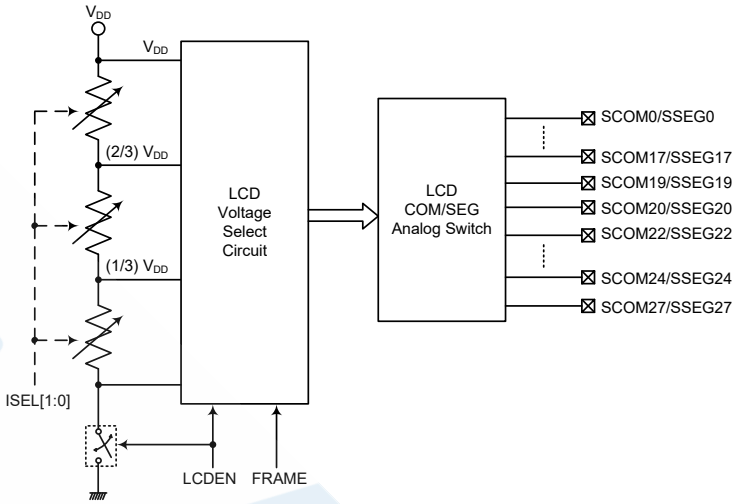
SCOM/SSEG Function for LCD

The device has the capability of driving external LCD panels. The common and segment pins for LCD driving, SCOM0~SCOM17, SCOM19~SCOM20, SCOM22~SCOM24, SCOM27 and SSEG0~SSEG17, SSEG19~SSEG20, SSEG22~SSEG24, SSEG27, are pin-shared with certain pins on the I/O ports. The LCD signals, COM and SEG, are generated using the application program.

LCD Operation

An external LCD panel can be driven using the device by configuring the I/O pins as common pins and segment pins. The LCD driver function is controlled using the LCD control registers which in addition to controlling the overall on/off function also controls the bias voltage setup function. This enables the LCD COM and SEG driver to generate the necessary V_{SS}, (1/3) V_{DD}, (2/3) V_{DD} and V_{DD} voltage levels for LCD 1/3 bias operation.

The LCDEN bit in the SLCDC0 register is the overall master control for the LCD driver. This bit is used in conjunction with the corresponding pin-shared function selection bits to select which I/O pins are used for LCD driving. Note that the corresponding Port Control register does not need to first setup the pins as outputs to enable the LCD driver operation.



Software Controlled LCD Driver Structure

LCD Frames

A cyclic LCD waveform includes two frames known as Frame 0 and Frame 1 for which the following offers a functional explanation.

Frame 0

To select Frame 0, clear the FRAME bit in the SLCDC0 register to 0.

In frame 0, the COM signal output can have a value of V_{DD} or a V_{BIAS} value of (1/3) V_{DD}. The SEG signal output can have a value of V_{SS} or a V_{BIAS} value of (2/3) V_{DD}.

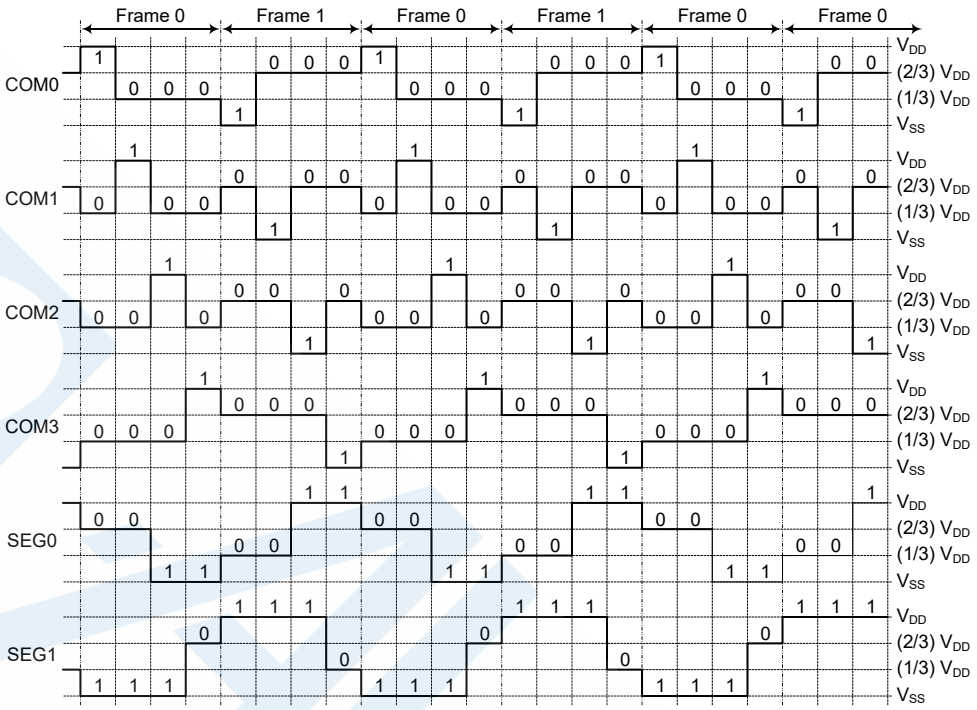
Frame 1

To select Frame 1, set the FRAME bit in the SLCDC0 register to 1.

In frame 1, the COM signal output can have a value of V_{SS} or a V_{BIAS} value of (2/3) V_{DD}. The SEG signal output can have a value of V_{DD} or a V_{BIAS} value of (1/3) V_{DD}.

The SCOMm waveform is controlled by the application program using the FRAME bit in the SLCDC0 register and the corresponding pin-shared I/O data bit for the respective SCOMm pin to determine whether the SCOMm output has a value of V_{DD}, V_{SS} or V_{BIAS}. The SSEGn waveform is controlled in a similar way using the FRAME bit and the corresponding pin-shared I/O data bit for the respective SSEGn pin to determine whether the SSEGn output has a value of V_{DD}, V_{SS} or V_{BIAS}.

The accompanying waveform diagram shows a typical 1/3 bias LCD waveform generated using the application program together with the LCD voltage select circuit. Note that the depiction of a “1” in the diagram illustrates an illuminated LCD pixel. The COM signal polarity generated on pins SCOM0~SCOM17, SCOM19~SCOM20, SCOM22~SCOM24, SCOM27, whether “0” or “1”, are generated using the corresponding pin-shared I/O data register bit.



Note: The logical values shown in the above diagram are the corresponding pin-shared I/O data bit value.

1/3 Bias LCD Waveform – 4-COM & 2-SEG Application

LCD Control Registers

The LCD SCOM and SSEG driver enables a range of selections to be provided to suit the requirement of the LCD panel which is being used. The bias resistor choice is implemented using the ISEL1 and ISEL0 bits in the SLCDC0 register. All SCOM and SSEG pins are pin-shared with I/O pins and selected as SCOM and SSEG pins using the corresponding pin function selection bits in the SLCDS0~SLCDS3 registers respectively.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SLCDC0	FRAME	ISEL1	ISEL0	LCDEN	—	—	—	—
SLCDS0	COMSEGS7	COMSEGS6	COMSEGS5	COMSEGS4	COMSEGS3	COMSEGS2	COMSEGS1	COMSEGS0
SLCDS1	COMSEGS15	COMSEGS14	COMSEGS13	COMSEGS12	COMSEGS11	COMSEGS10	COMSEGS9	COMSEGS8
SLCDS2	COMSEGS23	COMSEGS22	D5	COMSEGS20	COMSEGS19	D2	COMSEGS17	COMSEGS16
SLCDS3	—	—	D5	D4	COMSEGS27	D2	D1	COMSEGS24

LCD Driver Control Register List

- **SLCDC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	FRAME	ISEL1	ISEL0	LCDEN	—	—	—	—
R/W	R/W	R/W	R/W	R/W	—	—	—	—
POR	0	0	0	0	—	—	—	—

Bit 7 **FRAME**: SCOMn/SSEGn Output Frame selection
0: Frame 0
1: Frame 1

Bit 6~5 **ISEL1~ISEL0**: Select resistor for R type LCD bias current (@V_{DD}=3V)
00: 3×200kΩ (1/3 Bias), I_{BIAS}=8.3μA
01: 3×100kΩ (1/3 Bias), I_{BIAS}=16.6μA
10: 3×33.3kΩ (1/3 Bias), I_{BIAS}=50μA
11: 3×16.6kΩ (1/3 Bias), I_{BIAS}=100μA

Bit 4 **LCDEN**: LCD control bit
0: Off
1: On
When the LCDEN bit is cleared to 0, then the SCOMm and SSEGn outputs will be fixed at a V_{SS} level.

Bit 3~0 Unimplemented, read as "0"

• **SLCDS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	COMSEGS7	COMSEGS6	COMSEGS5	COMSEGS4	COMSEGS3	COMSEGS2	COMSEGS1	COMSEGS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **COMSEGS7**: SCOM7/SSEG7 pin function selection
0: SCOM7
1: SSEG7

Bit 6 **COMSEGS6**: SCOM6/SSEG6 pin function selection
0: SCOM6
1: SSEG6

Bit 5 **COMSEGS5**: SCOM5/SSEG5 pin function selection
0: SCOM5
1: SSEG5

Bit 4 **COMSEGS4**: SCOM4/SSEG4 pin function selection
0: SCOM4
1: SSEG4

Bit 3 **COMSEGS3**: SCOM3/SSEG3 pin function selection
0: SCOM3
1: SSEG3

Bit 2 **COMSEGS2**: SCOM2/SSEG2 pin function selection
0: SCOM2
1: SSEG2

Bit 1 **COMSEGS1**: SCOM1/SSEG1 pin function selection
0: SCOM1
1: SSEG1

Bit 0 **COMSEGS0**: SCOM0/SSEG0 pin function selection
0: SCOM0
1: SSEG0

• **SLCDS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	COMSEGS15	COMSEGS14	COMSEGS13	COMSEGS12	COMSEGS11	COMSEGS10	COMSEGS9	COMSEGS8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **COMSEGS15**: SCOM15/SSEG15 pin function selection
0: SCOM15
1: SSEG15

Bit 6 **COMSEGS14**: SCOM14/SSEG14 pin function selection
0: SCOM14
1: SSEG14

Bit 5 **COMSEGS13**: SCOM13/SSEG13 pin function selection
0: SCOM13
1: SSEG13

Bit 4 **COMSEGS12**: SCOM12/SSEG12 pin function selection
0: SCOM12
1: SSEG12

Bit 3 **COMSEGS11**: SCOM11/SSEG11 pin function selection
0: SCOM11
1: SSEG11

Bit 2 **COMSEGS10**: SCOM10/SSEG10 pin function selection
0: SCOM10
1: SSEG10

Bit 1 **COMSEGS9**: SCOM9/SSEG9 pin function selection
0: SCOM9
1: SSEG9

Bit 0 **COMSEGS8**: SCOM8/SSEG8 pin function selection
0: SCOM8
1: SSEG8

• **SLCDS2 Register**

Bit	7	6	5	4	3	2	1	0
Name	COMSEGS23	COMSEGS22	D5	COMSEGS20	COMSEGS19	D2	COMSEGS17	COMSEGS16
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **COMSEGS23**: SCOM23/SSEG23 pin function selection
0: SCOM23
1: SSEG23

Bit 6 **COMSEGS22**: SCOM22/SSEG22 pin function selection
0: SCOM22
1: SSEG22

Bit 5 **D5**: Reserved, must be fixed at "0"

Bit 4 **COMSEGS20**: SCOM20/SSEG20 pin function selection
0: SCOM20
1: SSEG20

Bit 3 **COMSEGS19**: SCOM19/SSEG19 pin function selection
0: SCOM19
1: SSEG19

Bit 2 **D2**: Reserved, must be fixed at "0"

Bit 1 **COMSEGS17**: SCOM17/SSEG17 pin function selection
0: SCOM17
1: SSEG17

Bit 0 **COMSEGS16**: SCOM16/SSEG16 pin function selection
0: SCOM16
1: SSEG16

• **SLCDS3 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	D5	D4	COMSEGS27	D2	D1	COMSEGS24
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as "0"

Bit 5~4 **D5~D4**: Reserved, must be fixed at "00"

Bit 3 **COMSEGS27**: SCOM27/SSEG27 pin function selection
0: SCOM27
1: SSEG27

Bit 2~1 **D2~D1**: Reserved, must be fixed at "00"

Bit 0 **COMSEGS24**: SCOM24/SSEG24 pin function selection
0: SCOM24
1: SSEG24

Low Voltage Detector – LVD

The device has a Low Voltage Detector function, also known as LVD. This enables the device to monitor the power supply voltage, V_{DD}, and provide a warning signal should it fall below a certain level. This function may be especially useful in battery applications where the supply voltage will gradually reduce as the battery ages, as it allows an early warning battery low signal to be generated. The Low Voltage Detector also has the capability of generating an interrupt signal.

LVD Register

The Low Voltage Detector function is controlled using a single register with the name LVDC. Three bits in this register, VLVD2~VLVD0, are used to select one of six fixed voltages below which a low voltage condition will be determined. A low voltage condition is indicated when the LVDO bit is set. If the LVDO bit is low, this indicates that the V_{DD} voltage is above the preset low voltage value. The LVDEN bit is used to control the overall on/off function of the low voltage detector. Setting the bit high will enable the low voltage detector. Clearing the bit to zero will switch off the internal low voltage detector circuits. As the low voltage detector will consume a certain amount of power, it may be desirable to switch off the circuit when not in use, an important consideration in power sensitive battery powered applications.

• **LVDC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	LVDO	LVDEN	—	VLVD2	VLVD1	VLVD0
R/W	—	—	R	R/W	—	R/W	R/W	R/W
POR	—	—	0	0	—	0	0	0

Bit 7~6 Unimplemented, read as "0"

Bit 5 **LVDO**: LVD output flag
0: No Low Voltage detected
1: Low Voltage detected

Bit 4 **LVDEN**: Low Voltage Detector Enable control
0: Disable
1: Enable

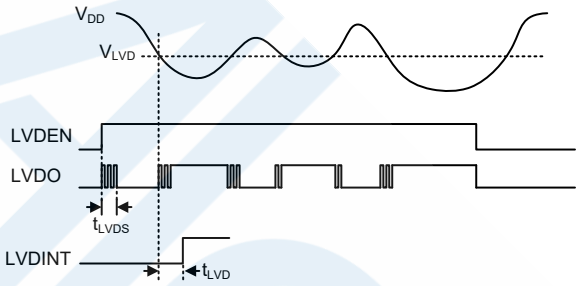
Bit 3 Unimplemented, read as "0"

Bit 2~0 **VLVD2~VLVD0**: LVD Voltage selection
000: 1.8V
001: 2.0V

010: 2.4V
011: 2.7V
100: 3.0V
101: 3.3V
110: Reserved
111: Reserved

LVD Operation

The Low Voltage Detector function operates by comparing the power supply voltage, V_{DD}, with a pre-specified voltage level stored in the LVDC register. This has a range of between 1.8V and 3.3V. When the power supply voltage, V_{DD}, falls below this pre-determined value, the LVDO bit will be set high indicating a low power supply voltage condition. When the device is in the SLEEP mode, the low voltage detector will be disabled even if the LVDEN bit is high. After enabling the Low Voltage Detector, a time delay, t_{LVDS}, should be allowed for the circuitry to stabilise before reading the LVDO bit. Note also that as the V_{DD} voltage may rise and fall rather slowly, at the voltage nears that of V_{LVD}, there may be multiple bit LVDO transitions.



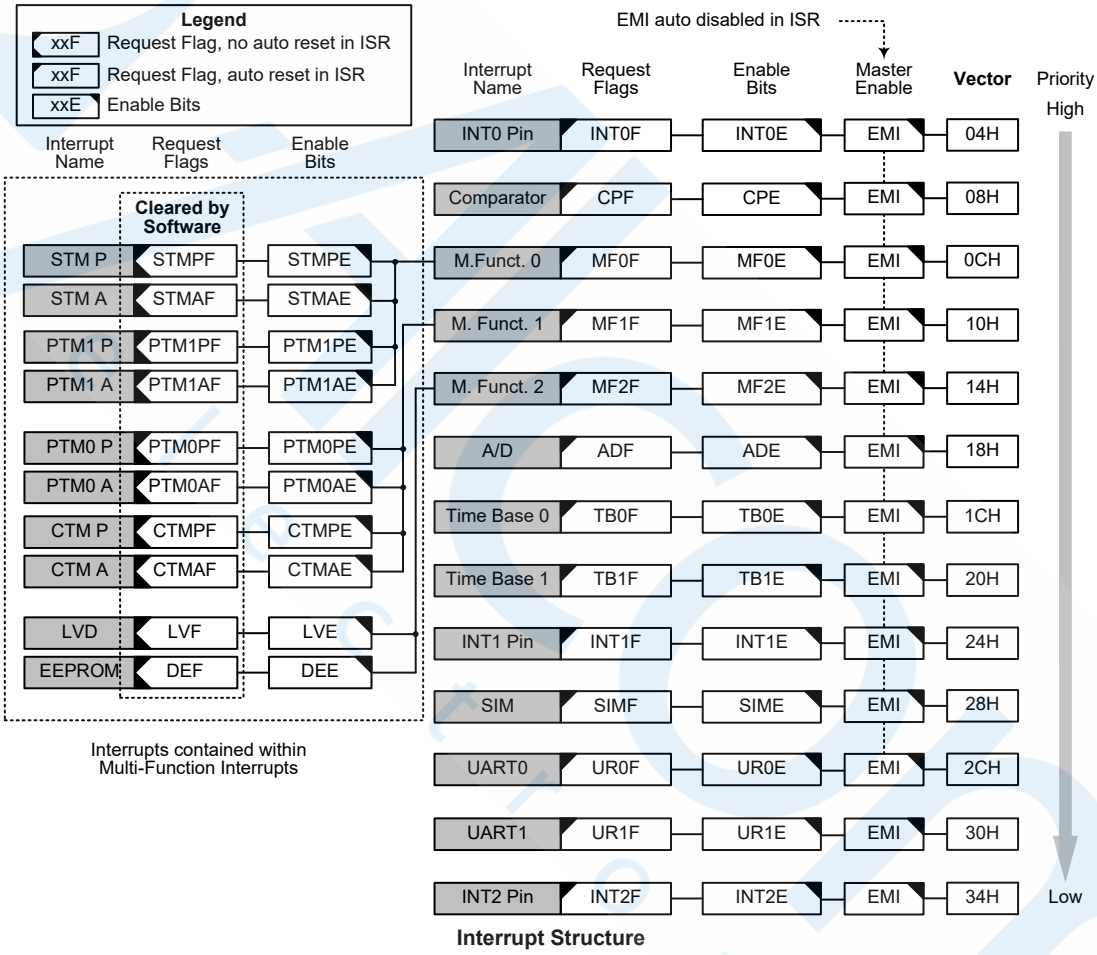
LVD Operation

The Low Voltage Detector also has its own interrupt which is contained within one of the Multi-function interrupts, providing an alternative means of low voltage detection, in addition to polling the LVDO bit. The interrupt will only be generated after a delay of t_{LVD} after the LVDO bit has been set high by a low voltage condition, i.e., V_{DD} falls below the preset LVD voltage. In this case, the LVF interrupt request flag will be set, causing an interrupt to be generated. This will cause the device to wake up from the IDLE Mode, however if the Low Voltage Detector wake-up function is not required then the LVF flag should be first set high before the device enters the IDLE Mode.

Interrupts

Interrupts are an important part of any microcontroller system. When an external event or an internal function such as a Timer Module or an A/D converter requires microcontroller attention, their corresponding interrupt will enforce a temporary suspension of the main program allowing the microcontroller to direct attention to their respective needs. The device contains several external interrupt and internal interrupt functions. The external interrupts are generated by the action of the external INT0, INT1and INT2 pins, while the internal interrupts are generated by various internal functions such as TMs, Time Base, LVD, EEPROM, SIM, UART and the A/D converter, etc.

The various interrupt enable bits, together with their associated request flags, are shown in the accompanying diagrams with their order of priority. Some interrupt sources have their own individual vector while others share the same multi-function interrupt vector.



Interrupt Structure

Interrupt Registers

Overall interrupt control, which basically means the setting of request flags when certain microcontroller conditions occur and the setting of interrupt enable bits by the application program, is controlled by a series of registers, located in the Special Purpose Data Memory. The registers fall into three categories. The first is the INTC0~INTC3 registers which setup the primary interrupts, the second is the MFI0~MFI2 registers which setup the Multi-function interrupts. Finally there is an INTEG register which setup the external interrupt trigger edge type.

Each register contains a number of enable bits to enable or disable individual registers as well as interrupt flags to indicate the presence of an interrupt request. The naming convention of these follows a specific pattern. First is listed an abbreviated interrupt type, then the (optional) number of that interrupt followed by either an "E" for enable/disable bit or "F" for request flag.

Function	Enable Bit	Request Flag	Notes
Global	EMI	—	—
INTn Pins	INTnE	INTnF	n=0~2
Comparator	CPE	CPF	—
Multi-function	MFnE	MFnF	n=0~2
A/D Converter	ADE	ADF	—
Time Bases	TBnE	TBnF	n=0~1
SIM	SIME	SIMF	—
UART	URnE	URnF	n=0~1
LVD	LVE	LVF	—
EEPROM	DEE	DEF	—
CTM	CTMPE	CTMPF	—
	CTMAE	CTMAF	—
STM	STMPE	STMPF	—
	STMAE	STMAF	—
PTM	PTMnPE	PTMnPF	n=0~1
	PTMnAE	PTMnAF	

Interrupt Register Bit Naming Conventions

Register Name	Bit							
	7	6	5	4	3	2	1	0
INTEG	—	—	INT2S1	INT2S0	INT1S1	INT1S0	INT0S1	INT0S0
INTC0	—	MF0F	CPF	INT0F	MF0E	CPE	INT0E	EMI
INTC1	TB0F	ADF	MF2F	MF1F	TB0E	ADE	MF2E	MF1E
INTC2	UR0F	SIMF	INT1F	TB1F	UR0E	SIME	INT1E	TB1E
INTC3	—	—	INT2F	UR1F	—	—	INT2E	UR1E
MFI0	PTM1AF	PTM1PF	STMAF	STMPF	PTM1AE	PTM1PE	STMAE	STMPE
MFI1	CTMAF	CTMPF	PTM0AF	PTM0PF	CTMAE	CTMPE	PTM0AE	PTM0PE
MFI2	—	—	DEF	LVF	—	—	DEE	LVE

Interrupt Register List

- **INTEG Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	INT2S1	INT2S0	INT1S1	INT1S0	INT0S1	INT0S0
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as "0"

Bit 5~4 **INT2S1~INT2S0**: Interrupt edge control for INT2 pin
00: Disable
01: Rising edge
10: Falling edge
11: Rising and falling edges

Bit 3~2 **INT1S1~INT1S0**: Interrupt edge control for INT1 pin
00: Disable
01: Rising edge
10: Falling edge
11: Rising and falling edges

Bit 1~0 **INT0S1~INT0S0**: Interrupt edge control for INT0 pin
00: Disable
01: Rising edge
10: Falling edge
11: Rising and falling edges

• **INTC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	MF0F	CPF	INT0F	MF0E	CPE	INT0E	EMI
R/W	—	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	0	0	0	0	0	0	0

Bit 7 Unimplemented, read as "0"

Bit 6 **MF0F**: Multi-function 0 interrupt request flag
0: No request
1: Interrupt request

Bit 5 **CPF**: Comparator interrupt request flag
0: No request
1: Interrupt request

Bit 4 **INT0F**: INT0 interrupt request flag
0: No request
1: Interrupt request

Bit 3 **MF0E**: Multi-function 0 interrupt control
0: Disable
1: Enable

Bit 2 **CPE**: Comparator interrupt control
0: Disable
1: Enable

Bit 1 **INT0E**: INT0 interrupt control
0: Disable
1: Enable

Bit 0 **EMI**: Global interrupt control
0: Disable
1: Enable

• **INTC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	TB0F	ADF	MF2F	MF1F	TB0E	ADE	MF2E	MF1E
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **TB0F**: Time Base 0 interrupt request flag
0: No request
1: Interrupt request

Bit 6 **ADF**: A/D Converter interrupt request flag
0: No request
1: Interrupt request

Bit 5 **MF2F**: Multi-function 2 interrupt request flag
0: No request
1: Interrupt request

Bit 4 **MF1F**: Multi-function 1 interrupt request flag
0: No request
1: Interrupt request

Bit 3 **TB0E**: Time Base 0 interrupt control
0: Disable
1: Enable

Bit 2 **ADE**: A/D Converter interrupt control
0: Disable
1: Enable

Bit 1 **MF2E**: Multi-function 2 interrupt control
0: Disable
1: Enable

Bit 0 **MF1E**: Multi-function 1 interrupt control
0: Disable
1: Enable

• **INTC2 Register**

Bit	7	6	5	4	3	2	1	0
Name	UR0F	SIMF	INT1F	TB1F	UR0E	SIME	INT1E	TB1E
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **UR0F**: UART0 interrupt request flag
0: No request
1: Interrupt request

Bit 6 **SIMF**: SIM interrupt request flag
0: No request
1: Interrupt request

Bit 5 **INT1F**: INT1 interrupt request flag
0: No request
1: Interrupt request

Bit 4 **TB1F**: Time Base 1 interrupt request flag
0: No request
1: Interrupt request

Bit 3 **UR0E**: UART0 interrupt control
0: Disable
1: Enable

Bit 2 **SIME**: SIM interrupt control
0: Disable
1: Enable

Bit 1 **INT1E**: INT1 interrupt control
0: Disable
1: Enable

Bit 0 **TB1E**: Time Base 1 interrupt control
0: Disable
1: Enable

• **INTC3 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	INT2F	UR1F	—	—	INT2E	UR1E
R/W	—	—	R/W	R/W	—	—	R/W	R/W
POR	—	—	0	0	—	—	0	0

Bit 7~6 Unimplemented, read as “0”

Bit 5 **INT2F**: INT2 interrupt request flag
0: No request
1: Interrupt request

Bit 4 **UR1F**: UART1 interrupt request flag
0: No request
1: Interrupt request

Bit 3~2 Unimplemented, read as "0"

Bit 1 **INT2E**: INT2 interrupt control
0: Disable
1: Enable

Bit 0 **UR1E**: UART1 interrupt control
0: Disable
1: Enable

• **MFI0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PTM1AF	PTM1PF	STMAF	STMPF	PTM1AE	PTM1PE	STMAE	STMPE
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **PTM1AF**: PTM1 Comparator A match interrupt request flag
0: No request
1: Interrupt request
Note that this bit must be cleared to zero by the application program when the interrupt is serviced.

Bit 6 **PTM1PF**: PTM1 Comparator P match interrupt request flag
0: No request
1: Interrupt request
Note that this bit must be cleared to zero by the application program when the interrupt is serviced.

Bit 5 **STMAF**: STM Comparator A match interrupt request flag
0: No request
1: Interrupt request
Note that this bit must be cleared to zero by the application program when the interrupt is serviced.

Bit 4 **STMPF**: STM Comparator P match interrupt request flag
0: No request
1: Interrupt request
Note that this bit must be cleared to zero by the application program when the interrupt is serviced.

Bit 3 **PTM1AE**: PTM1 Comparator A match interrupt control
0: Disable
1: Enable

Bit 2 **PTM1PE**: PTM1 Comparator P match interrupt control
0: Disable
1: Enable

Bit 1 **STMAE**: STM Comparator A match interrupt control
0: Disable
1: Enable

Bit 0 **STMPE**: STM Comparator P match interrupt control
0: Disable
1: Enable

• **MFI1 Register**

Bit	7	6	5	4	3	2	1	0
Name	CTMAF	CTMPF	PTM0AF	PTM0PF	CTMAE	CTMPE	PTM0AE	PTM0PE
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **CTMAF**: CTM Comparator A match interrupt request flag
0: No request
1: Interrupt request
Note that this bit must be cleared to zero by the application program when the interrupt is serviced.

Bit 6 **CTMPF**: CTM Comparator P match interrupt request flag
0: No request
1: Interrupt request
Note that this bit must be cleared to zero by the application program when the interrupt is serviced.

Bit 5 **PTM0AF**: PTM0 Comparator A match interrupt request flag
0: No request
1: Interrupt request
Note that this bit must be cleared to zero by the application program when the interrupt is serviced.

Bit 4 **PTM0PF**: PTM0 Comparator P match interrupt request flag
0: No request
1: Interrupt request
Note that this bit must be cleared to zero by the application program when the interrupt is serviced.

Bit 3 **CTMAE**: CTM Comparator A match interrupt control
0: Disable
1: Enable

Bit 2 **CTMPE**: CTM Comparator P match interrupt control
0: Disable
1: Enable

Bit 1 **PTM0AE**: PTM0 Comparator A match interrupt control
0: Disable
1: Enable

Bit 0 **PTM0PE**: PTM0 Comparator P match interrupt control
0: Disable
1: Enable

• **MFI2 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	DEF	LVF	—	—	DEE	LVE
R/W	—	—	R/W	R/W	—	—	R/W	R/W
POR	—	—	0	0	—	—	0	0

Bit 7~6 Unimplemented, read as "0"

Bit 5 **DEF**: Data EEPROM interrupt request flag
0: No request
1: Interrupt request
Note that this bit must be cleared to zero by the application program when the interrupt is serviced.

Bit 4 **LVF**: LVD interrupt request flag
0: No request
1: Interrupt request

Note that this bit must be cleared to zero by the application program when the interrupt is serviced.

Bit 3~2 Unimplemented, read as "0"

Bit 1 **DEE**: Data EEPROM interrupt control
0: Disable
1: Enable

Bit 0 **LVE**: LVD interrupt control
0: Disable
1: Enable

Interrupt Operation

When the conditions for an interrupt event occur, such as a TM Comparator P or Comparator A match or A/D conversion completion etc., the relevant interrupt request flag will be set. Whether the request flag actually generates a program jump to the relevant interrupt vector is determined by the condition of the interrupt enable bit. If the enable bit is set high then the program will jump to its relevant vector, if the enable bit is zero then although the interrupt request flag is set an actual interrupt will not be generated and the program will not jump to the relevant interrupt vector. The global interrupt enable bit, if cleared to zero, will disable all interrupts.

When an interrupt is generated, the Program Counter, which stores the address of the next instruction to be executed, will be transferred onto the stack. The Program Counter will then be loaded with a new address which will be the value of the corresponding interrupt vector. The microcontroller will then fetch its next instruction from this interrupt vector. The instruction at this vector will usually be a "JMP" which will jump to another section of program which is known as the interrupt service routine. Here is located the code to control the appropriate interrupt. The interrupt service routine must be terminated with a "RETI", which retrieves the original Program Counter address from the stack and allows the microcontroller to continue with normal execution at the point where the interrupt occurred.

Once an interrupt subroutine is serviced, all the other interrupts will be blocked, as the global interrupt enable bit, EMI bit will be cleared automatically. This will prevent any further interrupt nesting from occurring. However, if other interrupt requests occur during this interval, although the interrupt will not be immediately serviced, the request flag will still be recorded.

If an interrupt requires immediate servicing while the program is already in another interrupt service routine, the EMI bit should be set after entering the routine, to allow interrupt nesting. If the stack is full, the interrupt request will not be acknowledged, even if the related interrupt is enabled, until the Stack Pointer is decremented. If immediate service is desired, the stack must be prevented from becoming full. In case of simultaneous requests, the interrupt structure diagram shows the priority that is applied. All of the interrupt request flags when set will wake up the device if it is in SLEEP or IDLE Mode, however to prevent a wake-up from occurring the corresponding flag should be set before the device is in SLEEP or IDLE Mode.

External Interrupts

The external interrupts are controlled by signal transitions on the pins INT0~INT2. An external interrupt request will take place when the external interrupt request flags, INT0F~INT2F, are set, which will occur when a transition, whose type is chosen by the edge select bits, appears on the external interrupt pins. To allow the program to branch to their respective interrupt vector addresses, the global interrupt enable bit, EMI, and respective external interrupt enable bit, INT0E~INT2E, must first be set. Additionally the correct interrupt edge type must be selected using the INTEG register to enable the external interrupt function and to choose the trigger edge type. As the external interrupt pins are pin-shared with I/O pins, they can only be configured as external interrupt pins if

their external interrupt enable bit in the corresponding interrupt register has been set and the external interrupt pin is selected by the corresponding pin-shared function selection bits. The pin must also be setup as an input by setting the corresponding bit in the port control register. When the interrupt is enabled, the stack is not full and the correct transition type appears on the external interrupt pin, a subroutine call to the external interrupt vector, will take place. When the interrupt is serviced, the external interrupt request flags, INT0F~INT2F, will be automatically reset and the EMI bit will be automatically cleared to disable other interrupts. Note that any pull-high resistor selections on the external interrupt pins will remain valid even if the pin is used as an external interrupt input.

The INTEG register is used to select the type of active edge that will trigger the external interrupt. A choice of either rising or falling or both edge types can be chosen to trigger an external interrupt. Note that the INTEG register can also be used to disable the external interrupt function.

Comparator Interrupt

The comparator interrupt is controlled by the internal comparator. A comparator interrupt request will take place when the comparator interrupt request flag, CPF, is set, a situation that will occur when the comparator output bit changes state. To allow the program to branch to its interrupt vector address, the global interrupt enable bit, EMI, and comparator interrupt enable bit, CPE, must first be set. When the interrupt is enabled, the stack is not full and the comparator inputs generate a comparator output transition, a subroutine call to the comparator interrupt vector, will take place. When the interrupt is serviced, the comparator interrupt request flag will be automatically reset and the EMI bit will be automatically cleared to disable other interrupts.

Multi-function Interrupts

Within the device there are three Multi-function interrupts. Unlike the other independent interrupts, these interrupts have no independent source, but rather are formed from other existing interrupt sources, namely the TM interrupts, LVD interrupt and EEPROM interrupt.

A Multi-function interrupt request will take place when any of the Multi-function interrupt request flags MFnF are set. The Multi-function interrupt flag will be set when any of their included functions generate an interrupt request flag. To allow the program to branch to their respective interrupt vector addresses, when the Multi-function interrupt is enabled and the stack is not full, and any one of the interrupts contained within each of the Multi-function interrupt occurs, a subroutine call to the related Multi-function interrupt vector will take place. When the interrupt is serviced, the related Multi-function request flag will be automatically reset and the EMI bit will be automatically cleared to disable other interrupts.

However, it must be noted that, although the Multi-function Interrupt request flags will be automatically reset when the interrupt is serviced, the request flags from the original source of the Multi-function interrupt will not be automatically reset and must be manually reset by the application program.

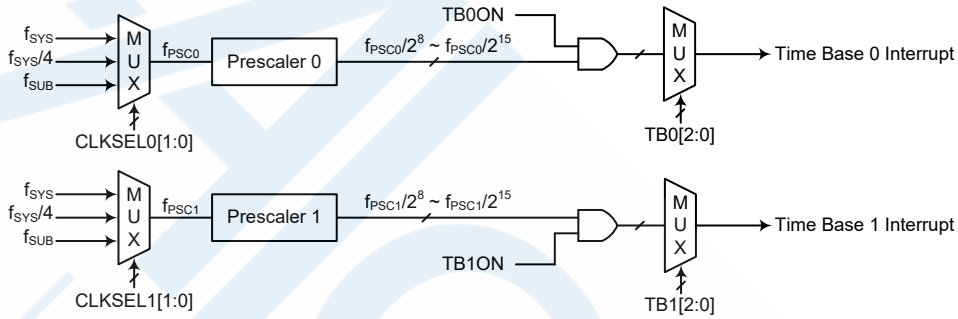
A/D Converter Interrupt

The A/D Converter Interrupt is controlled by the termination of an A/D conversion process. An A/D Converter Interrupt request will take place when the A/D Converter Interrupt request flag, ADF, is set, which occurs when the A/D conversion process finishes. To allow the program to branch to its interrupt vector address, the global interrupt enable bit, EMI, and A/D Interrupt enable bit, ADE, must first be set. When the interrupt is enabled, the stack is not full and the A/D conversion process has ended, a subroutine call to the A/D Converter Interrupt vector, will take place. When the interrupt is serviced, the A/D Converter Interrupt flag, ADF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

Time Base Interrupts

The function of the Time Base Interrupt is to provide regular time signal in the form of an internal interrupt. They are controlled by the overflow signals from their respective timer functions. When these happen their respective interrupt request flags, TB0F or TB1F, will be set. To allow the program to branch to their respective interrupt vector addresses, the global interrupt enable bit, EMI and Time Base enable bit, TB0E or TB1E, must first be set. When the interrupt is enabled, the stack is not full and the Time Base overflows, a subroutine call to their respective vector locations will take place. When the interrupt is serviced, the respective interrupt request flag, TB0F or TB1F, will be automatically reset and the EMI bit will be cleared to disable other interrupts.

The purpose of the Time Base Interrupt is to provide an interrupt signal at fixed time periods. Their clock sources f_{PSC0} or f_{PSC1}, originate from the internal clock source f_{SYS}, $f_{SYS}/4$ or f_{SUB} and then passes through a divider, the division ratio of which is selected by programming the appropriate bits in the TB0C or TB1C register to obtain longer interrupt periods whose value ranges. The clock source which in turn controls the Time Base interrupt period is selected using the CLKSEL0[1:0] and CLKSEL1[1:0] bits in the PSC0R and PSC1R register respectively.



Time Base Interrupts

• PSC0R Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CLKSEL01	CLKSEL00
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 **CLKSEL01~CLKSEL00**: Prescaler 0 clock source selection
00: f_{SYS}
01: $f_{SYS}/4$
1x: f_{SUB}

• PSC1R Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CLKSEL11	CLKSEL10
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 **CLKSEL11~CLKSEL10**: Prescaler 1 clock source selection
00: f_{SYS}
01: $f_{SYS}/4$
1x: f_{SUB}

• **TB0C Register**

Bit	7	6	5	4	3	2	1	0
Name	TB0ON	—	—	—	—	TB02	TB01	TB00
R/W	R/W	—	—	—	—	R/W	R/W	R/W
POR	0	—	—	—	—	0	0	0

Bit 7 **TB0ON**: Time Base 0 Enable Control
0: Disable
1: Enable

Bit 6~3 Unimplemented, read as "0"

Bit 2~0 **TB02~TB00**: Time Base 0 time-out period selection
000: $2^8/f_{PSC0}$
001: $2^9/f_{PSC0}$
010: $2^{10}/f_{PSC0}$
011: $2^{11}/f_{PSC0}$
100: $2^{12}/f_{PSC0}$
101: $2^{13}/f_{PSC0}$
110: $2^{14}/f_{PSC0}$
111: $2^{15}/f_{PSC0}$

• **TB1C Register**

Bit	7	6	5	4	3	2	1	0
Name	TB1ON	—	—	—	—	TB12	TB11	TB10
R/W	R/W	—	—	—	—	R/W	R/W	R/W
POR	0	—	—	—	—	0	0	0

Bit 7 **TB1ON**: Time Base 1 Enable Control
0: Disable
1: Enable

Bit 6~3 Unimplemented, read as "0"

Bit 2~0 **TB12~TB10**: Time Base 1 time-out period selection
000: $2^8/f_{PSC1}$
001: $2^9/f_{PSC1}$
010: $2^{10}/f_{PSC1}$
011: $2^{11}/f_{PSC1}$
100: $2^{12}/f_{PSC1}$
101: $2^{13}/f_{PSC1}$
110: $2^{14}/f_{PSC1}$
111: $2^{15}/f_{PSC1}$

TM Interrupts

The Compact, Standard and Periodic TMs each have two interrupts, one comes from the comparator A match situation and the other comes from the comparator P match situation. All of the TM interrupts are contained within the Multi-function Interrupts. For all of the TM types there are two interrupt request flags and two enable control bits. A TM interrupt request will take place when any of the TM request flags are set, a situation which occurs when a TM comparator P or A match situation happens.

To allow the program to branch to their respective interrupt vector addresses, the global interrupt enable bit, EMI, respective TM Interrupt enable bit, and relevant Multi-function Interrupt enable bit, MFnE, must first be set. When the interrupt is enabled, the stack is not full and a TM comparator match situation occurs, a subroutine call to the relevant Multi-function Interrupt vector location, will take place. When the TM interrupt is serviced, the EMI bit will be automatically cleared to disable other interrupts. However, only the related MFnF flag will be automatically cleared. As the TM

interrupt request flags will not be automatically cleared, they have to be cleared by the application program.

Serial Interface Module Interrupt

The Serial Interface Module Interrupt, also known as the SIM interrupt. An SIM Interrupt request will take place when the SIM Interrupt request flag, SIMF, is set, which occurs when a byte of data has been received or transmitted by the SIM interface, an I^2C slave address matches or I^2C bus time-out occurs. To allow the program to branch to its interrupt vector address, the global interrupt enable bit, EMI and the Serial Interface Interrupt enable bit, SIME, must first be set. When the interrupt is enabled, the stack is not full and any of the above described situations occurs, a subroutine call to the SIM Interrupt vector will take place. When the Serial Interface Interrupt is serviced, the SIM Interrupt flag, SIMF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

UART Interrupts

Each UART Interrupt is controlled by several UART transfer conditions. When one of these conditions occurs, an interrupt pulse will be generated to get the attention of the microcontroller. These conditions are a transmitter data register empty, transmitter idle, receiver reaching FIFO trigger level, receiver overrun, address detect and an RXn/TXn pin wake-up. To allow the program to branch to their corresponding interrupt vector addresses, the global interrupt enable bit, EMI, and UARTn Interrupt enable bit, URnE, must first be set. When the interrupt is enabled, the stack is not full and any of the conditions described above occurs, a subroutine call to the UARTn Interrupt vector, will take place. When the interrupt is serviced, the UARTn Interrupt flag, URnF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts. However, the UnSR register flags will only be cleared when certain actions are taken by the UARTn, the details of which are given in the UART Interfaces section.

LVD Interrupt

The Low Voltage Detector Interrupt is contained within the Multi-function Interrupt. An LVD Interrupt request will take place when the LVD Interrupt request flag, LVF, is set, which occurs when the Low Voltage Detector function detects a low power supply voltage. To allow the program to branch to its interrupt vector address, the global interrupt enable bit, EMI, Low Voltage Interrupt enable bit, LVE, and associated Multi-function interrupt enable bit, MFnE, must first be set. When the interrupt is enabled, the stack is not full and a low voltage condition occurs, a subroutine call to the Multi-function Interrupt vector, will take place. When the Low Voltage Interrupt is serviced, the EMI bit will be automatically cleared to disable other interrupts. However, only the Multi-function interrupt request flag will be also automatically cleared. As the LVF flag will not be automatically cleared, it has to be cleared by the application program.

EEPROM Interrupt

The EEPROM Interrupt is contained within the Multi-function Interrupt. An EEPROM Interrupt request will take place when the EEPROM Interrupt request flag, DEF, is set, which occurs when an EEPROM Erase or Write cycle ends. To allow the program to branch to its interrupt vector address, the global interrupt enable bit, EMI, EEPROM Interrupt enable bit, DEE, and associated Multi-function interrupt enable bit, MFnE, must first be set. When the interrupt is enabled, the stack is not full and an EEPROM Erase or Write cycle ends, a subroutine call to the respective Multi-function Interrupt vector will take place. When the EEPROM Interrupt is serviced, the EMI bit will be automatically cleared to disable other interrupts. However, only the Multi-function interrupt request flag will be automatically cleared. As the DEF flag will not be automatically cleared, it has to be cleared by the application program.

Interrupt Wake-up Function

Each of the interrupt functions has the capability of waking up the microcontroller when in the SLEEP or IDLE Mode. A wake-up is generated when an interrupt request flag changes from low to high and is independent of whether the interrupt is enabled or not. Therefore, even though the device is in the SLEEP or IDLE Mode and its system oscillator stopped, situations such as external edge transitions on the external interrupt pins, a low power supply voltage or comparator input change may cause their respective interrupt flag to be set high and consequently generate an interrupt. Care must therefore be taken if spurious wake-up situations are to be avoided. If an interrupt wake-up function is to be disabled then the corresponding interrupt request flag should be set high before the device enters the SLEEP or IDLE Mode. The interrupt enable bits have no effect on the interrupt wake-up function.

Programming Considerations

By disabling the relevant interrupt enable bits, a requested interrupt can be prevented from being serviced, however, once an interrupt request flag is set, it will remain in this condition in the interrupt register until the corresponding interrupt is serviced or until the request flag is cleared by the application program.

Where a certain interrupt is contained within a Multi-function interrupt, then when the interrupt service routine is executed, as only the Multi-function interrupt request flags, MFnF, will be automatically cleared, the individual request flag for the function needs to be cleared by the application program.

It is recommended that programs do not use the "CALL" instruction within the interrupt service subroutine. Interrupts often occur in an unpredictable manner or need to be serviced immediately. If only one stack is left and the interrupt is not well controlled, the original control sequence will be damaged once a CALL subroutine is executed in the interrupt subroutine.

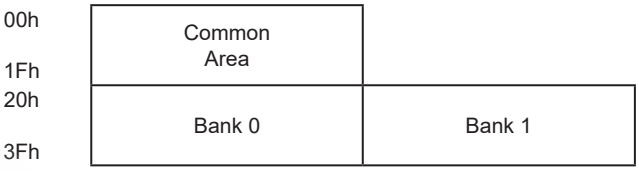
Every interrupt has the capability of waking up the microcontroller when it is in SLEEP or IDLE Mode, the wake-up being generated when the interrupt request flag changes from low to high. If it is required to prevent a certain interrupt from waking up the microcontroller then its respective request flag should be first set high before enter SLEEP or IDLE Mode.

As only the Program Counter is pushed onto the stack, then when the interrupt is serviced, if the contents of the accumulator, status register or other registers are altered by the interrupt service program, their contents should be saved to the memory at the beginning of the interrupt service routine.

To return from an interrupt subroutine, either a RET or RETI instruction may be executed. The RETI instruction in addition to executing a return to the main program also automatically sets the EMI bit high to allow further interrupts. The RET instruction however only executes a return to the main program leaving the EMI bit in its present zero state and therefore disabling the execution of further interrupts.

RF Transceiver

Memory Mapping

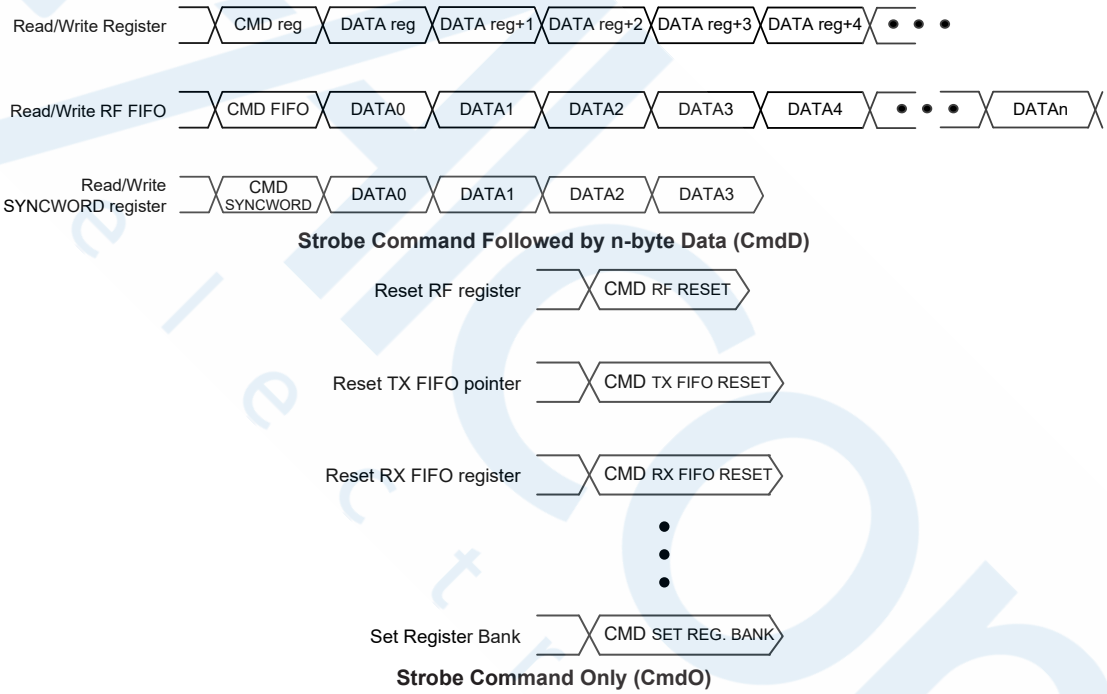


Common Area: It contains 32 bytes space. Accessing addresses 00h~1Fh always means to access the Common Area regardless of Bank Pointer configuration.

Bank 0~1: Each bank contains 32 bytes space. They are selected by the Bank Pointer.

The Bank Pointer, BANK[1:0], which is defined in the Common Area, can be set directly by the Set Register Bank command and read/written by the Control Register command.

Control Register Access



SFR Mapping and Bit Definition

Common Area Control Register

All control registers will be set to their initial value by power-on reset (POR). The software reset will set the control registers to their initial value except the FSYCK_EN, FSYCK_DIV[1:0], PWRON, GIO1S[2:0], GIO2S[2:0], PADDS[1:0], GIO4S[3:0], GIOPU4, GIOPU[2:1], SPIPU, SDO_TEN bits in the RC1, IO1, IO2 and IO3 registers. These bits keep unchanged after software reset.

Addr.	Name	Bit							
		7	6	5	4	3	2	1	0
00h	CFG1	OOK_EN	AGC_EN	RXCON_EN	DIR_EN	—	—	BANK[1:0]	
01h	RC1	PWRON	FSYCK_RDY	XCLK_RDY	XCLK_EN	FSYCK_DIV[1:0]		FSYCK_EN	RST_LL
02h	IRQ1	RXTO	RXFFOW	—	RXCRCF	RXDETS[1:0]		IRQCPOR	IRQPOR
03h	IRQ2	ARKTFIE	ATRCTIE	FIFOLTIE	RXERRIE	RXDETIE	CALCMPIE	RXCMPIE	TXCMPIE
04h	IRQ3	ARKTFIF	ATRCTIF	FIFOLTIF	RXERRIF	RXDETIF	CALCMPIF	RXCMPIF	TXCMPIF
06h	IO1	PADDS[1:0]		GIO2S[2:0]			GIO1S[2:0]		
07h	IO2	GIO4S[3:0]				—			
08h	IO3	SDO_TEN	SPIPU	—	GIOPU4	—	GIOPU[2:1]		—
09h	FIFO1	—	—	TXFFSA[5:0]					
0Ah	FIFO2	—	—	—	RXPL2F_EN	FFINF_EN	FFMG_EN	FFMG[1:0]	
0Bh	PKT1	TXPMLEN[7:0]							
0Ch	PKT2	PID[1:0]		TRAILER_EN	WHTFMT[0]	SYNCLEN[1:0]		—	
0Dh	PKT3	MCH_EN	FEC_EN	CRC_EN	CRCFMT	PLLEN_EN	PLHAC_EN	PLHLEN	PLH_EN
0Eh	PKT4	WHT_EN	WHTSD[6:0]						
0Fh	PKT5	TXDLEN[7:0]							
10h	PKT6	RXDLEN[7:0]							
11h	PKT7	RXPID[1:0]		DLY_RXS[2:0]			DLY_TXS[2:0]		
12h	PKT8	—		PLHA[5:0]					
13h	PKT9	PLHEA[7:0]							
14h	MOD1	DTR[7:0]							
15h	MOD2	RXFDOS[11:8]				DITHER[1:0]		—	DTR[8]
16h	MOD3	RXFDOS[7:0]							
17h	DM1	—	—	MDIV[5:0]					
18h	DM2	—	—	SDR[5:0]					
19h	DM3	CSF_SW_EN	FD_MOD[6:0]						
1Ah	DM4	—				CFO_DSEL	—	PH_DIFF_MOD	—
1Bh	DM5	FD_HOLD[7:0]							
1Eh	DM8	M_RATIO[7:0]							

Note: The addresses which are not listed in this table are reserved for future use, it is suggested not to change their initial values by any methods.

The reset value shown in the following register description tables means the software reset results of strobe command.

• **CFG1: Configuration Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	OOK_EN	AGC_EN	RXCON_EN	DIR_EN	—	—	BANK[1:0]	
R/W	R/W	R/W	R/W	R/W	—	—	R/W	
Reset	0	0	0	0	0	0	0	0

Bit 7 **OOK_EN**: OOK mode enable
0: GFSK mode
1: OOK mode, support Direct mode (DIR_EN=1) only

Bit 6 **AGC_EN**: AGC enable
0: Disable
1: Enable

Bit 5 **RXCON_EN**: RX continue mode enable
0: Disable
1: Enable
Note that this bit only affects normal RX mode and ATR RX mode without ARK function.

Bit 4 **DIR_EN**: Direct mode enable
0: TX/RX data from packet handling hardware
1: TX/RX data from/to external MCU directly

Bit 3~2 Reserved, must be "00"

Bit 1~0 **BANK[1:0]**: Control register bank selection
00: Bank 0
01: Bank 1
10: Bank 2
11: Reserved
This selection can be set by both the Set Register Bank command and Control Register command.

• **RC1: Reset/Clock Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	PWRON	FSYCK_RDY	XCLK_RDY	XCLK_EN	FSYCK_DIV[1:0]		FSYCK_EN	RST_LL
R/W	R/W	R	R	R/W	R/W		R/W	R/W
POR	1	—	—	—	0	0	0	—
Reset	—	0	0	1	—	—	—	0

Bit 7 **PWRON**: 3.3V power on flag
This bit is only set to 1 by power on reset and not affected by software reset of strobe command. After being set high, this bit should be cleared by application program. The firmware can check this flag status and determine whether to execute auto calibration in the Light Sleep mode.

Bit 6 **FSYCK_RDY**: FSYCK clock ready flag (ready only)
0: Not ready
1: Ready
This bit is used to indicate that whether the FSYCK clock is ready for operation. This bit will be automatically cleared when FSYCK_EN=0, when power on reset occurs or when a Deep Sleep command or an Idle command is received.

Bit 5 **XCLK_RDY**: XCLK clock ready flag (ready only)
0: Not ready
1: Ready
This bit is used to indicate whether the XCLK debounce counter is full and XCLK is ready for operation. Note that when exiting the Deep Sleep state, this flag may need a certain period before being set high. This bit will be automatically cleared to zero when XCLK_EN=0, when RST_LL=1, when power on reset occurs or when a software reset command, a Deep Sleep command or an Idle command is received.

Bit 4 **XCLK_EN**: XCLK clock enable
0: Disable
1: Enable
Setting this bit high will enable the XCLK path to the baseband block while clearing this bit to zero can save power if required. The XCLK clock should be enabled when writing data to the FIFO.

Bit 3~2 **FSYCK_DIV[1:0]**: FSYCK clock (XCLK division) selection
00: 1/1 XCLK
01: 1/2 XCLK
10: 1/4 XCLK
11: 1/8 XCLK

Bit 1 **FSYCK_EN**: FSYCK clock enable
0: Disable
1: Enable

Bit 0 **RST_LL**: Low voltage (1.2V) logic reset control
0: Release reset
1: Reset

• IRQ1: Interrupt Control Register 1

Bit	7	6	5	4	3	2	1	0
Name	RXTO	RXFFOW	—	RXCRCF	RXDETS[1:0]		IRQCPOR	IRQPOR
R/W	R	R	—	R/W	R/W		R/W	R/W
Reset	0	0	0	0	1	0	0	1

Bit 7 **RXTO**: RX time-out flag
0: RX time-out does not occur
1: RX time-out occurs
This flag will be set high by hardware when the RX time-out condition occurs and automatically cleared when a Light Sleep strobe command is received, when the RF transceiver enters the RX continuous mode, when WOR/WOT wake up occurs or when the RF transceiver enters the ARK TX/RX mode.

Bit 6 **RXFFOW**: RX FIFO overwrite flag
0: RX FIFO overwrite does not occur
1: RX FIFO overwrite occurs
This flag will be set high by hardware when the RX FIFO overwrite condition occurs and automatically cleared when an RX FIFO reset strobe command or an RX strobe command is received.

Bit 5 Reserved, must be "0"

Bit 4 **RXCRCF**: RX CRC failure flag

Bit 3~2 **RXDETS[1:0]**: RX detect selection
00: Detect carry
01: Reserved
10/11: Detect SYNCWORD

Bit 1 **IRQCPOR**: IRQ flags clearing polarity selection
0: IRQ flags are cleared by writing 0 to the corresponding bits
1: IRQ flags are cleared by writing 1 to the corresponding bits

Bit 0 **IRQPOR**: IRQ signal polarity selection
0: Active low
1: Active high
When an IRQ flag in the IRQ3 register is set high and the corresponding IRQ function is enabled, the active level of the IRQ signal is determined by this configuration.

• **IRQ2: Interrupt Control Register 2**

Bit	7	6	5	4	3	2	1	0
Name	ARKTFIE	ATRCTIE	FIFOLTIE	RXERRIE	RXDETIE	CALCMPIE	RXCMPIE	TXCMPIE
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

Bit 7 **ARKTFIE**: ARK TX Failure IRQ Enable
0: Disable
1: Enable

Bit 6 **ATRCTIE**: ATR Cycle Timer IRQ Enable
0: Disable
1: Enable

Bit 5 **FIFOLTIE**: FIFO Low Threshold IRQ Enable
0: Disable
1: Enable

Bit 4 **RXERRIE**: RX Error IRQ Enable
0: Disable
1: Enable

Bit 3 **RXDETIE**: RX Event Detected IRQ Enable
0: Disable
1: Enable

Bit 2 **CALCMPIE**: Calibration Complete IRQ Enable
0: Disable
1: Enable

Bit 1 **RXCMPIE**: RX Complete IRQ Enable
0: Disable
1: Enable

Bit 0 **TXCMPIE**: TX Complete IRQ Enable
0: Disable
1: Enable

• **IRQ3: Interrupt Control Register 3**

Bit	7	6	5	4	3	2	1	0
Name	ARKTFIF	ATRCTIF	FIFOLTIF	RXERRIF	RXDETIF	CALCMPIF	RXCMPIF	TXCMPIF
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

When the individual flag within this register is set high by the hardware, the corresponding IRQ will be generated. These flags can be cleared by writing 0 or 1 to the corresponding flag which is determined by the IRQCPOR bit configuration.

Bit 7 **ARKTFIF**: ARK TX Failure IRQ Flag
0: No request
1: Interrupt request

Bit 6 **ATRCTIF**: ATR Cycle Timer IRQ Flag
0: No request
1: Interrupt request
This flag will be set high when the ATRCT timer is full.

Bit 5 **FIFOLTIF**: FIFO Low Threshold IRQ Flag
0: No request
1: Interrupt request
When in the Burst TX mode, if this flag is set high, it means that TX FIFO data length is less than FFMG setting threshold and there are TX data to be written into the FIFO. When in the Burst RX mode, if this flag is set high, it means that RX FIFO remaining space is less than FFMG setting threshold and the remaining RX data length is longer than FFMG setting threshold.

Bit 4 **RXERRIF**: RX Error IRQ Flag
0: No request
1: Interrupt request
The RX error conditions include RX failure, CRC failure (CRC_EN=1) or RX FIFO overwrite.

Bit 3 **RXDETIF**: RX Event Detected IRQ Flag
0: No request
1: Interrupt request
The RX events include carry and syncword, and the actual trigger source is determined by the RXDETS[1:0] configuration.

Bit 2 **CALCMPIF**: Calibration Complete IRQ Flag
0: No request
1: Interrupt request
If ACAL_EN=0, the LIRC calibration is enabled by its individual calibration enable bit and the calibration completion will trigger IRQ. If ACAL_EN=1, VCO and RC calibrations are enabled and both completion will trigger IRQ.

Bit 1 **RXCMPIF**: RX Complete IRQ Flag
0: No request
1: Interrupt request
When the RX operation is completed without any error, this flag will be set high by hardware.

Bit 0 **TXCMPIF**: TX Complete IRQ Flag
0: No request
1: Interrupt request

- **IO1: I/O Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	PADDS[1:0]		GIO2S[2:0]			GIO1S[2:0]		
R/W	R/W		R/W			R/W		
POR	0	1	0	0	0	0	0	0

Bit 7~6 **PADDS[1:0]**: PAD driving strength selection (only reset by POR)
00: 0.5mA
01: 1mA
10: 5mA
11: 10mA

Bit 5~3 **GIO2S[2:0]**: GIO2 pin function selection (only reset by POR)
000/111: No function, input
001: SDO, 4-wire SPI data, output
010: TRXD, direct mode TXD/RXD, input/output
011: TXD, direct mode TXD, input
100: RXD, direct mode RXD, output
101: IRQ, interrupt request, output
110: ROSCi, ATR clock external input

Bit 2~0 **GIO1S[2:0]**: GIO1 pin function selection (only reset by POR)
000/111: No function, input
001: SDO, 4-wire SPI data, output
010: TRXD, direct mode TXD/RXD, input/output
011: TXD, direct mode TXD, intput
100: RXD, direct mode RXD, output
101: IRQ, interrupt request, output
110: ROSCi, ATR clock external input

• **IO2: I/O Control Register 2**

Bit	7	6	5	4	3	2	1	0
Name	GIO4S[3:0]				—	—	—	—
R/W	R/W				—	—	—	—
POR	0	0	0	0	0	0	0	0

Bit 7~4 **GIO4S[3:0]**: GIO4 pin function selection (only reset by POR)
0000/0111: No function, input
0001: SDO, 4-wire SPI data, output
0010: TRXD, direct mode TXD/RXD, input/output
0011: TXD, direct mode TXD, input
0100: RXD, direct mode RXD, output
0101: IRQ, interrupt request, output
0110: ROSCi, ATR clock external input
1000: TBCLK, TX bit (data) clock, output
1001: RBCLK, RX bit (recovery) clock, output
1010: FSYCK, i.e. XCLK 1/1, 1/2, 1/4, 1/8 output
1011: LIRCCLK, internal LIRC clock with debounce, output
1100: EPA_EN, external PA enable, output
1101: ELAN_EN, external LNA enable, output
1110: TRBCLK, TBCLK in TX mode or RBCLK in RX mode, output
1111: PDB, Power down bar and XO enable, output

Bit 3~0 Reserved, must be "0000"

• **IO3: I/O Control Register 3**

Bit	7	6	5	4	3	2	1	0
Name	SDO_TEN	SPIPU	—	GIOPU4	—	GIOPU[2:1]		—
R/W	R/W	R/W	—	R/W	—	R/W		—
POR	0	1	1	1	1	1	1	1

Bit 7 **SDO_TEN**: SDO tri-state enable (only reset by POR)
0: Disable
1: Enable

Bit 6 **SPIPU**: 3-wire SPI pull-up enable (only reset by POR)
0: Disable
1: Enable
When this bit is set high, it only controls the pull-up function for the CSN, SCK and SDIO pins. Note that the pull-up function of the SDO pin for the 4-wire SPI is configured using the GIOPU4 and GIOPU[2:1] bits.

Bit 5 Reserved, must be "1"

Bit 4 **GIOPU4**: GIO pin function pull-up enable control (only reset by POR)
This bit controls the pull-high function of the GIO4 pin.

Bit 3 Reserved, must be "1"

Bit 2~1 **GIOPU[2:1]**: GIO pin function pull-up enable control (only reset by POR)
These bits control the pull-high function of the GIO2~GIO1 pins respectively.

Bit 0 Reserved, must be "1"

• **FIFO1: FIFO Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	—	—	TXFFSA[5:0]					
R/W	—	—	R/W					
Reset	0	0	0	0	0	0	0	0

Bit 7~6 Reserved, must be "00"

Bit 5~0 **TXFFSA[5:0]**: TX FIFO start address, used for Block FIFO mode

• **FIFO2: FIFO Control Regsiter 2**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	RXPL2F_EN	FFINF_EN	FFMG_EN	FFMG[1:0]	
R/W	—	—	—	R/W	R/W	R/W	R/W	
Reset	0	0	0	0	0	0	0	1

Bit 7~5 Reserved, must be "000"

Bit 4 **RXPL2F_EN**: RX payload length byte to FIFO enable
0: Disable
1: Enable
Setting this bit high will place the payload length byte in the packet to RX FIFO. In the RX continue mode (RXCON_EN=1), this bit should be set to 1 to support multiple payload in single RX FIFO.

Bit 3 **FFINF_EN**: FIFO infinite length mode enable
0: Disable
1: Enable

Bit 2 **FFMG_EN**: FIFO length margin detect enable
0: Disable
1: Enable

Bit 1~0 **FFMG[1:0]**: FIFO length margin selection
Threshold of remaining data in TX FIFO:
00: 4 bytes
01: 8 bytes
10: 16 bytes
11: 32 bytes
Threshold of remaining space in RX FIFO:
00: 4 bytes
01: 8 bytes
10: 16 bytes
11: 32 bytes
After the FIFO length margin detect function has been enabled by setting the FFMG_EN bit high and the required FIFO length margin has been selected by setting these bits, when the selected condition occurs the FIFOLTIF flag will be set high. In this case, an interrupt signal will also be generated if the corresponding interrrpt function has been enabled.

• **PKT1: Packet Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	TXPMLEN[7:0]							
R/W	R/W							
Reset	0	0	0	0	0	0	0	1

Bit 7~0 **TXPMLEN[7:0]**: TX preamble length
Transmit preamble length=(TXPMLEN[7:0]+1) words; the word length, 1 byte or 2 bytes, is further determined by the PMLP_EN and PMLPLEN settings.

• **PKT2: Packet Control Register 2**

Bit	7	6	5	4	3	2	1	0
Name	PID[1:0]		TRAILER_EN	WHTFMT[0]	SYNCLEN[1:0]		—	—
R/W	R/W		R/W	R/W	R/W		—	—
Reset	0	0	1	0	0	1	0	0

Bit 7~6 **PID[1:0]**: TX Packet ID
This ID will be placed in the highest two bits of the payload header field when the header option is enabled using the PLH_EN bit.

Bit 5 **TRAILER_EN**: Trailer field enable
0: Disable
1: Enable

Bit 4 **WHTFMT[0]**: Whitening format selection bit 0
WHTFMT[1:0]=
00: BC66F36xx, G(X)=$X^7+X^6+X^5+X^4+1$
01: PN7, G(X)=X^7+X^4+1
10: PN9-CCITT, G(X)=X^9+X^5+1
11: PN9-IBM, G(X)=X^9+X^5+1
The WHTFMT[1] bit is located in the PKT10 register.

Bit 3~2 **SYNCLEN[1:0]**: TX/RX mode SYNCWORD length setting bit 1~0
Final SYNCWORD length=({SYNCLEN[1:0], SYNCLENLB})+1; SYNCLENLB is located in the PKT10 register.

Bit 1~0 Reserved, must be "00"

• **PKT3: Packet Control Register 3**

Bit	7	6	5	4	3	2	1	0
Name	MCH_EN	FEC_EN	CRC_EN	CRCFMT	PLLEN_EN	PLHAC_EN	PLHLEN	PLH_EN
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	1	0	0	0	0	0

Bit 7 **MCH_EN**: Manchester code enable
0: Disable
1: Enable

Bit 6 **FEC_EN**: FEC enable
0: Disable
1: Enable

Bit 5 **CRC_EN**: CRC field enable
0: Disable
1: Enable

Bit 4 **CRCFMT**: CRC format selection
0: CCITT-16-CRC, G(X)=$X^{16}+X^{12}+X^5+1$
1: IBC-16-CRC, G(X)=$X^{16}+X^{15}+X^2+1$

Bit 3 **PLLEN_EN**: Payload length field enable
0: Disable
1: Enable

Bit 2 **PLHAC_EN**: Payload header address correction enable control
0: Disable, PLHA[5:0] and PLHEA[7:0] can be used as software flags defined by users.
1: Enable, PLHA[5:0] and PLHEA[7:0] of TX/RX devices must include the same address, otherwise the packet will be regarded as a failed packet.
Note: Whether the PLHEA[7:0] is included or not is determined by the PLHLEN bit setting.

Bit 1 **PLHLEN**: Payload header length
0: 1 byte
1: 2 bytes

Bit 0 **PLH_EN**: Payload header field enable
0: Disable
1: Enable

• **PKT4: Packet Control Register 4**

Bit	7	6	5	4	3	2	1	0
Name	WHT_EN	WHTSD[6:0]						
R/W	R/W	R/W						
Reset	0	0	1	1	0	1	1	0

Bit 7 **WHT_EN**: Data whitening enable
0: Disable
1: Enable

Bit 6~0 **WHTSD[6:0]**: Data whitening seed bit 6~0
WHTSD[8:7] is located in the PKT15 register.

• **PKT5: Packet Control Register 5**

Bit	7	6	5	4	3	2	1	0
Name	TXDLEN[7:0]							
R/W	R/W							
Reset	0	1	0	0	0	0	0	0

Bit 7~0 **TXDLEN[7:0]**: TX data length (unit: byte, used in burst mode only)

• **PKT6: Packet Control Register 6**

Bit	7	6	5	4	3	2	1	0
Name	RXDLEN[7:0]							
R/W	R/W							
Reset	0	1	0	0	0	0	0	0

Bit 7~0 **RXDLEN[7:0]**: RX data length (unit: byte; used in burst mode only)
When the PLLEN_EN bit is cleared to 0, the received data length is determined by this field.

When this register is read, the read value indicates the RX data length in FIFO. The default read value is 00h.

• **PKT7: Packet Control Register 7**

Bit	7	6	5	4	3	2	1	0
Name	RXPID[1:0]		DLY_RXS[2:0]			DLY_TXS[2:0]		
R/W	R		R/W			R/W		
Reset	0	0	1	0	0	0	0	0

Bit 7~6 **RXPID[1:0]**: Received packet PID (read only)

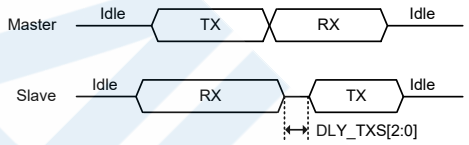
Bit 5~3 **DLY_RXS[2:0]**: RX block stable time after RX is enabled
000: 4μs
001: 8μs
010: 12μs
011: 16μs
100: 20μs
101: 32μs

110: 64μs
111: 100μs

These bits are used to select the waiting time between RX enable and RX stable. This time should be configured to a value greater than the default RX DCOC turbo mode delay time of 6μs.

Bit 2~0 **DLY_TXS[2:0]**: TX start (delay) time before entering the TX mode
000: 0μs
001: 10μs
010: 20μs
011: 40μs
100: 60μs
101: 80μs
110: 100μs
111: 120μs

It is used to align the timing between transmitter and receiver in ARK mode.



• **PKT8: Packet Control Register 8**

Bit	7	6	5	4	3	2	1	0
Name	—	—	PLHA[5:0]					
R/W	—	—	R/W					
Reset	0	0	0	0	0	0	0	0

Bit 7~6 Reserved, must be "00"

Bit 5~0 **PLHA[5:0]**: Payload header address to support broadcast
Address=0 in RX mode means not doing address correction check.
Write: write data to TX PLHA[5:0]. Read: read data from RX PLHA[5:0].

• **PKT9: Packet Control Register 9**

Bit	7	6	5	4	3	2	1	0
Name	PLHEA[7:0]							
R/W	R/W							
Reset	0	0	0	0	0	0	0	0

Bit 7~0 **PLHEA[7:0]**: Payload header extended address to support broadcast
Address=0 in RX mode means not doing address correction check.

• **MOD1: Modulator Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	DTR[7:0]							
R/W	R/W							
Reset	0	0	0	0	0	0	0	1

Bit 7~0 **DTR[7:0]**
DTR[8:0]: Data rate divider, DTR[8] is loaced in the MOD2 register.
Data Rate=f_{XTAL}/[32×(DTR[8:0]+1)], here data rate indicates TBCLK. Note that DTR[8:0] can only be an odd number.

• **MOD2: Modulator Control Register 2**

Bit	7	6	5	4	3	2	1	0
Name	RXFDOS[11:8]				DITHER[1:0]		—	DTR[8]
R/W	R/W				R/W		—	R/W
Reset	1	0	0	1	0	0	0	0

Bit 7~4 **RXFDOS[11:8]**

RXFDOS[11:0]: RX frequency diavation offset, RXFDOS[7:0] is located in the MOD3 register.

Write to RXFDOS[11:8] first and then write to RXFDOS[7:0] to fully update RXFDOS[11:0].

RXFDOS[11:0]=Floor{(f_{IF}/f_{XTAL})×2^{17}}

Bit 3~2 **DITHER[1:0]**: Dither value

Bit 1 Reserved, must be "0"

Bit 0 **DTR[8]**:

DTR[8:0]: Data rate divider, DTR[7:0] is loaced in the MOD1 register.

Data Rate=f_{XTAL}/[32×(DTR[8:0]+1)], here data rate indicates TBCLK. Note that DTR[8:0] can only be an odd number.

• **MOD3: Modulator Control Register 3**

Bit	7	6	5	4	3	2	1	0
Name	RXFDOS[7:0]							
R/W	R/W							
Reset	1	0	0	1	1	0	1	0

Bit 7~0 **RXFDOS[7:0]**

RXFDOS[11:0]: RX frequency diavation offset, RXFDOS[11:8] is loaced in the MOD2 register.

Write to RXFDOS[11:8] first and then write to RXFDOS[7:0] to fully update RXFDOS[11:0].

RXFDOS[11:0]=Floor{(f_{IF}/f_{XTAL})×2^{17}}

• **DM1: Demodulator Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	—	—	MDIV[5:0]					
R/W	—	—	R/W					
Reset	0	0	0	0	0	0	1	1

Bit 7~6 Reserved, must be "00"

Bit 5~0 **MDIV[5:0]**: Demodulator operation clock divider

DMCLK=ADCLK/(MDIV[5:0]+1)

• **DM2: Demodulator Control Register 2**

Bit	7	6	5	4	3	2	1	0
Name	—	—	SDR[5:0]					
R/W	—	—	R/W					
Reset	0	0	0	0	0	0	0	0

Bit 7~6 Reserved, must be "00"

Bit 5~0 **SDR[5:0]**: Decimator operation clock after phase extract

SDR[5:0]+1=DMCLK/(8×DATA_RATE), here DATA_RATE indicates RBCLK.

• **DM3: Demodulator Control Register 3**

Bit	7	6	5	4	3	2	1	0
Name	CSF_SW_EN	FD_MOD[6:0]						
R/W	R/W	R/W						
Reset	1	1	1	0	0	0	0	0

Bit 7 **CSF_SW_EN**: Channel selection filter auto bandwidth switch enable
0: Disable
1: Enable

Bit 6~0 **FD_MOD[6:0]**: Frequency deviation modifier
FD_MOD=Round((h/(SDR[5:0]+1))×128); h=modulation index
SDR[5:0]+1=DMCLK/(8×DATA_RATE)

• **DM4: Demodulator Control Register 4**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	CFO_DSEL	—	PH_DIFF_MOD	—
R/W	—	—	—	—	R/W	—	R/W	—
Reset	0	0	0	0	1	0	0	0

Bit 7~4 Reserved, must be "0000"

Bit 3 **CFO_DSEL**: CFO correction domain selection
0: Analog domain
1: Digital domain

Bit 2 Reserved, must be "0"

Bit 1 **PH_DIFF_MOD**: Phase difference extract mode setting
0: Phase extract range [-pi/2, pi/2]
1: Phase extract range [-pi, pi]

Bit 0 Reserved, must be "0"

• **DM5: Demodulator Control Register 5**

Bit	7	6	5	4	3	2	1	0
Name	FD_HOLD[7:0]							
R/W	R/W							
Reset	0	0	1	1	0	0	0	0

Bit 7~0 **FD_HOLD[7:0]**: Frequency deviation threshold for "preamble + SYNCWORD" detection

• **DM8: Demodulator Control Register 8**

Bit	7	6	5	4	3	2	1	0
Name	M_RATIO[7:0]							
R/W	R/W							
Reset	0	1	0	0	0	0	0	0

Bit 7~0 **M_RATIO[7:0]**: For CFO calulation
M_RATIO=round(1/(MDIV[5:0]+1)×2^8)

Bank 0 Control Registers

All control registers will be set to their initial value by power-on reset (POR). The software reset will set the control registers to their initial value except the LIRC_EN and LIRC_OW bits in the XO3 register. These bits keep unchanged after software reset.

Addr.	Name	Bit							
		7	6	5	4	3	2	1	0
20h	OM	PWR_SOFT	BAND_SEL[1:0]		—	ACAL_EN	RTX_EN	RTX_SEL	SX_EN
22h	SX1	—	D_N[6:0]						
23h	SX2	D_K[7:0]							
24h	SX3	D_K[15:8]							
25h	SX4	—	—	—	—	D_K[19:16]			
26h	STA1	—	—	—	CD_FLAG	—	OMST[2:0]		
28h	RSSI2	—				RSSI_CTHD[3:0]			
29h	RSSI3	RSSI_NEGDB[7:0]							
2Ah	RSSI4	RSSI_SYNC_OK[7:0]							
2Bh	ATR1	ATRCLK_DIV[1:0]		ATRCLKS	ATRTU	ATRCTM	ATRM[1:0]		ATR_EN
2Ch	ATR2	ATRCYC[7:0]							
2Dh	ATR3	ATRCYC[15:8]							
2Eh	ATR4	ATRRXAP[7:0]							
2Fh	ATR5	ATRRXEP[7:0]							
30h	ATR6	ATRRXEP[15:8]							
31h	ATR7	ARKNM[3:0]				—	ATR_WDLY[1:0]		ARK_EN
32h	ATR8	ARKRXAP[7:0]							
33h	ATR9	ATRCT[7:0]							
34h	ATR10	ATRCT[15:8]							
35h	ATR11	ATRCYCM	—				ATRRXAP[10:8]		
36h	PTK10	—	WHTFMT[1]	CRCBYTEO	CRCBITO	CRCINV	SYNCLENLB	PMLPLEN	PMLP_EN
37h	PTK11	PMLPAT[7:0]							
38h	PTK12	PMLPAT[15:8]							
39h	PTK13	CRCSD[7:0]							
3Ah	PTK14	CRCSD[15:8]							
3Bh	PTK15	WHTSD[8:7]		OOKDT_TS[3:0]				OOKDT_POR	OOKDT_EN
3Ch	XO1	XSHIFT[1:0]		—	XO_TRIM[4:0]				
3Dh	XO2	—			XO_SW	—			
3Eh	XO3	LIRCCAL_EN	LIRC_OW	LIRC_OP[4:0]					LIRC_EN
3Fh	TX2	—	CT_PAD[5:0]						

Note: The addresses which are not listed in this table are reserved for future use, it is suggested not to change their initial values by any methods.

The reset value shown in the following register description tables means the software reset results of strobe command.

- **OM: Operation Mode Control Register**

Bit	7	6	5	4	3	2	1	0
Name	PWR_SOFT	BAND_SEL[1:0]		—	ACAL_EN	RTX_EN	RTX_SEL	SX_EN
R/W	R/W	R/W		—	R/W	R/W	R/W	R/W
Reset	0	0	1	0	0	0	0	0

Bit 7 **PWR_SOFT**: RF operation mode selection
0: RF normal operation mode
1: RF engineering mode

Bit 6~5 **BAND_SEL[1:0]**: Band selection (when PWR_SOFT=0)
00: 315MHz band, ODDIV=4
01: 433MHz band, ODDIV=2
10: 470~510MHz band, ODDIV=2
11: 868/915MHz band, ODDIV=1

Bit 4 Reserved, must be "0"

Bit 3 **ACAL_EN**: Auto calibration enable
0: Disable
1: Enable
When this bit is set high, both the VCO and RC calibrations will be enabled. When the VCO and RC calibrations are completed, this bit will be cleared to zero by hardware.

Bit 2 **RTX_EN**: RX or TX mode enable
0: Disable
1: Enable
After the RX or TX mode has been selected by the RTX_SEL bit, setting this bit high will enable the selected mode.

Bit 1 **RTX_SEL**: RX or TX mode selection
0: RX mode
1: TX mode

Bit 0 **SX_EN**: Synthesizer enable (standby mode enable control)
0: Disable
1: Enable
Setting this bit high will enable the PFD, CP and VCO functions.

- **SX1: Fractional-N Synthesizer Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	—	D_N[6:0]						
R/W	—	R/W						
Reset	0	0	1	1	0	1	1	0

Bit 7 Reserved, must be "0"

Bit 6~0 **D_N[6:0]**: RF channel integer number code
D_N[6:0]=floor(f_{RF}×ODDIV/f_{XTAL})
For example, XO=16MHz and RF band=433.92MHz which are initial setup:
→ (433.92MHz×2)/16MHz=54.24
→ D_N=54
→ Dec2Hex(54)=36
→ Dec2Bin(54)=011_0110

- **SX2: Fractional-N Synthesizer Control Register 2**

Bit	7	6	5	4	3	2	1	0
Name	D_K[7:0]							
R/W	R/W							
Reset	0	0	0	0	1	0	1	0

Bit 7~0 **D_K[7:0]**: RF channel fractional number code lowest byte

- **SX3: Fractional-N Synthesizer Control Register 3**

Bit	7	6	5	4	3	2	1	0
Name	D_K[15:8]							
R/W	R/W							
Reset	1	1	0	1	0	1	1	1

Bit 7~0 **D_K[15:8]**: RF channel fractional number code medium byte

• **SX4: Fractional-N Synthesizer Control Register 4**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	D_K[19:16]			
R/W	—	—	—	—	R/W			
Reset	0	0	0	0	0	0	1	1

Bit 7~4 Reserved, must be "0000"

Bit 3~0 **D_K[19:16]**: RF channel fractional number code highest byte

D_K[19:0]=floor{(f_{RF}×ODDIV/f_{XTAL}-D_N[6:0])×2^{20}}

For example, XO=16MHz and RF band=433.92MHz which are initial setup:

→ (433.92MHz×2)/16MHz=54.24

→ D_K=0.24×220=251658

→ Dec2Hex(251658)=3D70A

→ Dec2Bin(251658)=0011_1101+0111_0000_1010

• **STA1: Status Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	CD_FLAG	—	OMST[2:0]		
R/W	—	—	—	R	—	R		
Reset	0	0	0	0	0	0	0	0

Bit 7~5 Reserved, must be "000"

Bit 4 **CD_FLAG**: Carrier detection flag (read only)

This flag will be set high by hardware when carrier detection is okey after pulling DEMOD_EN high. Here DEMOD_EN high level is an internal signal which is generated by the internal state machine when in the Direct mode (DIR_EN=1) or after the RX strobe command is received when in the Burst mode (DIR_EN=0). The flag will be automatically cleared when RX_EN rising edge occurs. Here RX_EN rising edge is generated after setting RTX_SEL=0 and RTX_EN=1 when in the Direct mode or by the internal state machine after the RX strobe command is received when in the Burst mode.

Bit 3 Reserved, must be "0"

Bit 2~0 **OMST[2:0]**: Operation mode state indication (read only)

000: Deep Sleep mode

001: Idle mode

010: Light Sleep mode

011: Standby mode

100: TX mode

101: RX mode

110: VCO calibration mode

111: Undefined

• **RSSI2: RSSI Control Register 2**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	RSSI_CTHD[3:0]			
R/W	—	—	—	—	R/W			
Reset	0	0	0	0	1	0	1	0

Bit 7~4 Reserved, must be "0000"

Bit 3~0 **RSSI_CTHD[3:0]**: RSSI threshold for carrier detection

(RSSI_CTHD[3:0]×2+1)+74=RSSI threshold for carrier detection

• **RSSI3: RSSI Control Register 3**

Bit	7	6	5	4	3	2	1	0
Name	RSSI_NEGDB[7:0]							
R/W	R							
Reset	0	0	0	0	0	0	0	0

Bit 7~0 **RSSI_NEGDB[7:0]**: RSSI value (unit: -dB)
It is a real time measurement value.

• **RSSI4: RSSI Control Register 4**

Bit	7	6	5	4	3	2	1	0
Name	RSSI_SYNC_OK[7:0]							
R/W	R							
Reset	0	0	0	0	0	0	0	0

Bit 7~0 **RSSI_SYNC_OK[7:0]**: RSSI snapshot when SYNCWORD is detected correct

• **ATR1: Auto TX/RX Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	ATRCLK_DIV[1:0]		ATRCLKS	ATRTU	ATRCTM	ATRM[1:0]		ATR_EN
R/W	R/W		R/W	R/W	R/W	R/W		R/W
Reset	1	1	0	0	1	0	0	0

Bit 7~6 **ATRCLK_DIV[1:0]**: ATR clock frequency division selection
00: 1/1, ATRCLK=32768Hz
01: 1/4, ATRCLK=8192Hz
10: 1/8, ATRCLK=4096Hz
11: 1/16, ATRCLK=2048Hz

Bit 5 **ATRCLKS**: ATRCLK clock source selection
0: From the internal LIRC clock
1: From the exteral ROSCi clock input on the GIOn pin

Bit 4 **ATRTU**: Auto TRX unit time selection
0: 250μs
1: 1ms, used to support low data rate applications
This bit is used to select the unit time for the ATR RX active period (ATRRXAP[10:0]), ATR RX extended period (ATRRXEP[15:0]) and ARK RX active period (ARKRXAP[7:0]).

Bit 3 **ATRCTM**: Auto TRX timer mode selection
0: Single mode, restart ATRCT timer when every ATR transaction occurs.
1: Continuous mode, start ATRCT timer upon receiving Idle command; stop ATRCT timer when ATR_EN=0 or ATRCTM=0 and exit the ATR active period.

Bit 2~1 **ATRM[1:0]**: Auto TRX mode selection
00: ATR WOT mode
01: ATR WOR mode
10/11: ATR WTM mode

Bit 0 **ATR_EN**: Auto TRX enable
0: Disable
1: Enable
Note that the ATR functions are activated by operation state transition from Deep Sleep/Light Sleep mode to Idle mode.

• **ATR2: Auto TX/RX Control Register 2**

Bit	7	6	5	4	3	2	1	0
Name	ATRCYC[7:0]							
R/W	R/W							
Reset	1	1	1	1	1	1	1	1

Bit 7~0 **ATRCYC[7:0]**: ATRCT timer expire value low byte

• **ATR3: Auto TX/RX Control Register 3**

Bit	7	6	5	4	3	2	1	0
Name	ATRCYC[15:8]							
R/W	R/W							
Reset	0	0	0	0	1	1	1	1

Bit 7~0 **ATRCYC[15:8]**: ATRCT timer expire value high byte

ATRCYCM=0 – BC66F36x2 compatible mode (default):

Wake-up period=ATRCLK period×ATRCYC[15:0]+LIRCCLK period, ATRCYC[15:0]≠0. Default wake-up period is 2 seconds.

ATRCYCM=1 – BC66F36x3 compatible mode:

Wake-up period=ATRCLK period×(ATRCYC[15:0]+1), ATRCYC[15:0]≠0. Default wake-up period is 2 seconds.

• **ATR4: Auto TX/RX Control Register 4**

Bit	7	6	5	4	3	2	1	0
Name	ATRRXAP[7:0]							
R/W	R/W							
Reset	0	0	1	0	0	1	1	1

Bit 7~0 **ATRRXAP[7:0]**: ATR RX active period low byte

ATR RX active period high byte ATRRXAP[10:8] is located in the ATR11 register.

Active period=unit time×(ATRRXAP[10:0]+1); the unit time can be 250μs or 1ms which is determined by the ATRTU bit. The default ATR RX active period is 10ms with a default time unit of 250μs.

• **ATR5: Auto TX/RX Control Register 5**

Bit	7	6	5	4	3	2	1	0
Name	ATRRXEP[7:0]							
R/W	R/W							
Reset	1	0	0	0	1	1	1	1

Bit 7~0 **ATRRXEP[7:0]**: ATR RX extend period low byte

• **ATR6: Auto TX/RX Control Register 6**

Bit	7	6	5	4	3	2	1	0
Name	ATRRXEP[15:8]							
R/W	R/W							
Reset	0	0	0	0	0	0	0	1

Bit 7~0 **ATRRXEP[15:8]**: ATR RX extended period high byte

Extended period=unit time×(ATRRXEP[15:0]+1); the unit time can be 250μs or 1ms which is determined by the ATRTU bit. The default ATR RX extended period is 100ms with a default time unit of 250μs.

• **ATR7: Auto TX/RX Control Register 7**

Bit	7	6	5	4	3	2	1	0
Name	ARKNM[3:0]				—	ATR_WDLY[1:0]		ARK_EN
R/W	R/W				—	R/W		R/W
Reset	0	1	1	1	0	0	1	0

Bit 7~4 **ARKNM[3:0]**: ARK repeat cycle number
Maximum repeat cycle number=ARKNM[3:0]+1

Bit 3 Reserved, must be "0"

Bit 2~1 **ATR_WDLY[1:0]**: Auto wake-up delay time
00: 244μs
01: 488μs
10: 732μs
11: 976μs

Bit 0 **ARK_EN**: Auto-Resend/ACK enable
0: Disable
1: Enable

• **ATR8: Auto TX/RX Control Register 8**

Bit	7	6	5	4	3	2	1	0
Name	ARKRXAP[7:0]							
R/W	R/W							
Reset	0	0	1	0	0	1	1	1

Bit 7~0 **ARKRXAP[7:0]**: ARK RX active period
Active period=unit time×(ARKRXAP[7:0]+1); the unit time can be 250μs or 1ms which is determined by the ATRTU bit. The default ARK RX active period is 10ms with a default time unit of 250μs.

• **ATR9: Auto TX/RX Control Register 9**

Bit	7	6	5	4	3	2	1	0
Name	ATRCT[7:0]							
R/W	R/W							
Reset	0	0	0	0	0	0	0	0

Bit 7~0 **ATRCT[7:0]**: ATR cycle timer low byte

• **ATR10: Auto TX/RX Control Register 10**

Bit	7	6	5	4	3	2	1	0
Name	ATRCT[15:8]							
R/W	R/W							
Reset	0	0	0	0	0	0	0	0

Bit 7~0 **ATRCT[15:8]**: ATR cycle timer high byte
Reading ATRCT[15:0] will get the current counting value. Due to the limitation of SPI 8-bit data length, reading the ATR9 register will take a snapshot of the whole 16-bit data into the read register buffer. Users should read ATR9 and ATR10 continuously (non-interrupted) to get correct data.
Writing to ATRCT[15:0] will update the counting value. Write to ATR9 first and then write to ATR10 to trigger the ATRCT write function. This timer update mechanism is used to align the time slot for the master and slave in a two-way RF system.

• **ATR11: Auto TX/RX Control Register 11**

Bit	7	6	5	4	3	2	1	0
Name	ATRCYCM	—	—	—	—	ATRRXAP[10:8]		
R/W	R/W	—	—	—	—	R/W		
Reset	0	0	0	0	0	0	0	0

Bit 7 **ATRCYCM**: ATR cycle calculation mode
0: BC66F36x2 compatible mode
1: BC66F36x3 compatible mode
Refer to the ATR2 and ATR3 registers for more cycle calculation details.

Bit 6~3 Reserved, must be "0000"

Bit 2~0 **ATRRXAP[10:8]**: ATR RX active period high byte
ATR RX active period low byte ATRRXAP[7:0] is located in the ATR4 register.
Active period=unit time×(ATRRXAP[10:0]+1); the unit time can be 250μs or 1ms which is determined by the ATRTU bit. The default ATR RX active period is 10ms with a default time unit of 250μs.

• **PKT10: Packet Control Register 10**

Bit	7	6	5	4	3	2	1	0
Name	—	WHTFMT[1]	CRCBYTEO	CRCBITO	CRCINV	SYNCLENLB	PMLPLEN	PMLP_EN
R/W	—	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	1	0	0

Bit 7 Reserved, must be "0"

Bit 6 **WHTFMT[1]**: Whitening format selection bit 1
WHTFMT[1:0]=
00: BC360X, $G(X)=X^7+X^6+X^5+X^4+1$
01: PN7, $G(X)=X^7+X^4+1$
10: PN9-CCITT, $G(X)=X^9+X^5+1$
11: PN9-IBM, $G(X)=X^9+X^5+1$
The WHTFMT[0] bit is located in the PKT2 register.

Bit 5 **CRCBYTEO**: CRC byte order
0: CRC high byte first
1: CRC low byte first

Bit 4 **CRCBITO**: CRC bit order
0: CRC high bit first
1: CRC low bit first

Bit 3 **CRCINV**: CRC bit inverse function control
0: Non-inverse
1: Inverse

Bit 2 **SYNCLENLB**: TX/RX mode SYNCWORD length setting lowest bit
Final SYNCWORD length=({SYNCLEN[1:0], SYNCLENLB)+1; SYNCLEN[1:0] is located in the PKT2 register.

Bit 1 **PMLPLEN**: Preamble pattern length selection (when PMLP_EN=1)
0: 1 byte – PMLPAT[7:0]
1: 2 bytes – PMLPAT[15:0], low byte first

Bit 0 **PMLP_EN**: Preamble pattern selection
0: Preamble pattern is derived from "SYNCWORD MSB+1/0 toggle"
Ex: SYNCWORD MSB=0, preamble=01010101(b)
SYNCWORD MSB=1, preamble=10101010(b)
1: Preamble pattern is derived from PMLPAT[15:0]

• **PKT11: Packet Control Register 11**

Bit	7	6	5	4	3	2	1	0
Name	PMLPAT[7:0]							
R/W	R/W							
Reset	0	1	0	1	0	1	0	1

Bit 7~0 **PMLPAT[7:0]**: Preamble pattern low byte

• **PKT12: Packet Control Register 12**

Bit	7	6	5	4	3	2	1	0
Name	PMLPAT[15:8]							
R/W	R/W							
Reset	0	1	0	1	0	1	0	1

Bit 7~0 **PMLPAT[15:8]**: Preamble pattern high byte

• **PKT13: Packet Control Register 13**

Bit	7	6	5	4	3	2	1	0
Name	CRCSD[7:0]							
R/W	R/W							
Reset	1	1	1	1	1	1	1	1

Bit 7~0 **CRCSD[7:0]**: CRC seed low byte

• **PKT14: Packet Control Register 14**

Bit	7	6	5	4	3	2	1	0
Name	CRCSD[15:8]							
R/W	R/W							
Reset	1	1	1	1	1	1	1	1

Bit 7~0 **CRCSD[15:8]**: CRC seed high byte

• **PKT15: Packet Control Register 15**

Bit	7	6	5	4	3	2	1	0
Name	WHTSD[8:7]		OOKDT_TS[3:0]				OOKDT_POR	OOKDT_EN
R/W	R/W		R/W				R/W	R/W
Reset	0	0	0	1	0	1	0	1

Bit 7~6 **WHTSD[8:7]**: Data whitening seed bit 8~7
WHTSD[6:0] is located in the PKT4 register. Note that only PN9-CCIT and PN9-IBM data whitening polynomials use the entire WHTSD[8:0] 9 bits.

Bit 5~2 **OOKDT_TS[3:0]**: OOK duty cycle tuning time selection
0000: 2μs
0001: 4μs
0010: 6μs
0011: 8μs
....(step: 2μs)
1100: 26μs
1101: 28μs
1110: 30μs
1111: 32μs

Bit 1 **OOKDT_POR**: OOK duty cycle tuning polarity
0: Extend 0 duty cycle
1: Extend 1 duty cycle

Bit 0 **OOKDT_EN**: OOK duty cycle tuning enable
0: Disable
1: Enable

• **XO1: XO Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	XSHIFT[1:0]		—	XO_TRIM[4:0]				
R/W	R/W		—	R/W				
Reset	0	0	0	1	0	0	0	0

Bit 7~6 **XSHIFT[1:0]**: Coarse tune of XO load capacitor for different crystal C_{LOAD}
Bit 5 Reserved, must be set to "1"
Bit 4~0 **XO_TRIM[4:0]**: Fine tune of XO load capacitor

• **XO2: XO Control Register 2**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	XO_SW	—	—	—	—
R/W	—	—	—	R/W	—	—	—	—
Reset	0	0	0	0	0	0	1	1

Bit 7~5 Reserved, must be "000"
Bit 4 **XO_SW**: Crystal oscillator load capacitance switch
0: About 12~16pF
1: About 8~12pF
Bit 3~0 Reserved, must be "0011"

• **XO3: XO Control Register 3**

Bit	7	6	5	4	3	2	1	0
Name	LIRCCAL_EN	LIRC_OW	LIRC_OP[4:0]					LIRC_EN
R/W	R/W	R/W	R/W					R/W
POR	—	0	—	—	—	—	—	0
Reset	0	—	0	1	1	0	1	—

Bit 7 **LIRCCAL_EN**: LIRC calibration enable
0: Disable
1: Enable
Bit 6 **LIRC_OW**: LIRC overwrite control (only reset by POR)
0: LIRC_OP[4:0] from calibration engine
1: LIRC_OP[4:0] from control register
Bit 5~1 **LIRC_OP[4:0]**: LIRC trim
After writing data to LIRC_OP[4:0], this trim will become active when the LIRC_OW bit is set high. When reading data from LIRC_OP[4:0], the actual data source is determined by the LIRC_OW bit setting.
Bit 0 **LIRC_EN**: LIRC enable (only reset by POR)
0: Disable
1: Enable

• **TX2: TX Control Register 2**

Bit	7	6	5	4	3	2	1	0
Name	—	—	CT_PAD[5:0]					
R/W	—	—	R/W					
Reset	0	0	1	1	0	0	0	0

Bit 7~6 Reserved, must be "00"
Bit 5~0 **CT_PAD[5:0]**: TX PAD linear power control

Bank 1 Control Registers

All control registers will be set to their initial value by power-on reset (POR). The software reset will set the control registers to their initial value.

Addr.	Name	Bit							
		7	6	5	4	3	2	1	0
20h	AGC1	MPT_0DB_EN	—	OFFSET_HBSEL[2:0]			OFFSET_LBSEL[2:0]		
21h	AGC2	—			DBFS_OFFSET[2:0]			AGC_CMP_THD[1:0]	
22h	AGC3	—					IF_DETOK_THD[2:0]		
23h	AGC4	GAIN_SEL[3:0]				—	AGC_ST[2:0]		
26h	AGC7	GAIN_STB[7:0]							
2Ch	FCF1	—	—	SFRATIO[1:0]		GFD_COMP[3:0]			
2Dh	FCF2	FSCALE[7:0]							
2Eh	FCF3	—				FSCALE[11:8]			
2Fh	FCF4	CF_B12[7:0]							
30h	FCF5	—						CF_B12[9:8]	
31h	FCF6	CF_B13[7:0]							
32h	FCF7	—						CF_B13[9:8]	
33h	FCF8	CF_A12[7:0]							
34h	FCF9	—						CF_A12[9:8]	
35h	FCF10	CF_A13[7:0]							
36h	FCF11	—						CF_A13[9:8]	
37h	FCF12	CF_B22[7:0]							
38h	FCF13	—						CF_B22[9:8]	
39h	FCF14	CF_B23[7:0]							
3Ah	FCF15	—						CF_B23[9:8]	
3Bh	FCF16	CF_A22[7:0]							
3Ch	FCF17	—						CF_A22[9:8]	
3Dh	FCF18	CF_A23[7:0]							
3Eh	FCF19	—						CF_A23[9:8]	

Note: The addresses which are not listed in this table are reserved for future use, it is suggested not to change their initial values by any methods.

The reset value shown in the following register description tables means the software reset results of strobe command.

- **AGC1: AGC Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	MPT_0DB_EN	—	OFFSET_HBSEL[2:0]			OFFSET_LBSEL[2:0]		
R/W	R/W	—	R/W			R/W		
Reset	0	0	0	1	1	0	0	0

Bit 7 **MPT_0DB_EN**: Force Curve1 max point=0dB
0: Disable
1: Enable

Bit 6 Reserved, must be "0"

Bit 5~3 **OFFSET_HBSEL[2:0]**: AGC max point selection
max point=-6+2×offset_HB

Bit 2~0 **OFFSET_LBSEL[2:0]**: AGC min point select
min point=max point - 22+2×offset_LB

OFFSET_HBSEL[2:0]	offset_HB
000	0
001	1
010	2
011	3
100	-4
101	-3
110	-2
111	-1

OFFSET_LBSEL[2:0]	offset_LB
000	0
001	1
010	2
011	3
100	-4
101	-3
110	-2
111	-1

Ex: OFFSET_HBSEL[2:0]=011, offset_HB=+3; OFFSET_LBSEL[2:0]=000, offset_LB=+0;
AGC max point=-6+2×3=0dBFs; AGC min point=0 - 22+2×0=-22dBFs

• AGC2: AGC Control Register 2

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	DBFS_OFFSET[2:0]			AGC_CMP_THD[1:0]	
R/W	—	—	—	R/W			R/W	
Reset	0	0	0	1	0	1	0	0

Bit 7~5 Reserved, must be "000"
Bit 4~2 **DBFS_OFFSET[2:0]**: Log offset for GFSK
Bit 1~0 **AGC_CMP_THD[1:0]**: AGC comparison number threshold
00: Continuous AGC comparison until SYNCWORD is detected
01~11: Comparison number threshold

• AGC3: AGC Control Register 3

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	IF_DETOK_THD[2:0]		
R/W	—	—	—	—	—	R/W		
Reset	0	0	0	0	0	1	0	0

Bit 7~3 Reserved, must be "00000"
Bit 2~0 **IF_DETOK_THD[2:0]**: IF detection OK threshold
After the gain stable time which determined by the AGC7 register, the AGC circuit will wait for (IF_DETOK_THD[2:0]×8) ADCLK cycles before starting to detect the IF signal strength.

• AGC4: AGC Control Register 4

Bit	7	6	5	4	3	2	1	0
Name	GAIN_SEL[3:0]				—	AGC_ST[2:0]		
R/W	R				—	R		
Reset	0	0	0	1	0	0	0	1

Bit 7~4 **GAIN_SEL[3:0]**: Gain curve selection
Bit 3 Reserved, must be "0"
Bit 2~0 **AGC_ST[2:0]**: AGC state machine state
100: AGC is completed

• **AGC7: AGC Control Register 7**

Bit	7	6	5	4	3	2	1	0
Name	GAIN_STB[7:0]							
R/W	R/W							
Reset	0	0	1	1	0	0	0	0

Bit 7~0 **GAIN_STB[7:0]**: Gain stable count
Gain stable count delay in ADCLK period=GAIN_STB[7:0]×2

• **FCF1: Filter Coefficient Control Register 1**

Bit	7	6	5	4	3	2	1	0
Name	—	—	SFRATIO[1:0]		GFD_COMP[3:0]			
R/W	—	—	R/W		R/W			
Reset	0	0	0	0	0	1	1	0

Bit 7~6 Reserved, must be "00"

Bit 5~4 **SFRATIO[1:0]**: Smooth filter ratio selection
00: 1/1
01: 1/16
10: 1/64
11: 1/128

Bit 3~0 **GFD_COMP[3:0]**: Compensate gaussian filter for 101/010 pattern
[3]: 1/8; [2]: 1/16; [1]: 1/32; [0]: 1/64
For example:
0000: no compensation
….
1111: value after compensation=(1+0.234375)x=[1+(1/8+1/16+1/32+1/64)]x

• **FCF2: Filter Coefficient Control Register 2**

Bit	7	6	5	4	3	2	1	0
Name	FSCALE[7:0]							
R/W	R/W							
Reset	0	1	0	0	0	1	0	0

Bit 7~0 **FSCALE[7:0]**: Frequency deviation scale parameter low byte

• **FCF3: Filter Coefficient Control Register 3**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	FSCALE[11:8]			
R/W	—	—	—	—	R/W			
Reset	0	0	0	0	0	1	0	0

Bit 7~4 Reserved, must be "0000"

Bit 3~0 **FSCALE[11:8]**: Frequency deviation scale parameter high byte
If the data rate is in the range of 250kbps~100kbps, the pre-filter is required.
FSCALE[11:0]=round{(h×f_S/f_{XTAL})×2^{15}}
where h=(2×frequency deviation)/(data rate).
Here "h" is the modulation index calculated from frequency deviation and data rate.

• **FCF4: Filter Coefficient Control Register 4**

Bit	7	6	5	4	3	2	1	0
Name	CF_B12[7:0]							
R/W	R/W							
Reset	1	0	0	0	0	1	0	1

• **FCF5: Filter Coefficient Control Register 5**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CF_B12[9:8]	
R/W	—	—	—	—	—	—	R/W	
Reset	0	0	0	0	0	0	1	0

• **FCF6: Filter Coefficient Control Register 6**

Bit	7	6	5	4	3	2	1	0
Name	CF_B13[7:0]							
R/W	R/W							
Reset	1	0	0	0	1	0	1	0

• **FCF7: Filter Coefficient Control Register 7**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CF_B13[9:8]	
R/W	—	—	—	—	—	—	R/W	
Reset	0	0	0	0	0	0	0	0

• **FCF8: Filter Coefficient Control Register 8**

Bit	7	6	5	4	3	2	1	0
Name	CF_A12[7:0]							
R/W	R/W							
Reset	0	0	0	1	0	0	1	0

• **FCF9: Filter Coefficient Control Register 9**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CF_A12[9:8]	
R/W	—	—	—	—	—	—	R/W	
Reset	0	0	0	0	0	0	0	0

When the data rate is in the range of 49kbps~2kbps, the following smooth filter is needed.

$CF\_A12[9:0]=mod(2^{10}+[(SFRATIO[1:0] - 1)\times 2^{8}], 2^{10})$

• **FCF10: Filter Coefficient Control Register 10**

Bit	7	6	5	4	3	2	1	0
Name	CF_A13[7:0]							
R/W	R/W							
Reset	0	0	1	0	1	0	1	1

• **FCF11: Filter Coefficient Control Register 11**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CF_A13[9:8]	
R/W	—	—	—	—	—	—	R/W	
Reset	0	0	0	0	0	0	1	1

• **FCF12: Filter Coefficient Control Register 12**

Bit	7	6	5	4	3	2	1	0
Name	CF_B22[7:0]							
R/W	R/W							
Reset	0	0	0	1	0	1	0	0

• **FCF13: Filter Coefficient Control Register 13**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CF_B22[9:8]	
R/W	—	—	—	—	—	—	R/W	
Reset	0	0	0	0	0	0	0	1

• **FCF14: Filter Coefficient Control Register 14**

Bit	7	6	5	4	3	2	1	0
Name	CF_B23[7:0]							
R/W	R/W							
Reset	0	0	1	0	0	0	0	1

• **FCF15: Filter Coefficient Control Register 15**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CF_B23[9:8]	
R/W	—	—	—	—	—	—	R/W	
Reset	0	0	0	0	0	0	0	0

• **FCF16: Filter Coefficient Control Register 16**

Bit	7	6	5	4	3	2	1	0
Name	CF_A22[7:0]							
R/W	R/W							
Reset	0	1	1	1	1	0	0	0

• **FCF17: Filter Coefficient Control Register 17**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CF_A22[9:8]	
R/W	—	—	—	—	—	—	R/W	
Reset	0	0	0	0	0	0	0	0

• **FCF18: Filter Coefficient Control Register 18**

Bit	7	6	5	4	3	2	1	0
Name	CF_A23[7:0]							
R/W	R/W							
Reset	0	0	1	0	1	0	0	0

• **FCF19: Filter Coefficient Control Register 19**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CF_A23[9:8]	
R/W	—	—	—	—	—	—	R/W	
Reset	0	0	0	0	0	0	0	0

The FCF4~FCF19 regsiters define eight groups of IIR coefficients, their recommended settings for different XTAL clock conditions are listed below.

f_{XTAL}	16MHz	16MHz	16MHz	16MHz	16MHz
f_S	250kbps	125kbps	50kbps	10kbps	2kbps
f_D	93.75kHz	46.875kHz	18.75kHz	40kHz	8kHz
D_K[19:0] (H)	f_{RF}×ODDIV/f_{XTAL}, take decimal number				
D_N[6:0] (H)	f_{RF}×ODDIV/f_{XTAL}, take integer number				
SFRATIO[1:0] (D)	0	0	0	1	3
FSCALE[11:0] (H)	294	119	4C	A4	20
CF_B12[9:0] (H)	2CA	01D	0	0	0
CF_B13[9:0] (H)	62	346	0	0	0
CF_A12[9:0] (H)	358	22	0	310	302
CF_A13[9:0] (H)	3E9	331	0	0	0
CF_B22[9:0] (H)	3B3	386	0	0	0
CF_B23[9:0] (H)	3E	12	0	0	0
CF_A22[9:0] (H)	3E9	8	0	0	0
CF_A23[9:0] (H)	39	8	0	0	0

Bank 2 Control Registers

All control registers will be set to their initial value by power-on reset (POR). The software reset will set the control registers to their initial value.

Addr.	Name	Bit							
		7	6	5	4	3	2	1	0
21h	CFG0	Reserved							
2Eh	CFG1	Reserved							
2Fh	CFG2	Reserved							
33h	CFG3	Reserved							
34h	CFG4	Reserved							
39h	CFG5	Reserved							

Note: The addresses which are not listed in this table are reserved for future use, it is suggested not to change their initial values by any methods.

The recommended values for the Bank 2 registers are listed below:

Addr.	Name	Frequency Band	
		433MHz	868MHz
21h	CFG0	97h	
2Eh	CFG1	68h	
2Fh	CFG2	16h (≤50kbps: 06h)	96h (≤50kbps: 86h)
33h	CFG3	01h (OOK mode)/41h (GFSK mode)	
34h	CFG4	90h	
39h	CFG5	9Ch	

Functional Description

Sub-1GHz RF Transceiver

The RF transceiver adopts a fully-integrated, low-IF receiver architrecture. The received RF signal is first amplified by a low noise amplifier (LNA), after which the frequency is down-converted to an intermediate frequency (IF) by a quadrature mixer. The mixer output is filtered by a channel-selected filter which rejects the unwanted out-of-band (OOB) interference and image signals. After filtering, the IF signal is amplified by a analog programmable gain amplifier (PGA). Then the IF signal is digitized by a 10-bit ΣΔ ADC.

The RF transceiver features an Automatic Gain Control (AGC) unit to adjust the receiver gain according to the RSSI, generated at the digital modem. The AGC enables the RF transceiver to operate from sensitivity level to +10dBm input power.

The RF transceiver adopts a fully integrated fractional-N synthesizer which includes RF VCO, loop filter, digital controlled XO (DCXO) and integrated load capacitors for XO. Placing VCO load inductor on the PCB is to lower VCO resonant frequency to achieve an RX mode current consumption low to 5.8mA. The fractional-N synthesizer architecture allows the users to extend their potential usage to a wider frequency range.

The transmit session is a VCO direct modulation architecture. Different from the conventional direct up-conversion transmitters, the GFSK modulation signal is fed into the VCO directly to take advantange of fractional-N synthesizer. As a result, both layout area and current consumption are much smaller compared with direct up-conversion transmitters. The fine resolution can generate a low FSK error GFSK signal. The modulated signal is fed into a Class-E Power Amplifier (PA) and the maximum output power can be up to +20dBm.

Serial Interface

The RF transceiver communicates with a host MCU via a 3-wire SPI interface (CSN, SCK, SDIO) or a 4-wire SPI interface (SDO from GIO1 or GIO2) with a data rate up to 4Mbps. An SPI transmission is an (8+8×n) bits sequence which consists of an 8-bit command and n×8 bits of data, where n can be 0 or any natural number. If the number n is greater than the address boundary, the address will return to zero. The host MCU should pull the CSN (SPI select) pin low in order to access the RF transceiver. Using the SPI interface, user can access the control registers and issue Strobe commands. When writing data to the RF transceiver, the SPI data will be latched into the registers at the rising edge of the SCK signal. When reading data from the RF transceiver registers, the bit data will be transferred at the falling edge of the SCK signal after the target register address has been input.

Command (8 bits)								Data (8 bits)							
C7	C6	C5	C4	C3	C2	C1	C0	D7	D6	D5	D4	D3	D2	D1	D0

SPI Command Format

Two kinds of command are defined. One is 1-byte command only, named CmdO, and the other is 1-byte command followed by n-byte data, named CmdD.

C7	C6	C5	C4	C3	C2	C1	C0	Description	CmdO	CmdD
0	1	A5	A4	A3	A2	A1	A0	Write to control registers		√
1	1	A5	A4	A3	A2	A1	A0	Read from control registers		√
0	0	1	x	x	x	B1	B0	Set register bank	√	
0	0	0	1	x	x	x	0	Write SYNCWORD command		√
1	0	0	1	x	x	x	0	Read SYNCWORD command		√
0	0	0	1	x	x	x	1	TX FIFO write command		√

C7	C6	C5	C4	C3	C2	C1	C0	Description	CmdO	CmdD
1	0	0	1	x	x	x	1	RX FIFO read command		√
1	0	0	1	1	1	1	1	Read RF transceiver ID command		√
0	0	0	0	1	0	0	0	Software reset command	√	
0	0	0	0	1	0	0	1	TX FIFO address pointer reset command	√	
1	0	0	0	1	0	0	1	RX FIFO address pointer reset command	√	
0	0	0	0	1	0	1	0	Deep Sleep mode	√	
0	0	0	0	1	0	1	1	Idle mode	√	
0	0	0	0	1	1	0	0	Light Sleep mode	√	
0	0	0	0	1	1	0	1	Standby mode	√	
0	0	0	0	1	1	1	0	TX mode	√	
1	0	0	0	1	1	1	0	RX mode	√	

A5~A0: The address of control registers;

x: Hardware doesn't care but it is recommended to set to 0 by software;

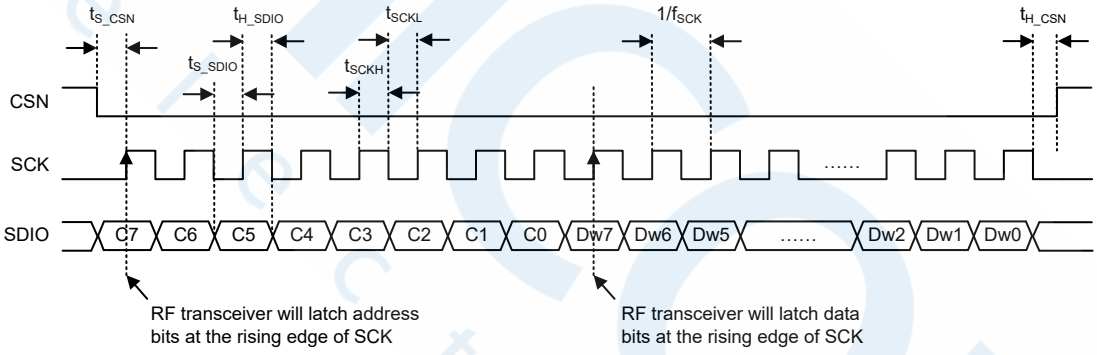
B1~B0: Bank number.

Note: 1. The RF transceiver supports multi-byte read/write operations and the address is increased automatically after each read or write operation.

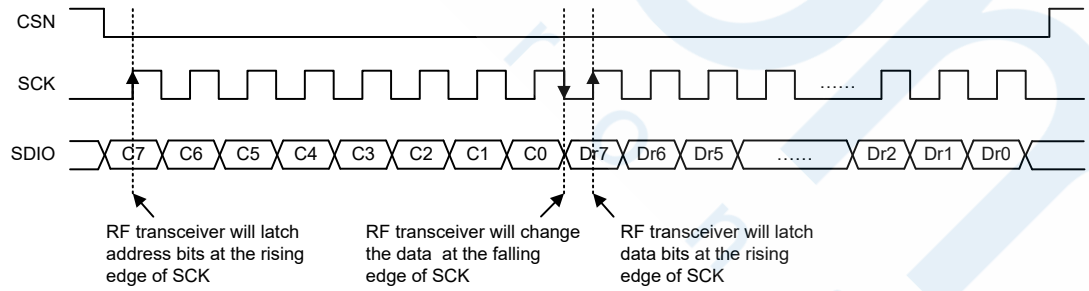
2. Using software to read/write multiple bytes is allowed after one read/write command in a single CSN enabled cycle.

3. In the sleep mode, GIOs will keep the same level of the last operation mode.

SPI Timing



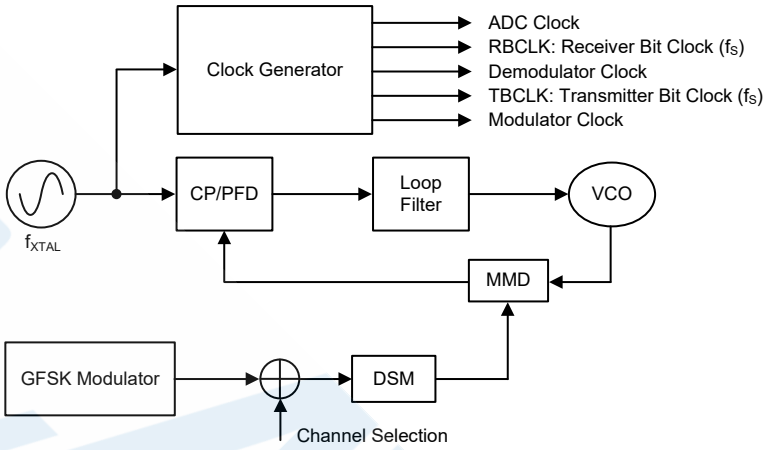
3-Wire SPI Interface Write 1-byte Data Operation



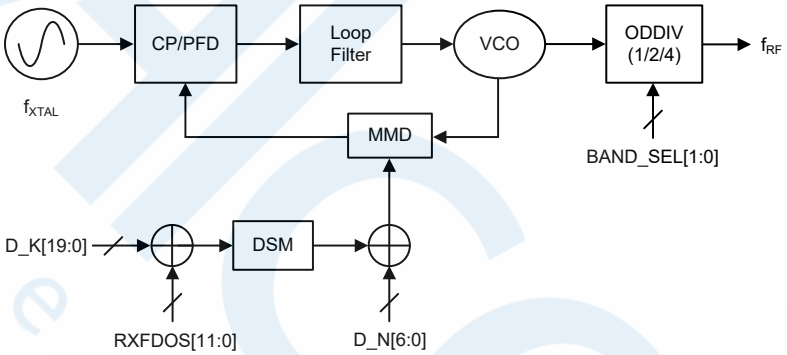
3-Wire SPI Interface Read 1-byte Data Operation

System Clock

The main system clock of the RF transceiver comes from the X'tal oscillator. All internal operation clocks of various functional blocks are derived from the X'tal oscillator.



Frequency Synthesizer



The RF transceiver frequency is generated by a high resolution fractional-N Delta Sigma frequency synthesizer. By appropriate setting on the configuration parameters D_N[6:0] and D_K[19:0], a low-noise LO frequency can be generated to comply with various radio regulatory standards including ETSI EN, FCC, etc.

$$D\_N[6:0]=Floor(\frac{f_{RF}\times ODDIV}{f_{XTAL}})$$

$$D\_K[19:0]=Floor((\frac{f_{RF}\times ODDIV}{f_{XTAL}}-D\_N[6:0])\times 2^{20})$$

$$RXFDOS[11:0]=Floor((\frac{f_{IF}}{f_{XTAL}})\times 2^{17})$$

State Machine

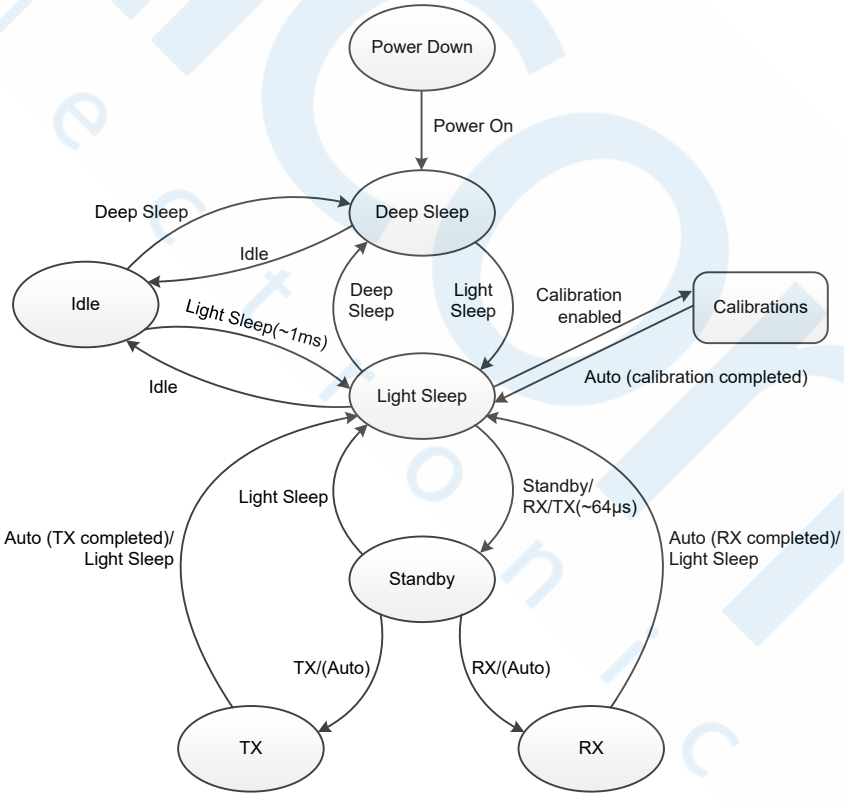
There are seven operating modes in the RF transceiver. The operation modes and key funcitons on/off state in the corresponding mode are listed below.

1. Power Down mode
2. Deep Sleep mode
3. Light Sleep mode
4. Standby mode
5. Idle mode
6. TX mode
7. RX mode

Mode	Register Retention	3.3V	LIRC	Regulator	XO	Standby+VCO	TX	RX	Strobe Command
Power Down	No	OFF	OFF	OFF	OFF	OFF	OFF	OFF	—
Deep Sleep	Yes	ON	OFF	OFF	OFF	OFF	OFF	OFF	0000_1010
Light Sleep	Yes	ON	OFF	ON	ON	OFF	OFF	OFF	0000_1100
Idle	Yes	ON	ON	OFF	OFF	OFF	OFF	OFF	0000_1011
Standby	Yes	ON	OFF	ON	ON	ON	OFF	OFF	0000_1101
TX	Yes	ON	OFF	ON	ON	ON	ON	OFF	0000_1110
RX	Yes	ON	OFF	ON	ON	ON	OFF	ON	1000_1110

TX/RX FIFO Mode (DIR_EN=0) State Machine

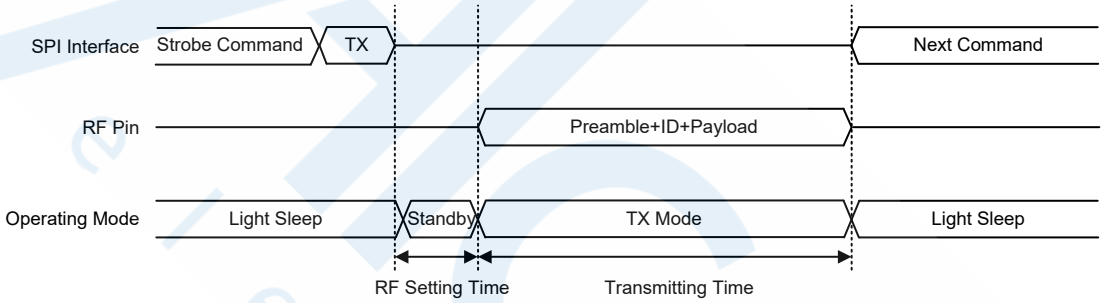
If the DIR_EN bit is cleared to 0, the RF transceiver mode transactions are implemented by strobe command from the host MCU and the TX/RX data are derived from the packet handling hardware.



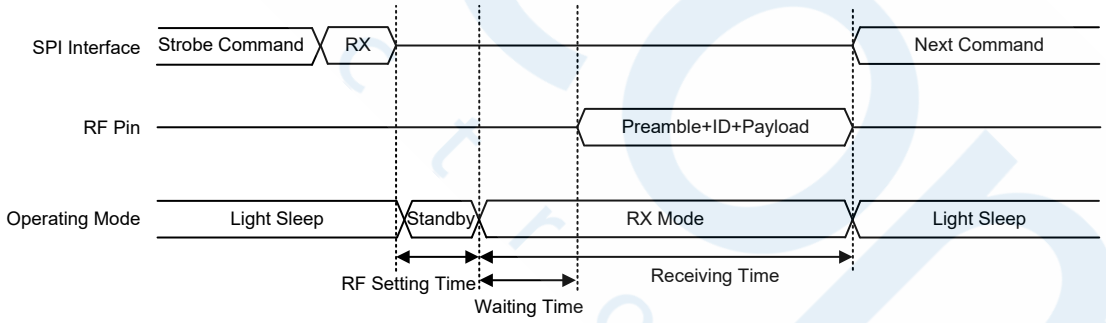
FIFO Mode State Diagram

Initially, the RF transceiver is in the Power Down mode. After the RF transceiver completes the internal power on reset, it will enter the Deep Sleep mode and wait for further strobe commands from the host MCU. If the Light Sleep command is received, the RF transceiver will enable the internal LDO, oscillate the XO and enter the Light Sleep mode. In this state, the host MCU can have the RF transceiver execute calibration process if necessary. For normal TRX operations, the host MCU can issue an RX or TX command to the RF transceiver. After receiving the TX or RX command, the RF transceiver will first enter the Standby mode which lasts a certain period known as TX/RX settling time. After the settling time has escaped, the RF transceiver will finally enter the RX or TX mode. The RF transceiver will stay in the TX/RX sate until the TX/RX event is completed, after which the RF transceiver will return to the Light Sleep mode automatically.

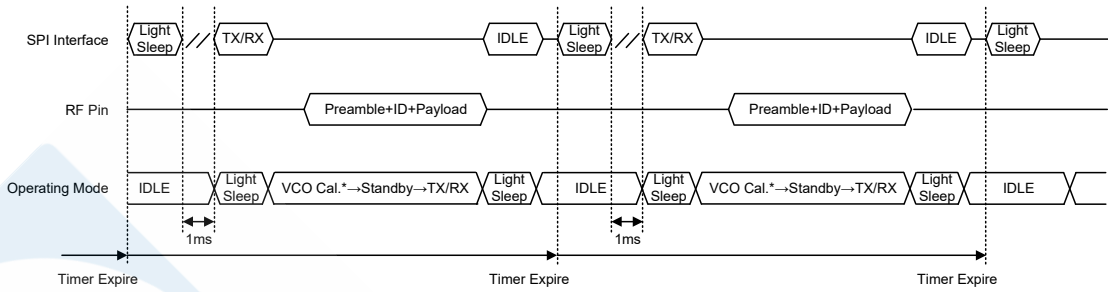
For low power periodical wireless transmission, the RF transceiver supports low power Idle mode where the LIRC and wake-up timer are turned on. By appropriate timer setting and issuing the Idle mode command, the RF transceiver will turn off the LDO and XO and enter the Idle mode. The wake-up timer starts to count after ATR_EN is set to "1". The RF transceiver stays in the Idle mode until the timer expires and then an IRQ will be asserted on the GIO to wake up the host MCU. Then the host MCU can have the RF transceiver enter the Light Sleep mode and continue to execute normal TX/RX operations. After the TX/RX event is completed, the host MCU can issue the Idle command to have the RF transceiver return to the Idle mode again.



TX Timing in FIFO Mode



RX Timing in FIFO Mode



Note: VCO Cal.(VCO Calibration) time: ~152μs@433MHz / ~96μs@868MHz.

Periodical TX/RX Timing

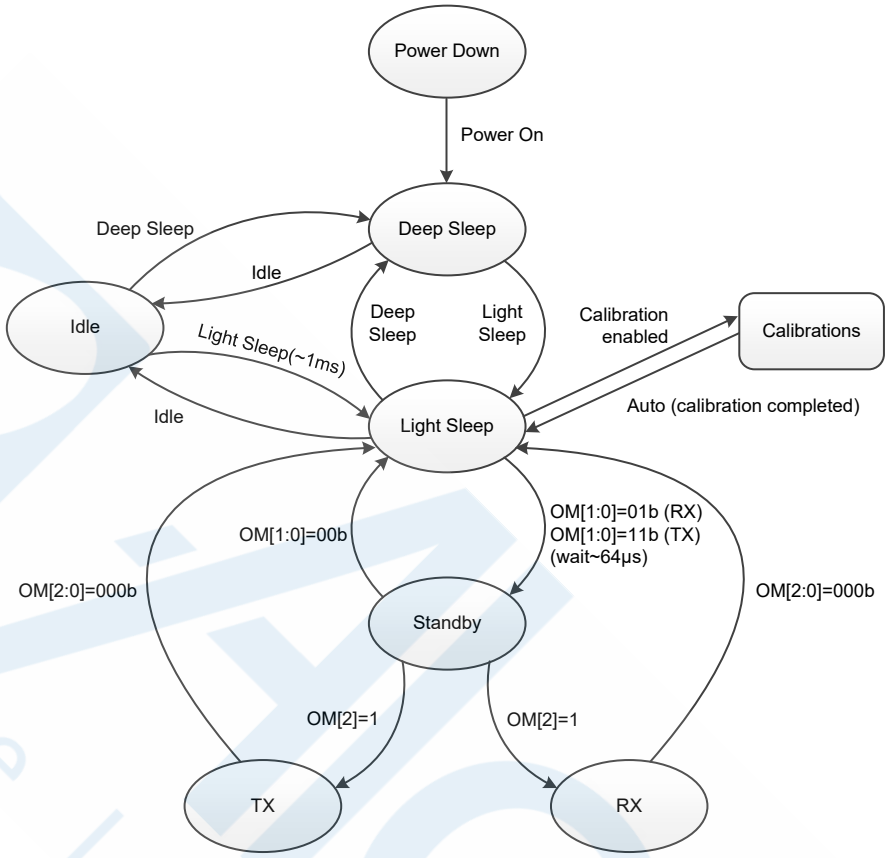
TX/RX Direct Mode (DIR_EN=1) State Machine

If the DIR_EN bit is set to 1, TX data is derived directly from the host MCU to RF transceiver and RX data is sent directly from the RF transceiver to the host MCU. In order to simplify the data bit clock synchronization between the RF transceiver and the host MCU, the RF transceiver outputs the TBCLK/RCLK from GIO4 by setting GIO4S[3:0]. Both TBCLK and RBCLK are in 50/50 duty cycle. In the transmitting mode, the host MCU outputs bit data at the rising edge of the TBCLK signal and the RF transceiver samples the TX bit data at the falling edge of the TBCLK signal. In the receiving mode, the host MCU receives data at the rising edge of the RBCLK signal and the RF transceiver outputs bit data at the falling edge of the RBCLK signal. The host MCU can select GIO1 or GIO2 for the TX/RX bit data transmission by setting GIO1S[2:0] or GIO2S[2:0].

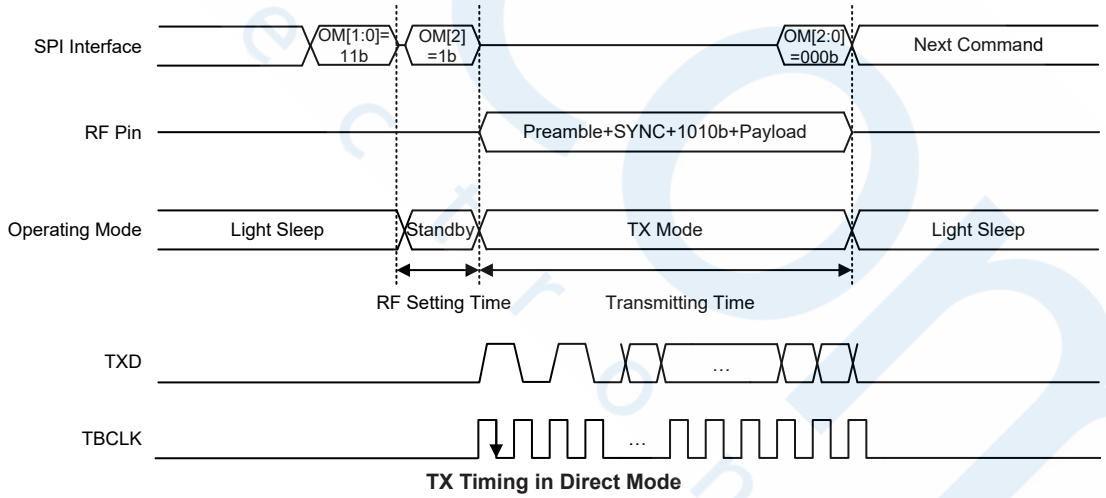
For TX operations in the direct mode, the host MCU needs to set the OM[1:0] bits, i.e. RTX_SEL and SX_EN, to 11b to select the TX mode and have the RF transceiver enter standby mode first, then set the OM[2] bit, RTX_EN, to 1 to have the RF transceiver start to transmit the TX data. As long as the host MCU sets OM[2:0] to 000b, the RF transceiver will return to the Light Sleep mode.

For RX operations in the direct mode, the host MCU needs to set OM[1:0] to 01b first, then set OM[2] to 1 to have the RF transceiver start to receive data from the air. After the RF transceiver receives the matched SYNCWORD code, it will output the RBCLK clock, receive data bit (payload part) and then transmit to the host MCU.

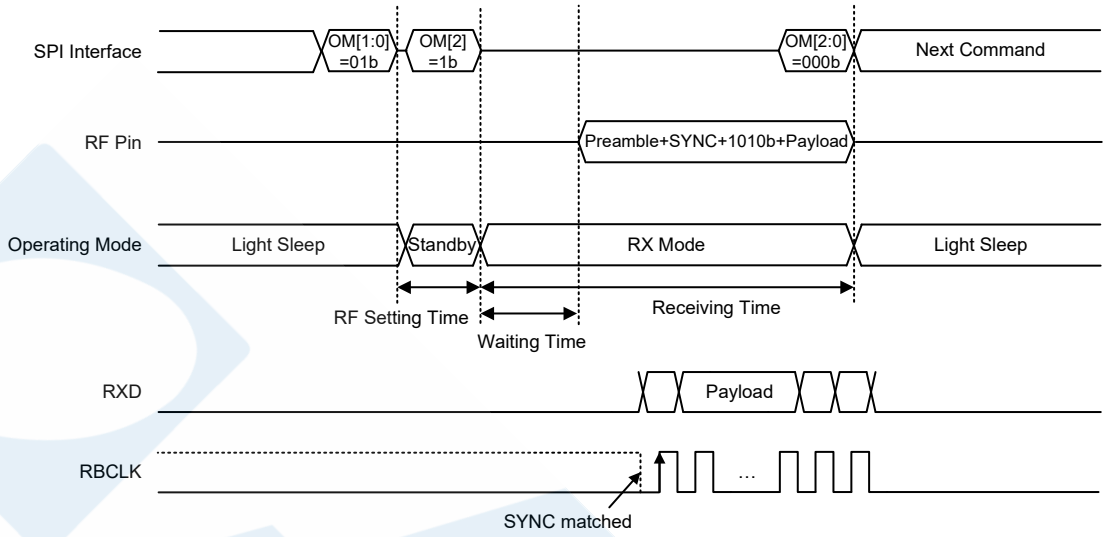
In direct mode, the transmission data length has no limit.



Direct Mode State Diagram



TX Timing in Direct Mode



RX Timing in Direct Mode

Calibration

The RF transceiver has three calibration functions, VCO, RC and LIRC calibrations, allowing users to auto select proper setting to compensate the PVT (Process-Voltage-Temperature) variation effect. The control bit, ACAL_EN, is used to enable the VCO and RC calibration functions at the same time and both calibration functions will be automatically implemented after this bit is set high. When the calibrations are completed, the ACAL_EN bit is cleared to zero by hardware. The host MCU can poll the ACAL_EN bit status or use the calibration complete interrupt flag CALCMPF to check the calibration status. The RF transceiver also has an independent enable bit, LIRCCAL_EN, for the LIRC calibration function, allowing to independently implementing the LIRC calibration function.

LIRC Calibration

There is an internal low frequency RC oscillator in the RF transceiver providing a clock source for the wake-up timer in the Idle mode. In order to compensate the PVT (Process-Voltage-Temperature) variation error (up to ±10%) impact on the accuracy of LIRC, the host MCU can trigger the LIRC calibration to improve the wake-up timer accuracy error to be less than ±1%.

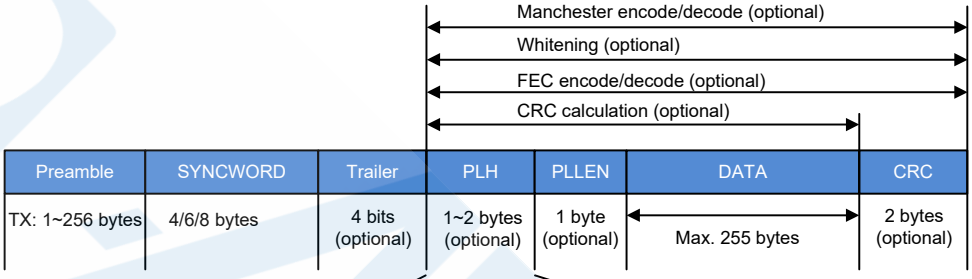
The host MCU need to configure LIRC_OW=0 and LIRC_EN=1 before the LIRC calibration. Then the RF transceiver will do LIRC calibration when LIRCCAL_EN is set to 1 by the host MCU during the Light Sleep mode. The LIRCCAL_EN bit is reset to 0 by hardware on the completion of LIRC calibration. The LIRC calibration process would take about 4ms.

Packet Handler

In the TX mode, the packet handler is used to move the transmitting data out of FIFO and implement channel coding according to the packet format, then sends the packet to the modulator. In the RX mode, the packet handler is used to implement channel decoding with data from the demodulator and store the payload data into FIFO.

The packet handler performs several tasks such as Preamble and SYNCWORD insertion, Forward Error Correction, CRC calculation/checking, whitening/de-whiteing and Manchester encode/decode.

Packet Format



Preamble	MSB	SYNCWORD	LSB	Trailer
0101...01	0	…	0	1010
1010...10	1	…	1	0101

Note: 1. Preamble format will follow SYNCWORD MSB to inverse.
If MSB=0, Preamble format=0101…01
If MSB=1, Preamble format=1010…10
2. Trailer format will follow SYNCWORD LSB to inverse.
If LSB=0, Trailer format=1010
If LSB=1, Trailer format=0101
3. The Trailer field contains 4 bits and is an optional field which is enabled by TRAILER_EN.

- Preamble

  The packet starts with a preamble with a length of 1~256 words set by TXPMLEN[7:0] in the TX mode. The word length is determined by the preamble mode. There are two preamble modes switched by PMLP_EN.

  PMLP_EN=0 – auto preamble pattern mode (compatible with BC66F36x2)

  The first bit of the preamble pattern is equal to the inverted SYNCWORD MSB and then continue with 1/0 toggle. In this mode, word unit=1 byte.

  PMLP_EN=1 – preamble pattern by register mode

  The preamble pattern is defined by PMLPAT[15:0] and the length is defined by PMLPLEN.

  PMLPLEN=0, preamble pattern=PMLPAT[7:0]; word length=1 byte

  PMLPLEN=1, preamble pattern=PMLPAT[15:0]; word length=2 bytes

- SYNCWORD

  The SYNCWORD length, which is set by ({SYNCLEN[1:0], SYNCLENLB}+1), can be 4/6/8 bytes in both TX and RX modes. When the RX side receives a matched SYNCWORD packet, the DATA field will be stored in the FIFO. Syncword Low Byte cannot be 0x55 or 0xAA.

- Trailer

  The trailer field length is fixed at 4 bits which is optional and enabled by the TRAILER_EN bit.

- PLH (Payload Header)

  The PLH is optional and enabled by PLH_EN. The payload header length can be 1 or 2 bytes set by PLHLEN. When the PLHLEN bit is 0, only {PID[1:0], PLHA[5:0]} field appears in the packet. PID[1:0] is located in bit[7:6] of the payload header field. When the PLHLEN bit is set to 1, the payload header address is extended to 2 bytes which is formed by PLHA[5:0] and PLHEA[7:0].

  PLHA[5:0] and PLHEA[7:0](when PLHLEN=1) have two functions controlled by the PLHAC_EN bit. If PLHAC_EN=0, PLHA[5:0] and PLHEA[7:0] can be used as software flags and the actual function can be defined by users. If PLHAC_EN=1, the RF transceiver will compare the local {PLHA[5:0], PLHEA[7:0]} field with the received {PLHA[5:0], PLHEA[7:0]} field. If matched, the receiving data will be moved into the RX FIFO, otherwise the following incoming data will be abandoned. PLHA[5:0]=0 and PLHEA[7:0]=0(checked when PLHLEN=1) is a special address that the hardware will not check whether the address is matched or not. It is used to support broadcast function for RF transceivers with same SYNCWORD but different payload header address values.

- PLLEN (Payload Length)

  The PLLEN field is optional and its length is fixed at 1 byte once being enabled by PLLEN_EN. When this bit is set high, the DATA field length is variable and is determined by the PLLEN field. In TX mode, the transmitter sends the TXDLEN defined bytes of data from TX FIFO and the TXDLEN is automatically latched into the PLLEN field. In RX mode, the receiver gets the PLLEN and receives the PLLEN defined bytes data into RX FIFO.

- DATA

  When in the TX mode, the TX data length is determined by the TXDLEN[7:0] field. The maximum length is 255 bytes in the extend FIFO mode. In the special case of infinite FIFO mode, the length can exceed 255 with an infinite length. If PLLEN_EN=1, the PLLEN field in the TX packet is enabled and the PLLEN content is equal to TXDLEN[7:0]. When in the RX mode, the RX data length is set by RXDLEN[7:0] ifPLLEN_EN=0 and by PLLEN field in the receiving packet if PLLEN_EN=1.

- CRC

  The CRC field is optional and is enabled byCRC_EN. It is recommend to always set CRC_EN to 1 for data correctness checking. There are two CRC formulas selected by setting the CRCFMT bit.

  CRCFMT=0: CCITT-16-CRC $G(X)=X^{16}+X^{12}+X^{5}+1$

  CRCFMT=1: IBC-16-CRC $G(X)=X^{16}+X^{15}+X^{2}+1$

- FEC

  The optional data encode/decode function can be enabled by FEC_EN. Use (7,4) Hamming code to correct 1-bit error and more than 1-bit error detect for each 4-bit data. After FEC, the data length for each data will be (4+3)×2=14 bits.

Bit	7	6	5	4	3	2	1
Transmitted Bit	D3	D2	D1	P2	D0	P1	P0
P0	Y	N	Y	N	Y	N	Y
P1	Y	Y	N	N	Y	Y	N
P2	Y	Y	Y	Y	N	N	N

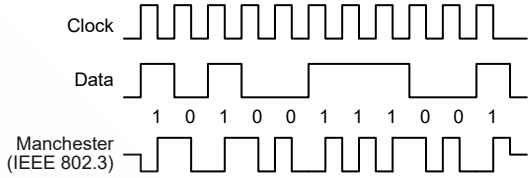
Hamming Code Function Table

- Data Whitening

  The optional data whitening/de-whitening function can be enabled by WHT_EN. Use PN7/PN9 code to implement XOR operation with the transmitted data. The whitening seed is set by WHTSD[8:0].

- Manchester Code

  The optional Manchester encode/decode function can be enabled by MCH_EN. Each bit after Manchester encoding will be extended into two bits and recovered to one bit data after decoding.



Manchester Code Example

FIFO Operation Modes

In Burst mode, data transmission to the RF transmitter is derived from FIFO and is pre-written by the host MCU. There are 4 FIFO modes to support various applications. They are the Simple FIFO mode, Block FIFO mode, Extend FIFO mode and Infinite FIFO mode.

FIFO Reset

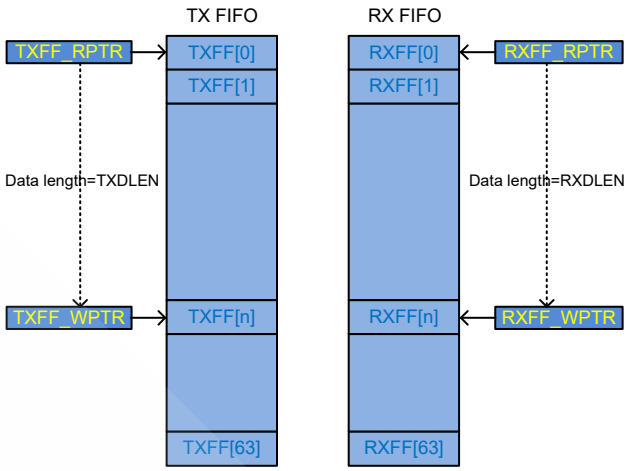
To use the FIFO in the burst mode, issue the TX FIFO address pointer reset command and RX FIFO address pointer reset command to reset the FIFO pointer and buffer first. After this, the FIFO is in the initial state same as reset.

Simple FIFO Mode

This FIFO mode is used for general applications with a TX/RX data length less then or equal to 64 bytes. The data length should not exceed 64 bytes. To use the simple FIFO mode, the host MCU must write the transmitting data to FIFO by the SPI write FIFOcommand. The transmitting sequence is first written byte first out and the MSB in each byte first out to the transmitter. Users should determine all transmitting data packet format including the preamble, SYNCWORDand packet encoding such as FEC, CRC, whiting. After FIFO data filling out is completed, clear the TXFFSA[5:0] field and set TXDLEN[7:0]/RXDLEN[7:0] field to the desired transmitting/receiving length in bytes. Then issue the TX command to start the transmission. After the current transmitting is completed, the data will be kept in FIFO to wait for the next transmission.

Programming procedure:

1. Reset TX FIFO by the SPI reset TX FIFO command.
2. Reset RX FIFO by the SPI reset RX FIFO command.
3. TXFFSA[5:0] must be cleared to 0.
4. Fill out TX FIFO by the SPI write FIFO command.
5. Set TXDLEN[7:0]/RXDLEN[7:0] to control the TX/RX length in bytes.
6. Issue the TX command for transmitter and RX command for receiver.
7. TX/RX completion is acknowledged by the TX/RX complete IRQ.
8. Re-transmitting TX packet with the same data will auto-reset TXFF_RPTR to 0.

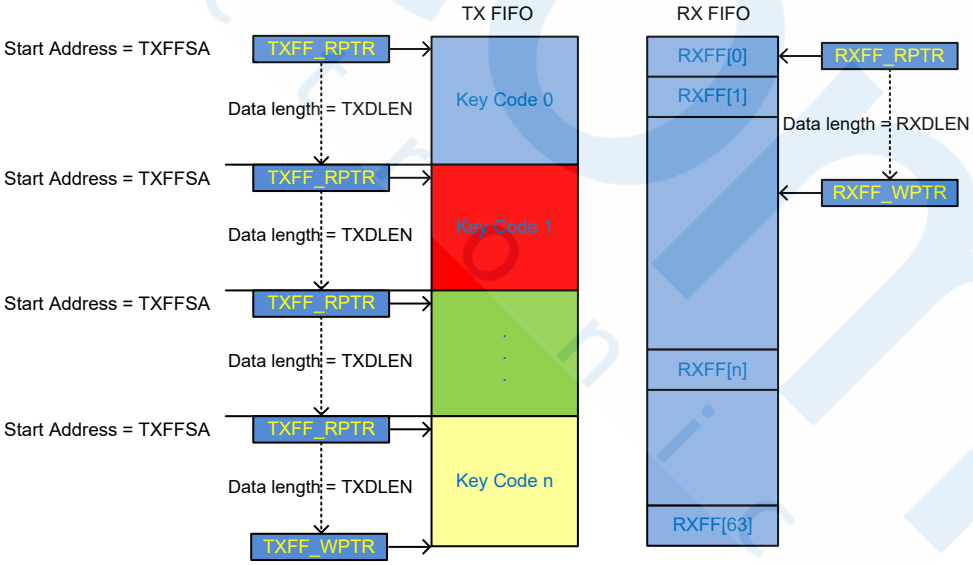


Block FIFO Mode

The Bock FIFO mode is used to support multi-key code applications. Users should write all the key codes to FIFO first. When a key is pressed, the host MCU will detect the key and set TXFFSA[5:0] to the target key code start address and set TXDLEN[7:0] to indicate the key code length and issue the TX strobe command to start the transmission. The maximum FIFO length is also limited to 64 bytes.

Programming procedure:

1. Write key code 0~n to TX FIFO by SPI write FIFO command.
2. When a key is pressed, the host MCU will set TXFFSA[5:0] to the start address of the corresponding key code.
3. Set TXDLEN[7:0] for key code length.
4. Set the RXDLEN[7:0] to key code length and then enter the RX mode by SPI command.
5. Issue TX command for transmitter and RX command for receiver.
6. TX/RX completion is acknowledged by the TX/RX complete IRQ.



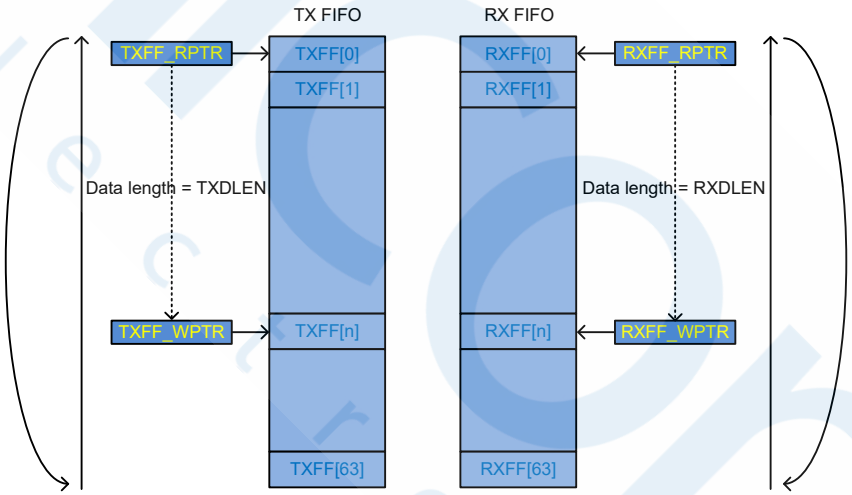
Extend FIFO Mode

The Extend FIFO mode is used for transmissions with a long payload data packet. The maximum length is 255 bytes. As the physical FIFO length is 64 bytes, to extend the available transmitting length in one packet, a handshake mechanism is needed between the host MCU and the FIFO controller.

Set FFMG[1:0] to determine the FIFO data length margin and set FFMG_EN to enable the margin detect function to inform the MCU when the TX FIFO data fullness level is less than the margin. The MCU should write data to TX FIFO fast enough when receiving this reminding signal to avoid transmission being terminated by TX FIFO data length low to zero.

Programming procedure:

1. Set FFMG_EN to enable FIFO length margin detection function (i.e. FIFO low threshold detect function) and set FFMG[1:0] to select the threshold, 4, 8, 16 or 32 bytes.
2. Set the FIFOLTIE bit to 1 to enable the FIFO low threshold IRQ.
3. Set GIOnS field (n=1, 2, 4)=101b to output IRQ on GIO1, GIO2, GIO4.
4. TX: If MCU detects the FIFO low threshold IRQ signal, it will move data into TX FIFO with a data length less than or equal to (64-threshold). Then the MCU clears the FIFO low threshold IRQ flag FIFOLTIF and repeats the same routine until all TX data are completely written to TX FIFO.
5. RX: If MCU detects the FIFO low threshold IRQ signal, it will read data from RX FIFO. Then the MCU clears the FIFO low threshold IRQ flag FIFOLTIF and repeats the same routine until receiving the RX completion IRQ to read the remaining data from RX FIFO.



Infinite FIFO Mode

Programming procedure:

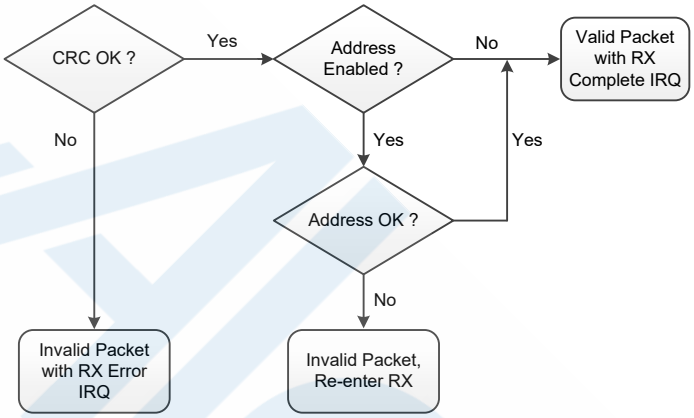
1. Set FFINF_EN to 1 to enable the Infinite FIFO mode.
2. The handshaking and IRQ function are identical with the Extend FIFO mode.
3. TX: If receiving the FIFO low threshold IRQ, the MCU continues to write TX data to TX FIFO with a data length less than or equal to (64-threshold). Then the MCU clears the FIFO low threshold IRQ flag and repeats the same routine until it wants to terminate the infinite FIFO mode. To terminate the infinite FIFO mode, after receiving IRQ and moving data to TX FIFO, the MCU should clear FFINF_EN to zero and set TXDLEN[7:0] to the remaining data length if

the remaining transmitting length is less than 192 bytes and longer than 64 bytes. The packet will be terminated when all of the target data are transmitted completely.

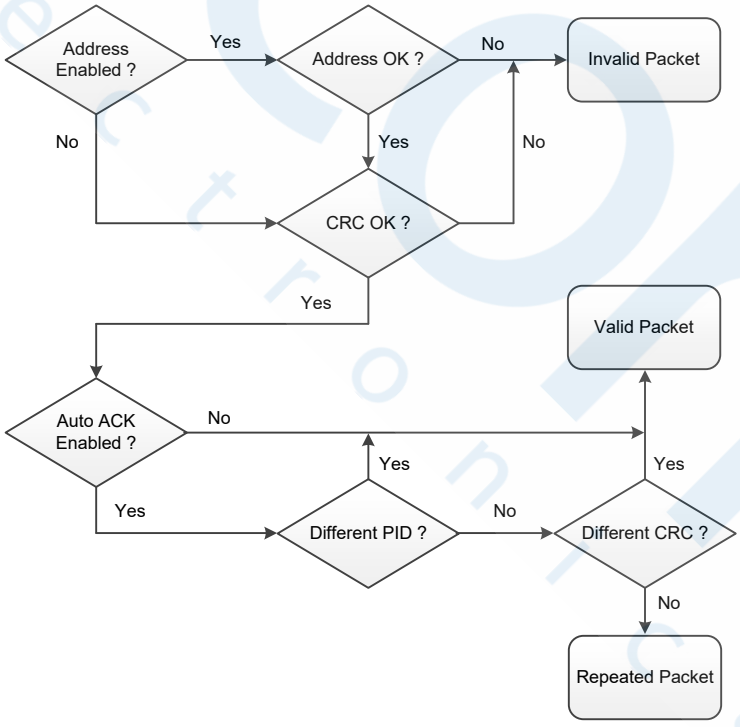
4. RX: If receiving the FIFO low threshold IRQ, the MCU reads data from RX FIFO. Then the MCU clears the FIFO low threshold IRQ flag and repeats the same routine until it wants to terminate the infinite FIFO mode. To terminate the infinite FIFO mode, after receiving IRQ and reading data from RX FIFO, the MCU should clear FFINF_EN to zero and set RXDLEN[7:0] to the remaining data length if the remaining receiving length is less than 192 bytes and longer than 64 bytes. The packet will be terminated when all of the target data are received completely.

Receiving Packet Judgement

In normal RX operating mode, package reception follows the following judgement criteria.

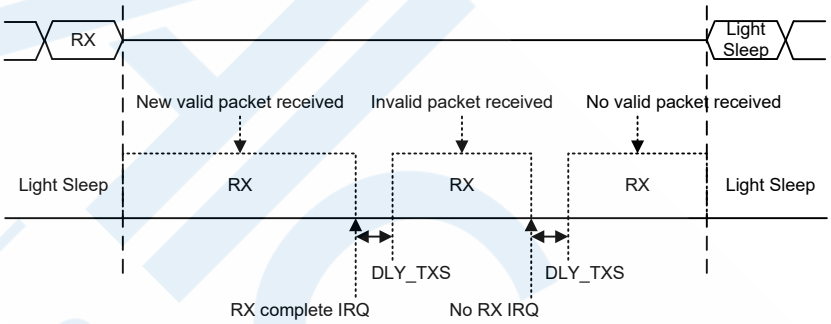


The RF transceiver adopts extra receiver packet judgment for the continuous RX mode and auto-acknowledge mode. The main purpose of these special link layer functions are used to alleviate MCU loading when handling TRX packet transaction.



Continuous RX Mode

There is a special continuous RX operating mode supported in the RF transceiver. The MCU can enable this continuous RX mode by setting the RXCON_EN bit high and start the continuous RX mode by issuing the RX strobe command to the RF transceiver. If there is a valid RX packet received, the RF transceiver will issue an RX completion IRQ to the MCU. The RF transceiver then repeats the RX operation after a duration defined by DLY_TXS[2:0] to keep listening for incoming packets. If an invalid packet is received, the RF transceiver would only repeat the RX operation without issuing the RX completion IRQ to the MCU. The MCU stops the continuous RX by issuing the Light Sleep strobe command to the RF transceiver. In the continuous RX mode, only simple FIFO mode can be used. In order to prevent the receiving packet data length field from being corrupted by new incoming packets before the MCU reads data from RX FIFO, users should set RXPL2F_EN=1 and PLLEN_EN=1 to have the PLLEN information stored into the RX FIFO. Because of the existence of PLLEN byte, the maximum packet data length becomes 63 bytes. If a new incoming packet arrives before the MCU reads RX FIFO, a FIFO overflow error will happen, in which condition the RF transceiver will issue an RX error IRQ to the MCU with FIFO overflow error flag RXERRIF set. At this moment, the MCU should exit the continuous RX mode and reset the RX FIFO pointer.

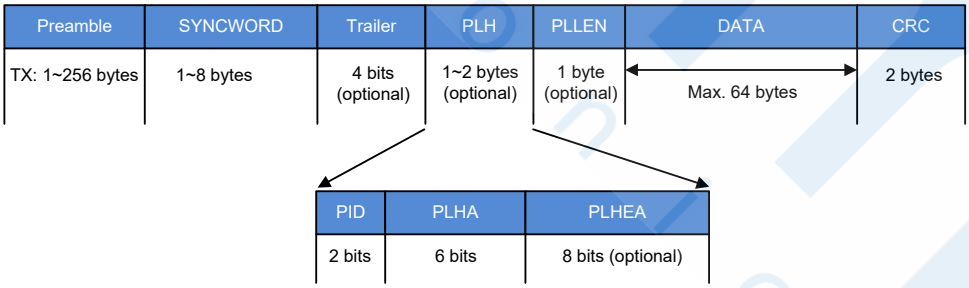


Continuous RX Mode

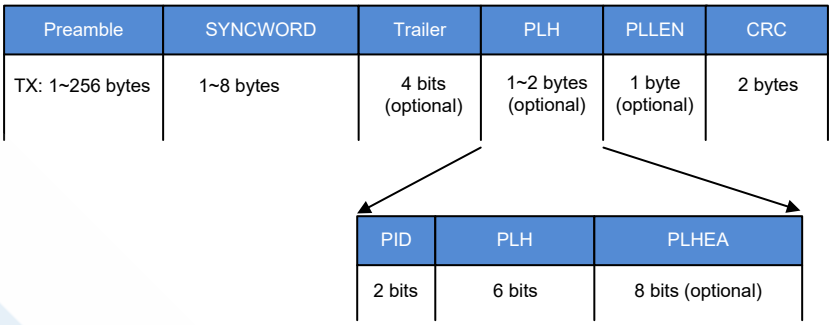
ARK Mode: Auto-Resend and Auto-Ack

The RF transceiver supports auto-resend and auto-ack mechanism by setting the ARK_EN bit high. This mechanism enables an easy two-way communication implementation however can only be operated in the simple FIFO mode.

Set ARK_EN to 1 to enable the RF transceiver to enter the auto-resend and auto-ack ready mode. Then, auto-resend is triggered by the TX strobe command from the MCU and auto-ack is triggered by RX strobe command from the MCU. Packet format transmitted from the master to the slave in the auto-resend mode are illustrated below.

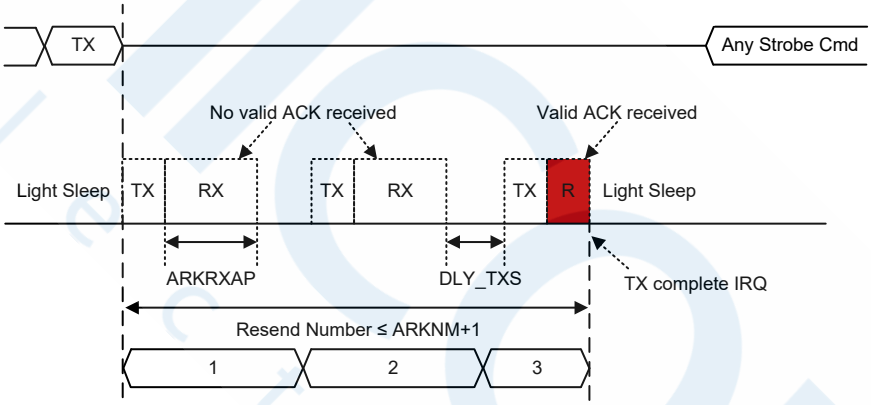


The slave side in the auto-ack mode uses the packet format as the following to be an acknowledge packet transmitted to master. Note that there is no payload data field used in the acknowledge packet.

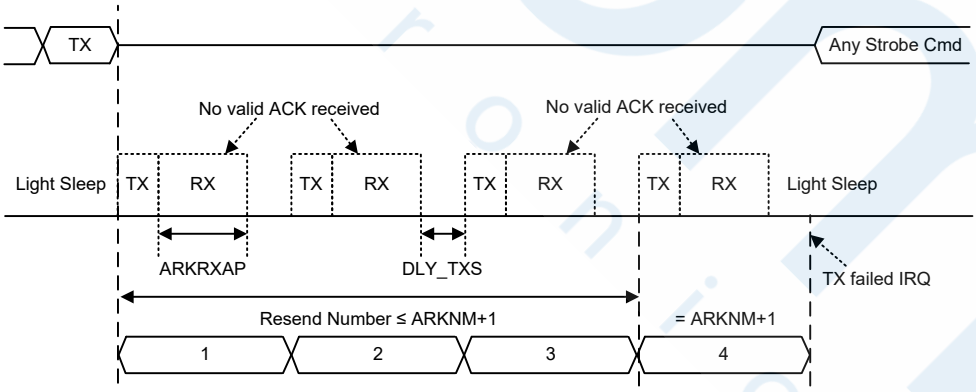


If the address field is used for the ARK mode, the auto-resend (master) side should configure the same address as the auto-ack (slave) side.

After configuring ARKNM[3:0], ARK_EN and ARKRXAP[7:0], the MCU starts the auo-resend process by issuing the TX strobe command. The RF transceiver starts to transmit data from the TX FIFO and then enters the RX mode after the TX completion. The RX period is in multiples of 250μs (default) which is determined by (ARKRXAP[7:0]+1). If the RF transceiver receives a valid acknowledge packet from the slave side within the RX period with CRC checked correct, it will return to the Light Sleep mode and issue a TX completion IRQ to the MCU. Otherwise, the RF transceiver will check if the resend number has reached the limit set by (ARKNM[3:0]+1), if not, it will go to the TX mode to transmit the same TX data from the TX FIFO and the resend number will be increased by one.



Auto-resend: ACK Packet Received before ARKNM Limit

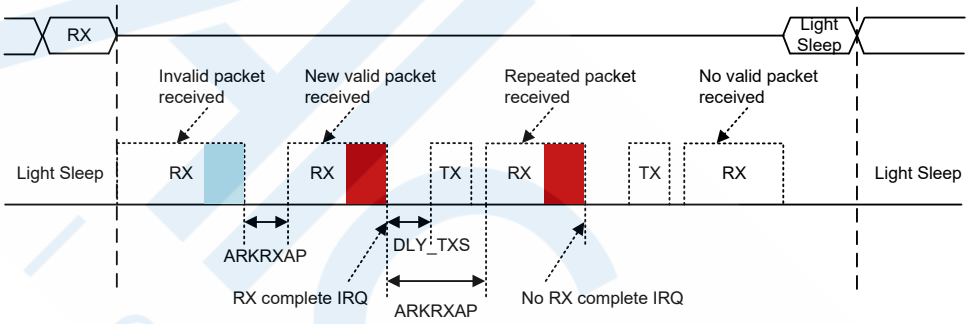


Auto-resend: No Valid Packet Received before ARKNM Limit

Regarding the auto-ack in the slave side, the MCU issues the RX strobe command to start the auto-ack process and issues the Light Sleep strobe command to stop the auto-ack process. In the auto-ack mode, an extra PID/CRC filtering function will be applied for the slave side to check the packet received. If the PID/CRC of the new incoming packet is same as the stored PID/CRC of the last packet, then the newly received packet would be treated as a repeated packet.

During the auto-ack process, if the RF transceiver receives a valid packet with different PID/CRC and CRC/address checked correct, it will issue an RX completion IRQ to the MCU and auto-transmit the ACK packet to the master. If the RF transceiver receives a packet with the same PID/CRC and CRC/address checked correct, it will treat this packet as the repeated packet. Then the RF transceiver will not issue the RX IRQ to the MCU but still auto-transmit the ACK packet to the master. If the RF transceiver receives a packet with CRC/address checked failed, no IRQ is issued and the RF transceiver will automatically re-do the RX operation to continually listening for incoming packets.

The gap period for the RF transceiver to restart the next RX operation after the current RX completion is defined by ARKRXAP[7:0]. In general cases, the MCU should fetch the receiver FIFO data within this period after receiving the RX completion IRQ. Besides, the MCU needs to wait for a same duration if it wants to leave the ARK mode after receiving the RX completion IRQ.



Auto-ACK Process

ATR Mode: Auto-Transmit-Receive

There is a special ATR operation mode in the RF transceiver to reduce the external host's loading. Two ATR functions are implemented within the RF transceiver, one is WOR (Wake-On-RX) and the other is WOT (Wake-On-TX). They can only be operated with simple FIFO mode. These two operating modes need to co-work with an Idle mode timer which operates at a low frequency. The low frequency clock can be sourced from the internal LIRC or from the external ROSCi clock by setting the ATRCLKS bit in the ATR1 register. There are two operation modes for the ATRCT timer which is selected using the ATRCTM bit. Clearing the ATRCTM bit to 0 will select the single mode, where the ATRCT timer will restart upon every ATR transation when entering the Idle state. The ATRCT timer will stop and leave the ATR mode upon receiving the Light Sleep command. Setting the ATRCTM bit to 1 will select the continuous mode, where the ATRCT timer will start to operate upon receiving the Idle command and continuously run until the ATR_EN bit or the ATRCTM bit is cleared to zero.

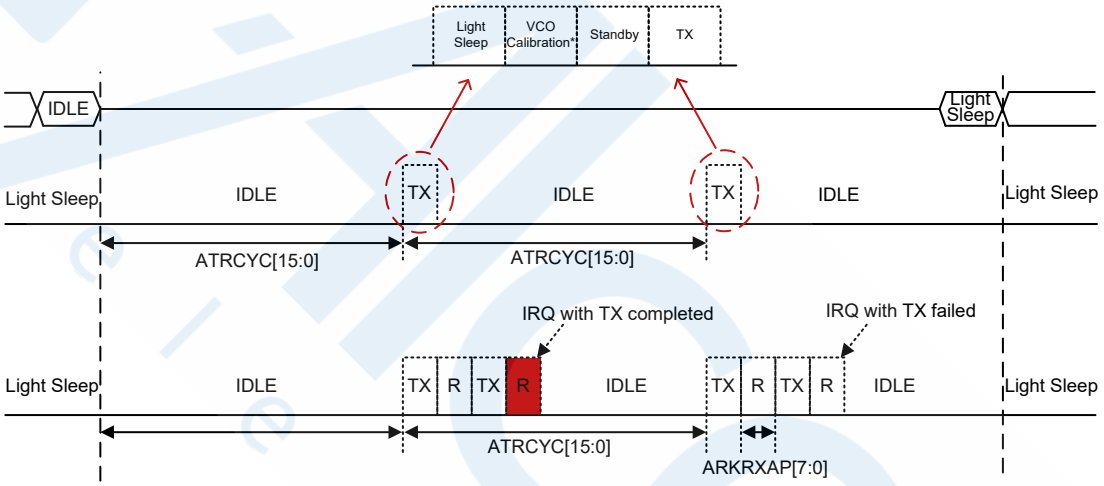
After entering the ATR mode, only the Idle, Light Sleep, Set Register Bank and control register read/write commands can be recognized by the RF transceiver.

WOT (Wake-On-TX) Function

When the WOT function is enabled by setting the ATR_EN bit to 1 and the ATRM[1:0] bits to 00b, the RF transceiver will periodically wake up from the Idle mode and transmit TX FIFO contents

without interaction with the host MCU. The RF transceiver starts the WOT process upon receiving the Idle strobe command from the MCU and stops the WOT process upon receiving the Light Sleep strobe command from the MCU. The ATRCYC[15:0] bits are used to set the wake-up period for the WOT function. At the moment of timer expiration, the wake-up timer will trigger the RF transceiver to leave the Idle state and enter the active state to transmit data, at the same time the ATRCYC[15:0] content will be reloaded into the timer's counter. After finishing the TX operation, the RF transceiver will return to the Idle mode and stay in this state until next wake-up timer expiration occurs. In the active state, the RF transceiver only implements wake-up transmission once by default. Users can extend the wake-up transmitting mechanism by combining with the ARK function. The repeated transmitting number is controlled by (ARKNM[3:0]+1). The time duration between the repeated transmitting packets is inserted with one RX slot and controlled by ARKRXAP[7:0] in the ATR8 register. If the RF transceiver receives ACK in the RX slot, a TX completion IRQ will be issued to inform the host MCU.

Note: 1. VCO Calibration time: ~152μs@433MHz / ~96μs@868MHz
2. Both new TX and RX need to first go through "Light Sleep→VCO Cal.→Standby"



WOT Process

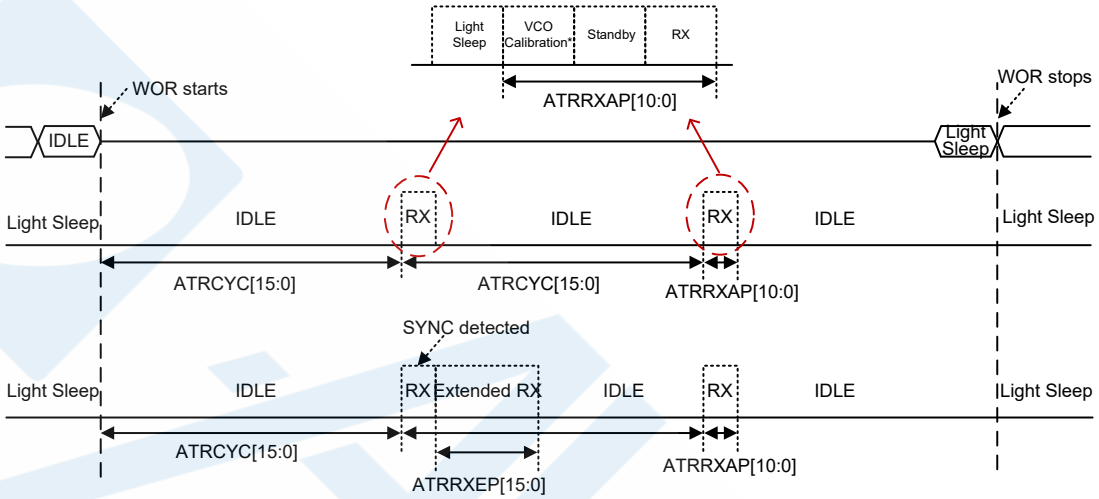
WOR (Wake-On-RX) Function

When the WOR function is enabled by setting the ATR_EN bit to 1 and the ATRM[1:0] bits to 01b, the RF transceiver will periodically wake up from the Idle mode and listen for the incoming packets without interaction with the host MCU. The RF transceiver starts the WOR process upon receiving the Idle strobe command from the MCU and stops the WOR process upon receiving the Light Sleep strobe command from the MCU. The ATRCYC[15:0] bits are used to set the wake-up period for the WOR function. At the moment of ATR timer expiration, the wake-up timer will trigger the RF transceiver to leave the Idle mode and enter the active state to listen for the incoming packet, at the same time the ATRCYC[15:0] content will be reloaded into the timer's counter. The receiving active period is defined by the ATRRXAP[10:0] bits. The active period is in multiples of 250μs (defalut) and starts from 250μs. If there is no incoming packet received in the RX active period, the RF transceiver will return to the Idle mode and wait for the next WOR cycle.

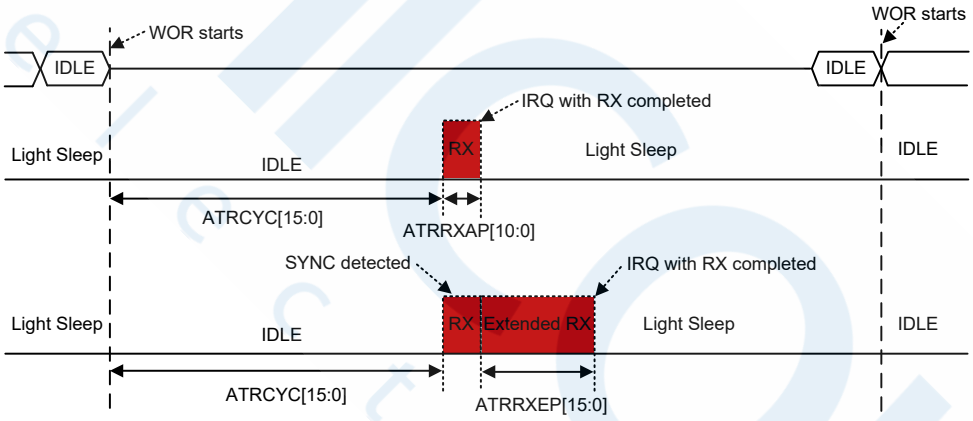
The active period is auto-extended when the "preamble+SYNCWORD" is detected. The extend period is defined by (ATRRXEP[15:0]+1). The extend period is also in multiples of 250μs (defalut) and starts from 250μs. Once the SYNCWORD is received, the receiving period would be auto-extended until the whole packet is completely received. After the RX receiving is done with CRC checked correct, the RF transceiver would acknowledge the MCU with RX complete IRQ and stay

at light sleep mode. MCU can read the incoming packet from the RX FIFO and then restart the next WOR session by issuing Idle strobe command. If MCU wants to leave WOR mode, MCU still needs to issue Light Sleep command to the RF transceiver.

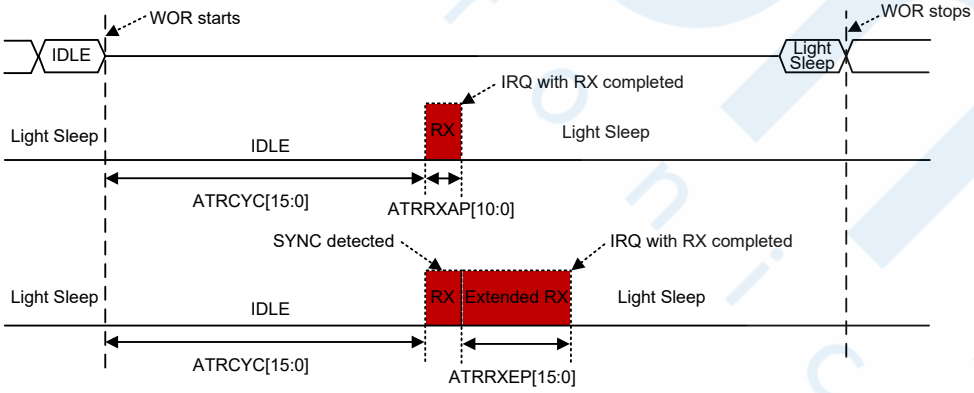
Note: 1. VCO Calibration time: ~152μs@433MHz / ~96μs@868MHz
2. Both new RX and TX need to first go through"Light Sleep→VCO Cal.→Standby"



WOR Without Incoming Packet Received

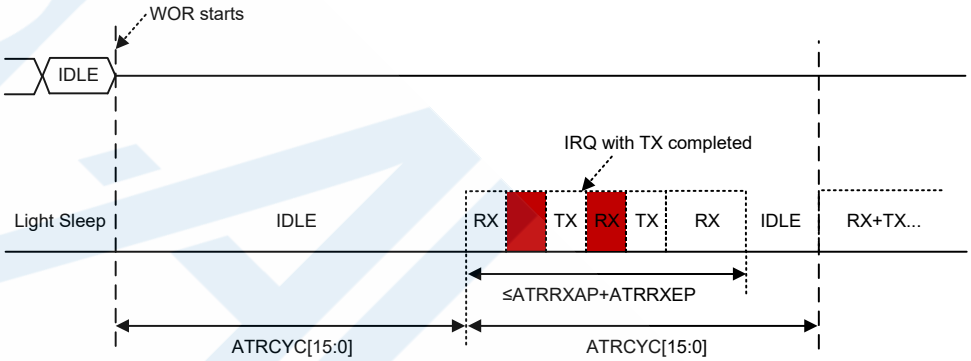


WOR With Incoming Packet Received



WOR Stops after Receiving Incoming Packet

In the WOR active period, the RF transceiver only implements RX operation once by default. Users can extend the wake-up receiving mechanism by combining with the ARK function. In WOR+ARK mode, the time duration between the repeated receiving packets is inserted with one TX slot for acknowledgement. The TX duration depends on the transmitting data rate. The RF transceiver stays in the RX mode for a maximum period of time defined by ATRRXAP+ATRRXEP. If a valid imcoming packet, with CRC checked correct and a different PID/CRC, is received before the timer expires, the RF transceiver will issue an RX completion IRQ to the MCU and automatically enter the TX mode. If a repeated packet, with CRC checked correct and a same PID/CRC, is received, the RF transceiver will only automatically enter the TX mode with no IRQ to the MCU. After the TX completion, the RF transceiver will return to the RX mode again and listen for the incoming packets until the timer expires if no incoming packet is received.



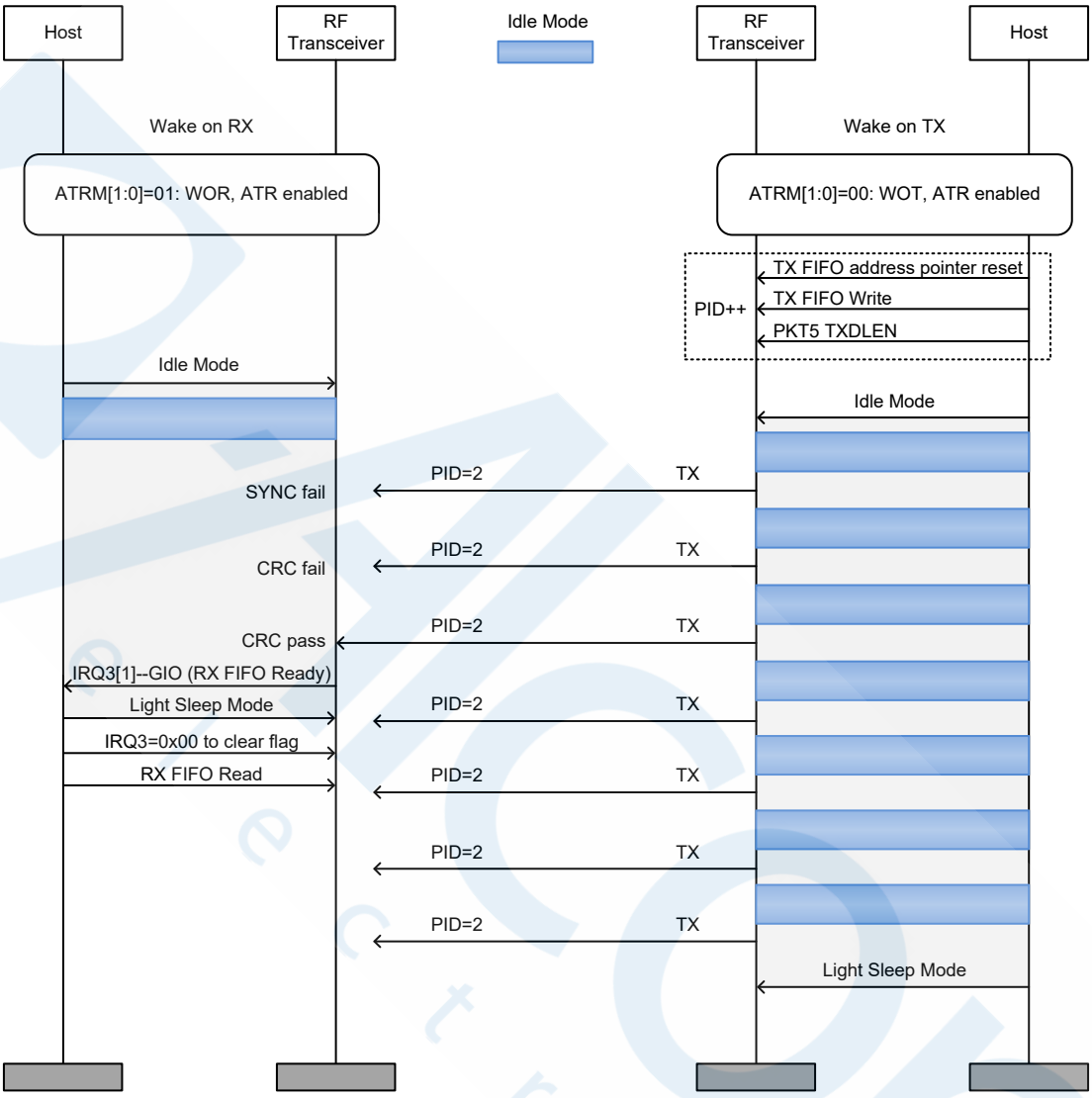
WOR+ARK Process

WTM (Wake-up Timer Mode)

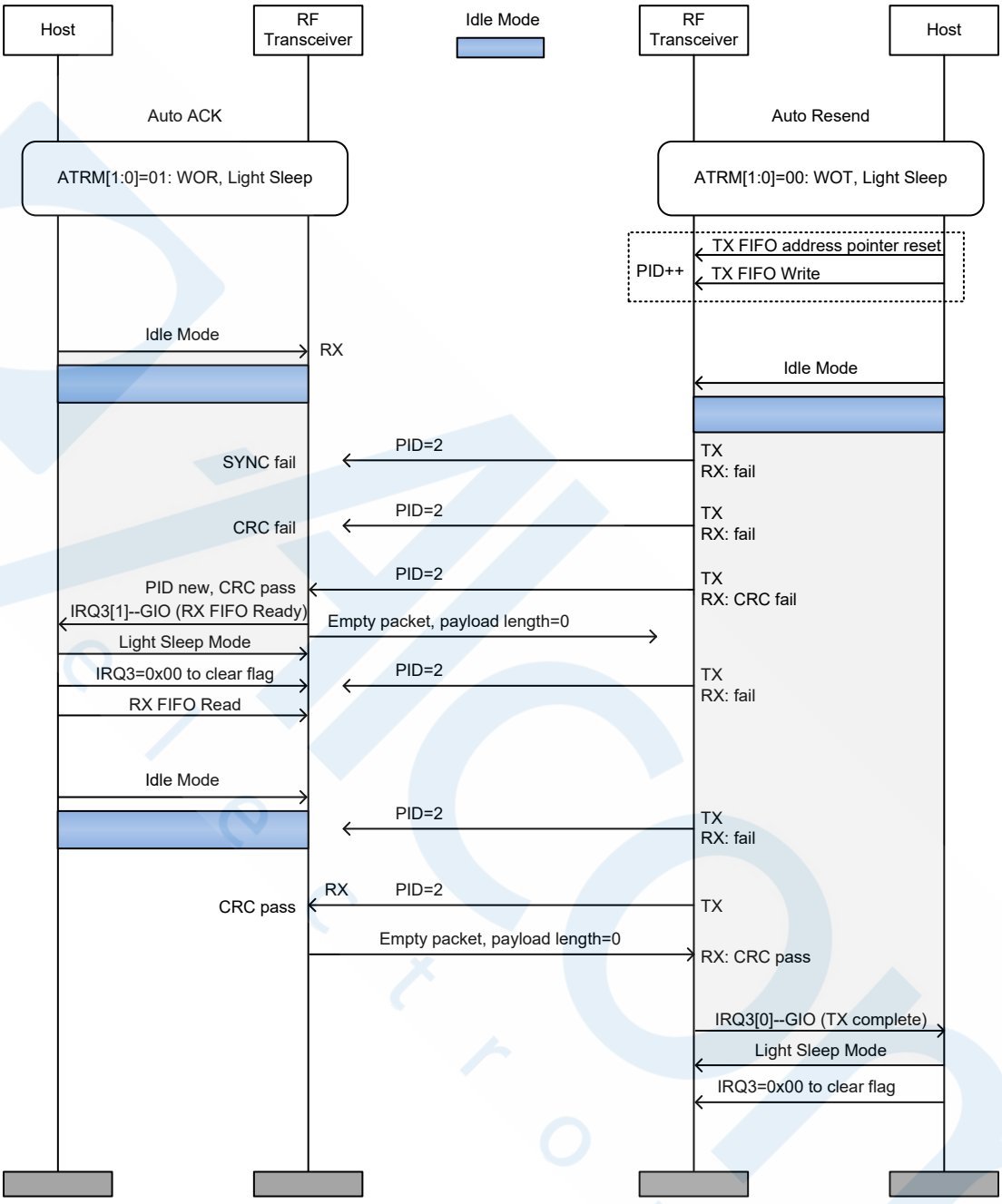
The RF transceiver can be set as a programmable timer to output a periodical waveform on GIOs. User can use this signal to wake up the CPU. Set ATR_EN=1 and ATRM=10b/11b to enable the WTM mode. The RF transceiver starts the WTM mode upon receiving the Idle strobe command from the MCU and stops the WTM mode upon receiving the Light Sleep strobe command. The RF transceiver will stay in the Idle mode for the whole WTM process.

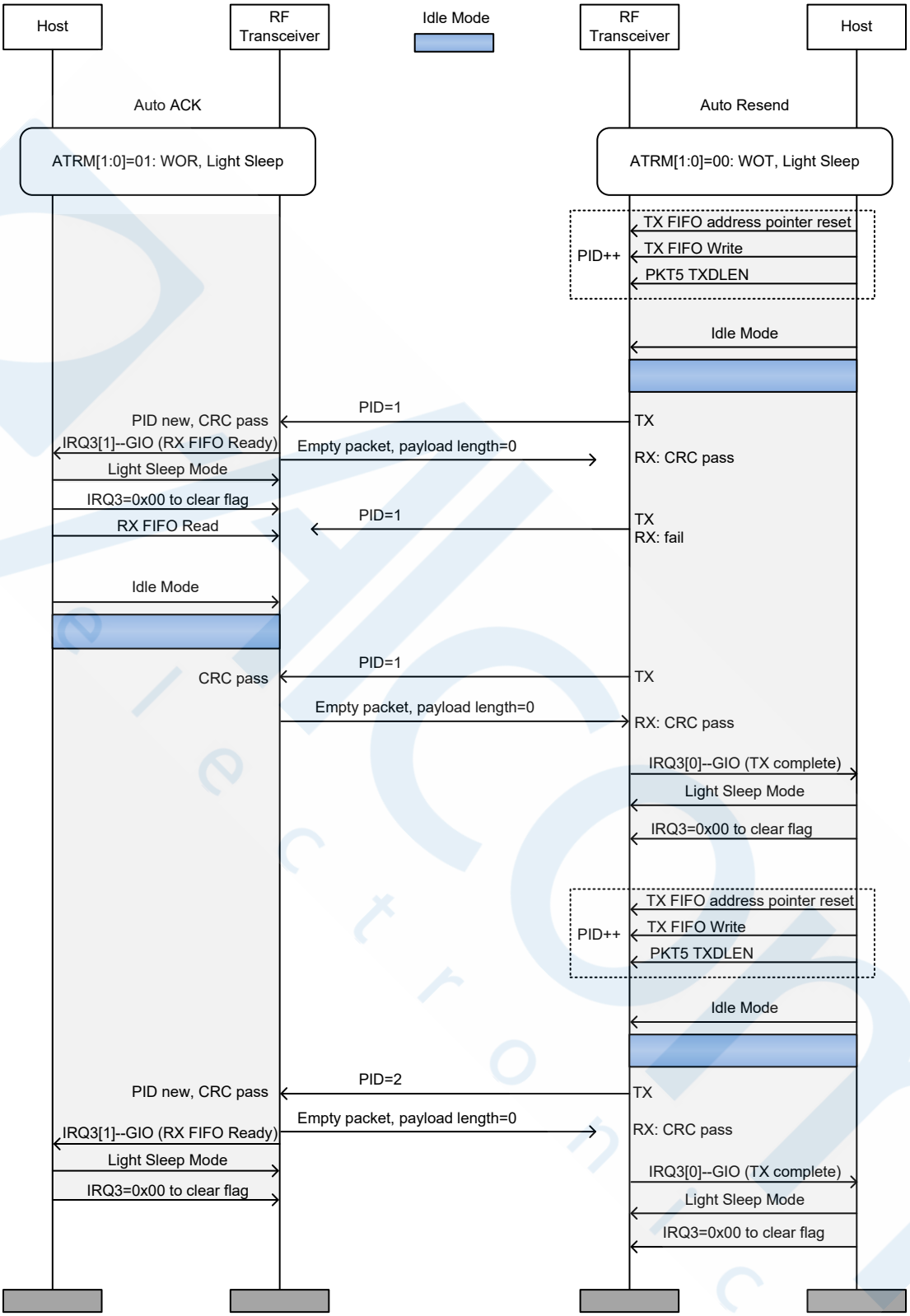
Message Flowchart Examples

ATR: WOT & WOR



ATR+ARK: WOT+Auto-Resend & WOR+Auto-Ack




Host
RF Transceiver
Idle Mode
RF Transceiver
Host
Auto ACK
ATRM[1:0]=01: WOR, Light Sleep
Auto Resend
ATRM[1:0]=00: WOT, Light Sleep
TX FIFO address pointer reset
PID++
TX FIFO Write
PKT5 TXDLEN
Idle Mode
PID=1
PID new, CRC pass
TX
IRQ3[1]--GIO (RX FIFO Ready)
Empty packet, payload length=0
RX: CRC pass
Light Sleep Mode
IRQ3=0x00 to clear flag
PID=1
RX FIFO Read
TX
RX: fail
Idle Mode
PID=1
CRC pass
TX
Empty packet, payload length=0
RX: CRC pass
IRQ3[0]--GIO (TX complete)
Light Sleep Mode
IRQ3=0x00 to clear flag
TX FIFO address pointer reset
PID++
TX FIFO Write
PKT5 TXDLEN
Idle Mode
PID=2
PID new, CRC pass
TX
Empty packet, payload length=0
IRQ3[1]--GIO (RX FIFO Ready)
RX: CRC pass
Light Sleep Mode
IRQ3[0]--GIO (TX complete)
IRQ3=0x00 to clear flag
Light Sleep Mode
IRQ3=0x00 to clear flag

Abbreviation

ADC: Analog to Digital Converter

AFC: Automatic Frequency Compensation

AGC: Automatic Gain Control

ARK: Auto-Resend and Auto-Ack

ATR: Automatic-Transmit-Receive

BER: Bit Error Rate

BPF: Band Pass Filter

CD: Carrier Detect

CFO: Carrier Frequency Offset

CP: Charge Pump

CRC: Cyclic Redundancy Check

DCOC: DC Offset Correct

DSM: Delta Sigma Modulator

FEC: Forward Error Correction

FIFO: First In First Out

GFSK: Gaussian Frequency Shift Keying

HPF: High Pass Filter

ID: Identifier

IF: Intermedia Frequency

IIR: Infinite Impulse Response

IRQ: Interrupt Request

ISM: Industrial, Scientific and Medical

LNA: Low Noise Amplifier

LO: Local Oscillator

LPF: Low Pass Filter

MCU: Mico Controller Unit

MMD: Multi-Mode Divider

OW: Overwrite

PA: Power Amplifier

PD: Power Down

PFD: Phase Frequency Detector (for PLL)

PLL: Phase Lock Loop

POR: Power On Reset

PVT: Process-Voltage-Temperature

RBCLK: RX Bit Clock

RSSI: Received Signal Strength Indicator

RX: Receiver

SNR: Signal Noise Ratio

SPI: Serial Port Interface

SX: Synthesizer

SYCK: System Clock for digital circuit

SYNC/SYNCWORD: Synchronization Word

TBCLK: TX Bit Clock

TRX: TX/RX

TX: Transmitter

VCO: Voltage Controlled Oscillator

WOR: Wake-on-RX

WOT: Wake-on-TX

WTM: Wake-up Timer Mode

XCLK: Crystal Clock

XO/XOSC: Crystal Oscillator

XTAL: Crystal

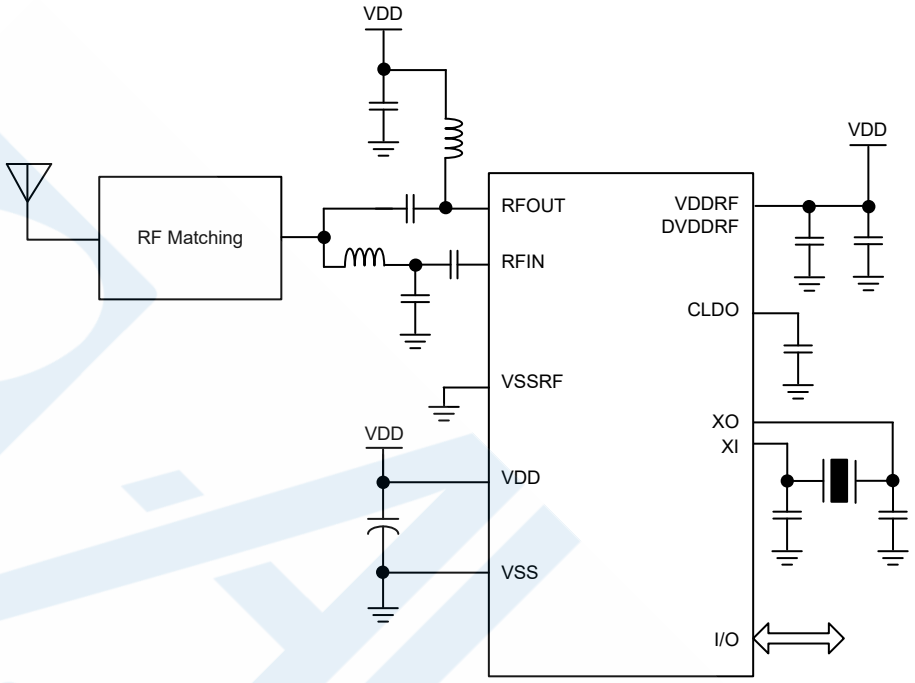
Configuration Options

Configuration options refer to certain options within the MCU that are programmed into the device during the programming process. During the development process, these options are selected using the HT-IDE software development tools. The option must be defined for proper system function, the details of which are shown in the table.

No.	Options
Oscillator Option	
1	HIRC frequency selection – f_{HIRC}: 8MHz, 12MHz or 16MHz

Note: When the HIRC has been configured at a frequency shown in this table, the HIRC1 and HIRC0 bits should also be set to select the same frequency to achieve the HIRC frequency accuracy specified in the A.C. Characteristics.

Application Circuits



Instruction Set

Introduction

Central to the successful operation of any microcontroller is its instruction set, which is a set of program instruction codes that directs the microcontroller to perform certain operations. In the case of Holtek microcontroller, a comprehensive and flexible set of over 60 instructions is provided to enable programmers to implement their application with the minimum of programming overheads.

For easier understanding of the various instruction codes, they have been subdivided into several functional groupings.

Instruction Timing

Most instructions are implemented within one instruction cycle. The exceptions to this are branch, call, or table read instructions where two instruction cycles are required. One instruction cycle is equal to 4 system clock cycles, therefore in the case of an 8MHz system oscillator, most instructions would be implemented within 0.5μs and branch or call instructions would be implemented within 1μs. Although instructions which require one more cycle to implement are generally limited to the JMP, CALL, RET, RETI and table read instructions, it is important to realize that any other instructions which involve manipulation of the Program Counter Low register or PCL will also take one more cycle to implement. As instructions which change the contents of the PCL will imply a direct jump to that new address, one more cycle will be required. Examples of such instructions would be "CLR PCL" or "MOV PCL, A". For the case of skip instructions, it must be noted that if the result of the comparison involves a skip operation then this will also take one more cycle, if no skip is involved then only one cycle is required.

Moving and Transferring Data

The transfer of data within the microcontroller program is one of the most frequently used operations. Making use of several kinds of MOV instructions, data can be transferred from registers to the Accumulator and vice-versa as well as being able to move specific immediate data directly into the Accumulator. One of the most important data transfer applications is to receive data from the input ports and transfer data to the output ports.

Arithmetic Operations

The ability to perform certain arithmetic operations and data manipulation is a necessary feature of most microcontroller applications. Within the Holtek microcontroller instruction set are a range of add and subtract instruction mnemonics to enable the necessary arithmetic to be carried out. Care must be taken to ensure correct handling of carry and borrow data when results exceed 255 for addition and less than 0 for subtraction. The increment and decrement instructions such as INC, INCA, DEC and DECA provide a simple means of increasing or decreasing by a value of one of the values in the destination specified.

Logical and Rotate Operation

The standard logical operations such as AND, OR, XOR and CPL all have their own instruction within the Holtek microcontroller instruction set. As with the case of most instructions involving data manipulation, data must pass through the Accumulator which may involve additional programming steps. In all logical data operations, the zero flag may be set if the result of the operation is zero. Another form of logical data manipulation comes from the rotate instructions such as RR, RL, RRC and RLC which provide a simple means of rotating one bit right or left. Different rotate instructions exist depending on program requirements. Rotate instructions are useful for serial port programming applications where data can be rotated from an internal register into the Carry bit from where it can be examined and the necessary serial bit set high or low. Another application which rotate data operations are used is to implement multiplication and division calculations.

Branches and Control Transfer

Program branching takes the form of either jumps to specified locations using the JMP instruction or to a subroutine using the CALL instruction. They differ in the sense that in the case of a subroutine call, the program must return to the instruction immediately when the subroutine has been carried out. This is done by placing a return instruction "RET" in the subroutine which will cause the program to jump back to the address right after the CALL instruction. In the case of a JMP instruction, the program simply jumps to the desired location. There is no requirement to jump back to the original jumping off point as in the case of the CALL instruction. One special and extremely useful set of branch instructions are the conditional branches. Here a decision is first made regarding the condition of a certain data memory or individual bits. Depending upon the conditions, the program will continue with the next instruction or skip over it and jump to the following instruction. These instructions are the key to decision making and branching within the program perhaps determined by the condition of certain input switches or by the condition of internal data bits.

Bit Operations

The ability to provide single bit operations on Data Memory is an extremely flexible feature of all Holtek microcontrollers. This feature is especially useful for output port bit programming where individual bits or port pins can be directly set high or low using either the "SET [m].i" or "CLR [m].i" instructions respectively. The feature removes the need for programmers to first read the 8-bit output port, manipulate the input data to ensure that other bits are not changed and then output the port with the correct new data. This read-modify-write process is taken care of automatically when these bit operation instructions are used.

Table Read Operations

Data storage is normally implemented by using registers. However, when working with large amounts of fixed data, the volume involved often makes it inconvenient to store the fixed data in the Data Memory. To overcome this problem, Holtek microcontrollers allow an area of Program Memory to be setup as a table where data can be directly stored. A set of easy to use instructions provides the means by which this fixed data can be referenced and retrieved from the Program Memory.

Other Operations

In addition to the above functional instructions, a range of other instructions also exist such as the "HALT" instruction for Power-down operations and instructions to control the operation of the Watchdog Timer for reliable program operations under extreme electric or electromagnetic environments. For their relevant operations, refer to the functional related sections.

Instruction Set Summary

The instructions related to the data memory access in the following table can be used when the desired data memory is located in Data Memory sector 0.

Table Conventions

x: Bits immediate data
m: Data Memory address
A: Accumulator
i: 0~7 number of bits
addr: Program memory address

Mnemonic	Description	Cycles	Flag Affected
Arithmetic			
ADD A,[m]	Add Data Memory to ACC	1	Z, C, AC, OV, SC
ADDM A,[m]	Add ACC to Data Memory	1Note	Z, C, AC, OV, SC
ADD A,x	Add immediate data to ACC	1	Z, C, AC, OV, SC
ADC A,[m]	Add Data Memory to ACC with Carry	1	Z, C, AC, OV, SC
ADCM A,[m]	Add ACC to Data memory with Carry	1Note	Z, C, AC, OV, SC
SUB A,x	Subtract immediate data from the ACC	1	Z, C, AC, OV, SC, CZ
SUB A,[m]	Subtract Data Memory from ACC	1	Z, C, AC, OV, SC, CZ
SUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory	1Note	Z, C, AC, OV, SC, CZ
SBC A,x	Subtract immediate data from ACC with Carry	1	Z, C, AC, OV, SC, CZ
SBC A,[m]	Subtract Data Memory from ACC with Carry	1	Z, C, AC, OV, SC, CZ
SBCM A,[m]	Subtract Data Memory from ACC with Carry, result in Data Memory	1Note	Z, C, AC, OV, SC, CZ
DAA [m]	Decimal adjust ACC for Addition with result in Data Memory	1Note	C
Logic Operation			
AND A,[m]	Logical AND Data Memory to ACC	1	Z
OR A,[m]	Logical OR Data Memory to ACC	1	Z
XOR A,[m]	Logical XOR Data Memory to ACC	1	Z
ANDM A,[m]	Logical AND ACC to Data Memory	1Note	Z
ORM A,[m]	Logical OR ACC to Data Memory	1Note	Z
XORM A,[m]	Logical XOR ACC to Data Memory	1Note	Z
AND A,x	Logical AND immediate Data to ACC	1	Z
OR A,x	Logical OR immediate Data to ACC	1	Z
XOR A,x	Logical XOR immediate Data to ACC	1	Z
CPL [m]	Complement Data Memory	1Note	Z
CPLA [m]	Complement Data Memory with result in ACC	1	Z
Increment & Decrement			
INCA [m]	Increment Data Memory with result in ACC	1	Z
INC [m]	Increment Data Memory	1Note	Z
DECA [m]	Decrement Data Memory with result in ACC	1	Z
DEC [m]	Decrement Data Memory	1Note	Z
Rotate			
RRA [m]	Rotate Data Memory right with result in ACC	1	None
RR [m]	Rotate Data Memory right	1Note	None
RRCA [m]	Rotate Data Memory right through Carry with result in ACC	1	C
RRC [m]	Rotate Data Memory right through Carry	1Note	C
RLA [m]	Rotate Data Memory left with result in ACC	1	None
RL [m]	Rotate Data Memory left	1Note	None
RLCA [m]	Rotate Data Memory left through Carry with result in ACC	1	C
RLC [m]	Rotate Data Memory left through Carry	1Note	C

Mnemonic	Description	Cycles	Flag Affected
Data Move			
MOV A,[m]	Move Data Memory to ACC	1	None
MOV [m],A	Move ACC to Data Memory	1Note	None
MOV A,x	Move immediate data to ACC	1	None
Bit Operation			
CLR [m].i	Clear bit of Data Memory	1Note	None
SET [m].i	Set bit of Data Memory	1Note	None
Branch Operation			
JMP addr	Jump unconditionally	2	None
SZ [m]	Skip if Data Memory is zero	1Note	None
SZA [m]	Skip if Data Memory is zero with data movement to ACC	1Note	None
SZ [m].i	Skip if bit i of Data Memory is zero	1Note	None
SNZ [m]	Skip if Data Memory is not zero	1Note	None
SNZ [m].i	Skip if bit i of Data Memory is not zero	1Note	None
SIZ [m]	Skip if increment Data Memory is zero	1Note	None
SDZ [m]	Skip if decrement Data Memory is zero	1Note	None
SIZA [m]	Skip if increment Data Memory is zero with result in ACC	1Note	None
SDZA [m]	Skip if decrement Data Memory is zero with result in ACC	1Note	None
CALL addr	Subroutine call	2	None
RET	Return from subroutine	2	None
RET A,x	Return from subroutine and load immediate data to ACC	2	None
RETI	Return from interrupt	2	None
Table Read Operation			
TABRD [m]	Read table (specific page) to TBLH and Data Memory	2Note	None
TABRDL [m]	Read table (last page) to TBLH and Data Memory	2Note	None
ITABRD [m]	Increment table pointer TBLP first and Read table (specific page) to TBLH and Data Memory	2Note	None
ITABRDL [m]	Increment table pointer TBLP first and Read table (last page) to TBLH and Data Memory	2Note	None
Miscellaneous			
NOP	No operation	1	None
CLR [m]	Clear Data Memory	1Note	None
SET [m]	Set Data Memory	1Note	None
CLR WDT	Clear Watchdog Timer	1	TO, PDF
SWAP [m]	Swap nibbles of Data Memory	1Note	None
SWAPA [m]	Swap nibbles of Data Memory with result in ACC	1	None
HALT	Enter power down mode	1	TO, PDF

Note: 1. For skip instructions, if the result of the comparison involves a skip then two cycles are required, if no skip takes place only one cycle is required.

2. Any instruction which changes the contents of the PCL will also require 2 cycles for execution.

Extended Instruction Set

The extended instructions are used to support the full range address access for the data memory. When the accessed data memory is located in any data memory sector except sector 0, the extended instruction can be used to directly access the data memory instead of using the indirect addressing access. This can not only reduce the use of Flash memory space but also improve the CPU execution efficiency.

Mnemonic	Description	Cycles	Flag Affected
Arithmetic			
LADD A,[m]	Add Data Memory to ACC	2	Z, C, AC, OV, SC
LADDM A,[m]	Add ACC to Data Memory	2Note	Z, C, AC, OV, SC
LADC A,[m]	Add Data Memory to ACC with Carry	2	Z, C, AC, OV, SC
LADCM A,[m]	Add ACC to Data memory with Carry	2Note	Z, C, AC, OV, SC
LSUB A,[m]	Subtract Data Memory from ACC	2	Z, C, AC, OV, SC, CZ
LSUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory	2Note	Z, C, AC, OV, SC, CZ
LSBC A,[m]	Subtract Data Memory from ACC with Carry	2	Z, C, AC, OV, SC, CZ
LSBCM A,[m]	Subtract Data Memory from ACC with Carry, result in Data Memory	2Note	Z, C, AC, OV, SC, CZ
LDAA [m]	Decimal adjust ACC for Addition with result in Data Memory	2Note	C
Logic Operation			
LAND A,[m]	Logical AND Data Memory to ACC	2	Z
LOR A,[m]	Logical OR Data Memory to ACC	2	Z
LXOR A,[m]	Logical XOR Data Memory to ACC	2	Z
LANDM A,[m]	Logical AND ACC to Data Memory	2Note	Z
LORM A,[m]	Logical OR ACC to Data Memory	2Note	Z
LXORM A,[m]	Logical XOR ACC to Data Memory	2Note	Z
LCPL [m]	Complement Data Memory	2Note	Z
LCPLA [m]	Complement Data Memory with result in ACC	2	Z
Increment & Decrement			
LINCA [m]	Increment Data Memory with result in ACC	2	Z
LINC [m]	Increment Data Memory	2Note	Z
LDECA [m]	Decrement Data Memory with result in ACC	2	Z
LDEC [m]	Decrement Data Memory	2Note	Z
Rotate			
LRRA [m]	Rotate Data Memory right with result in ACC	2	None
LRR [m]	Rotate Data Memory right	2Note	None
LRRCA [m]	Rotate Data Memory right through Carry with result in ACC	2	C
LRRC [m]	Rotate Data Memory right through Carry	2Note	C
LRLA [m]	Rotate Data Memory left with result in ACC	2	None
LRL [m]	Rotate Data Memory left	2Note	None
LRLCA [m]	Rotate Data Memory left through Carry with result in ACC	2	C
LRLC [m]	Rotate Data Memory left through Carry	2Note	C
Data Move			
LMOV A,[m]	Move Data Memory to ACC	2	None
LMOV [m],A	Move ACC to Data Memory	2Note	None
Bit Operation			
LCLR [m].i	Clear bit of Data Memory	2Note	None
LSET [m].i	Set bit of Data Memory	2Note	None

Mnemonic	Description	Cycles	Flag Affected
Branch			
LSZ [m]	Skip if Data Memory is zero	2Note	None
LSZA [m]	Skip if Data Memory is zero with data movement to ACC	2Note	None
LSNZ [m]	Skip if Data Memory is not zero	2Note	None
LSZ [m].i	Skip if bit i of Data Memory is zero	2Note	None
LSNZ [m].i	Skip if bit i of Data Memory is not zero	2Note	None
LSIZ [m]	Skip if increment Data Memory is zero	2Note	None
LSDZ [m]	Skip if decrement Data Memory is zero	2Note	None
LSIZA [m]	Skip if increment Data Memory is zero with result in ACC	2Note	None
LSDZA [m]	Skip if decrement Data Memory is zero with result in ACC	2Note	None
Table Read			
LTABRD [m]	Read table (specific page) to TBLH and Data Memory	3Note	None
LTABRDL [m]	Read table (last page) to TBLH and Data Memory	3Note	None
LITABRD [m]	Increment table pointer TBLP first and Read table (specific page) to TBLH and Data Memory	3Note	None
LITABRDL [m]	Increment table pointer TBLP first and Read table (last page) to TBLH and Data Memory	3Note	None
Miscellaneous			
LCLR [m]	Clear Data Memory	2Note	None
LSET [m]	Set Data Memory	2Note	None
LSWAP [m]	Swap nibbles of Data Memory	2Note	None
LSWAPA [m]	Swap nibbles of Data Memory with result in ACC	2	None

Note: 1. For these extended skip instructions, if the result of the comparison involves a skip then three cycles are required, if no skip takes place two cycles is required.

2. Any extended instruction which changes the contents of the PCL register will also require three cycles for execution.

Instruction Definition

ADC A,[m] Add Data Memory to ACC with Carry

Description: The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the Accumulator.

Operation: ACC ← ACC + [m] + C

Affected flag(s): OV, Z, AC, C, SC

ADCM A,[m] Add ACC to Data Memory with Carry

Description: The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the specified Data Memory.

Operation: [m] ← ACC + [m] + C

Affected flag(s): OV, Z, AC, C, SC

ADD A,[m] Add Data Memory to ACC

Description: The contents of the specified Data Memory and the Accumulator are added. The result is stored in the Accumulator.

Operation: ACC ← ACC + [m]

Affected flag(s): OV, Z, AC, C, SC

ADD A,x Add immediate data to ACC

Description: The contents of the Accumulator and the specified immediate data are added. The result is stored in the Accumulator.

Operation: ACC ← ACC + x

Affected flag(s): OV, Z, AC, C, SC

ADDM A,[m] Add ACC to Data Memory

Description: The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory.

Operation: [m] ← ACC + [m]

Affected flag(s): OV, Z, AC, C, SC

AND A,[m] Logical AND Data Memory to ACC

Description: Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator.

Operation: ACC ← ACC "AND" [m]

Affected flag(s): Z

AND A,x Logical AND immediate data to ACC

Description: Data in the Accumulator and the specified immediate data perform a bit wise logical AND operation. The result is stored in the Accumulator.

Operation: ACC ← ACC "AND" x

Affected flag(s): Z

ANDM A,[m] Logical AND ACC to Data Memory

Description: Data in the specified Data Memory and the Accumulator perform a bitwise logical AND operation. The result is stored in the Data Memory.

Operation: [m] ← ACC "AND" [m]

Affected flag(s): Z

CALL addr Subroutine call

Description: Unconditionally calls a subroutine at the specified address. The Program Counter then increments by 1 to obtain the address of the next instruction which is then pushed onto the stack. The specified address is then loaded and the program continues execution from this new address. As this instruction requires an additional operation, it is a two cycle instruction.

Operation: Stack ← Program Counter + 1
Program Counter ← addr

Affected flag(s): None

CLR [m] Clear Data Memory

Description: Each bit of the specified Data Memory is cleared to 0.

Operation: [m] ← 00H

Affected flag(s): None

CLR [m].i Clear bit of Data Memory

Description: Bit i of the specified Data Memory is cleared to 0.

Operation: [m].i ← 0

Affected flag(s): None

CLR WDT Clear Watchdog Timer

Description: The TO, PDF flags and the WDT are all cleared.

Operation: WDT cleared
TO ← 0
PDF ← 0

Affected flag(s): TO, PDF

CPL [m] Complement Data Memory

Description: Each bit of the specified Data Memory is logically complemented (1′s complement). Bits which previously contained a 1 are changed to 0 and vice versa.

Operation: [m] ← $\overline{[m]}$

Affected flag(s): Z

CPLA [m] Complement Data Memory with result in ACC

Description: Each bit of the specified Data Memory is logically complemented (1′s complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged.

Operation: ACC ← $\overline{[m]}$

Affected flag(s): Z

DAA [m] Decimal-Adjust ACC for addition with result in Data Memory

Description: Convert the contents of the Accumulator value to a BCD (Binary Coded Decimal) value resulting from the previous addition of two BCD variables. If the low nibble is greater than 9 or if AC flag is set, then a value of 6 will be added to the low nibble. Otherwise the low nibble remains unchanged. If the high nibble is greater than 9 or if the C flag is set, then a value of 6 will be added to the high nibble. Essentially, the decimal conversion is performed by adding 00H, 06H, 60H or 66H depending on the Accumulator and flag conditions. Only the C flag may be affected by this instruction which indicates that if the original BCD sum is greater than 100, it allows multiple precision decimal addition.

Operation: [m] ← ACC + 00H or
[m] ← ACC + 06H or
[m] ← ACC + 60H or
[m] ← ACC + 66H

Affected flag(s): C

DEC [m] Decrement Data Memory

Description Data in the specified Data Memory is decremented by 1.

Operation [m] ← [m] − 1

Affected flag(s) Z

DECA [m] Decrement Data Memory with result in ACC

Description Data in the specified Data Memory is decremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.

Operation ACC ← [m] − 1

Affected flag(s) Z

HALT Enter power down mode

Description This instruction stops the program execution and turns off the system clock. The contents of the Data Memory and registers are retained. The WDT and prescaler are cleared. The power down flag PDF is set and the WDT time-out flag TO is cleared.

Operation TO ← 0
PDF ← 1

Affected flag(s) TO, PDF

INC [m] Increment Data Memory

Description Data in the specified Data Memory is incremented by 1.

Operation [m] ← [m] + 1

Affected flag(s) Z

INCA [m] Increment Data Memory with result in ACC

Description Data in the specified Data Memory is incremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.

Operation ACC ← [m] + 1

Affected flag(s) Z

JMP addr Jump unconditionally

Description The contents of the Program Counter are replaced with the specified address. Program execution then continues from this new address. As this requires the insertion of a dummy instruction while the new address is loaded, it is a two cycle instruction.

Operation Program Counter ← addr

Affected flag(s) None

MOV A,[m] Move Data Memory to ACC

Description The contents of the specified Data Memory are copied to the Accumulator.

Operation ACC ← [m]

Affected flag(s) None

MOV A,x Move immediate data to ACC

Description The immediate data specified is loaded into the Accumulator.

Operation ACC ← x

Affected flag(s) None

MOV [m],A Move ACC to Data Memory

Description The contents of the Accumulator are copied to the specified Data Memory.

Operation [m] ← ACC

Affected flag(s) None

NOP No operation

Description No operation is performed. Execution continues with the next instruction.

Operation No operation

Affected flag(s) None

OR A,[m] Logical OR Data Memory to ACC

Description Data in the Accumulator and the specified Data Memory perform a bitwise logical OR operation. The result is stored in the Accumulator.

Operation ACC ← ACC "OR" [m]

Affected flag(s) Z

OR A,x Logical OR immediate data to ACC

Description Data in the Accumulator and the specified immediate data perform a bitwise logical OR operation. The result is stored in the Accumulator.

Operation ACC ← ACC "OR" x

Affected flag(s) Z

ORM A,[m] Logical OR ACC to Data Memory

Description Data in the specified Data Memory and the Accumulator perform a bitwise logical OR operation. The result is stored in the Data Memory.

Operation [m] ← ACC "OR" [m]

Affected flag(s) Z

RET Return from subroutine

Description The Program Counter is restored from the stack. Program execution continues at the restored address.

Operation Program Counter ← Stack

Affected flag(s) None

RET A,x Return from subroutine and load immediate data to ACC

Description The Program Counter is restored from the stack and the Accumulator loaded with the specified immediate data. Program execution continues at the restored address.

Operation Program Counter ← Stack
ACC ← x

Affected flag(s) None

RETI Return from interrupt

Description The Program Counter is restored from the stack and the interrupts are re-enabled by setting the EMI bit. EMI is the master interrupt global enable bit. If an interrupt was pending when the RETI instruction is executed, the pending Interrupt routine will be processed before returning to the main program.

Operation Program Counter ← Stack
EMI ← 1

Affected flag(s) None

RL [m] Rotate Data Memory left

Description The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0.

Operation [m].(i+1) ← [m].i; (i=0~6)
[m].0 ← [m].7

Affected flag(s) None

RLA [m] Rotate Data Memory left with result in ACC

Description: The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.

Operation:
ACC.(i+1) ← [m].i; (i=0~6)
ACC.0 ← [m].7

Affected flag(s): None

RLC [m] Rotate Data Memory left through Carry

Description: The contents of the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into bit 0.

Operation:
[m].(i+1) ← [m].i; (i=0~6)
[m].0 ← C
C ← [m].7

Affected flag(s): C

RLCA [m] Rotate Data Memory left through Carry with result in ACC

Description: Data in the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into the bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.

Operation:
ACC.(i+1) ← [m].i; (i=0~6)
ACC.0 ← C
C ← [m].7

Affected flag(s): C

RR [m] Rotate Data Memory right

Description: The contents of the specified Data Memory are rotated right by 1 bit with bit 0 rotated into bit 7.

Operation:
[m].i ← [m].(i+1); (i=0~6)
[m].7 ← [m].0

Affected flag(s): None

RRA [m] Rotate Data Memory right with result in ACC

Description: Data in the specified Data Memory is rotated right by 1 bit with bit 0 rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.

Operation:
ACC.i ← [m].(i+1); (i=0~6)
ACC.7 ← [m].0

Affected flag(s): None

RRC [m] Rotate Data Memory right through Carry

Description: The contents of the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7.

Operation:
[m].i ← [m].(i+1); (i=0~6)
[m].7 ← C
C ← [m].0

Affected flag(s): C

RRCA [m] Rotate Data Memory right through Carry with result in ACC

Description: Data in the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.

Operation:
ACC.i ← [m].(i+1); (i=0~6)
ACC.7 ← C
C ← [m].0

Affected flag(s): C

SBC A,[m] Subtract Data Memory from ACC with Carry

Description: The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.

Operation: $ACC \leftarrow ACC - [m] - \overline{C}$

Affected flag(s): OV, Z, AC, C, SC, CZ

SBC A, x Subtract immediate data from ACC with Carry

Description: The immediate data and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.

Operation: $ACC \leftarrow ACC - [m] - \overline{C}$

Affected flag(s): OV, Z, AC, C, SC, CZ

SBCM A,[m] Subtract Data Memory from ACC with Carry and result in Data Memory

Description: The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.

Operation: $[m] \leftarrow ACC - [m] - \overline{C}$

Affected flag(s): OV, Z, AC, C, SC, CZ

SDZ [m] Skip if decrement Data Memory is 0

Description: The contents of the specified Data Memory are first decremented by 1. If the result is 0 the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.

Operation:
[m] ← [m] − 1
Skip if [m]=0

Affected flag(s): None

SDZA [m] Skip if decrement Data Memory is zero with result in ACC

Description: The contents of the specified Data Memory are first decremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction.

Operation:
ACC ← [m] − 1
Skip if ACC=0

Affected flag(s): None

SET [m] Set Data Memory

Description Each bit of the specified Data Memory is set to 1.

Operation [m] ← FFH

Affected flag(s) None

SET [m].i Set bit of Data Memory

Description Bit i of the specified Data Memory is set to 1.

Operation [m].i ← 1

Affected flag(s) None

SIZ [m] Skip if increment Data Memory is 0

Description The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.

Operation [m] ← [m] + 1
Skip if [m]=0

Affected flag(s) None

SIZA [m] Skip if increment Data Memory is zero with result in ACC

Description The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.

Operation ACC ← [m] + 1
Skip if ACC=0

Affected flag(s) None

SNZ [m].i Skip if bit i of Data Memory is not 0

Description If bit i of the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction.

Operation Skip if [m].i ≠ 0

Affected flag(s) None

SNZ [m] Skip if Data Memory is not 0

Description The contents of the specified Data Memory are read out and then written back to the specified Data Memory again. If the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction.

Operation Skip if [m]≠ 0

Affected flag(s) None

SUB A,[m] Subtract Data Memory from ACC

Description The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.

Operation ACC ← ACC − [m]

Affected flag(s) OV, Z, AC, C, SC, CZ

SUBM A,[m] Subtract Data Memory from ACC with result in Data Memory

Description The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.

Operation [m] ← ACC − [m]

Affected flag(s) OV, Z, AC, C, SC, CZ

SUB A,x Subtract immediate data from ACC

Description The immediate data specified by the code is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.

Operation ACC ← ACC − x

Affected flag(s) OV, Z, AC, C, SC, CZ

SWAP [m] Swap nibbles of Data Memory

Description The low-order and high-order nibbles of the specified Data Memory are interchanged.

Operation [m].3~[m].0 ↔ [m].7~[m].4

Affected flag(s) None

SWAPA [m] Swap nibbles of Data Memory with result in ACC

Description The low-order and high-order nibbles of the specified Data Memory are interchanged. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.

Operation ACC.3~ACC.0 ← [m].7~[m].4
ACC.7~ACC.4 ← [m].3~[m].0

Affected flag(s) None

SZ [m] Skip if Data Memory is 0

Description The contents of the specified Data Memory are read out and then written back to the specified Data Memory again. If the contents of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.

Operation Skip if [m]=0

Affected flag(s) None

SZA [m] Skip if Data Memory is 0 with data movement to ACC

Description The contents of the specified Data Memory are copied to the Accumulator. If the value is zero, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.

Operation ACC ← [m]
Skip if [m]=0

Affected flag(s) None

SZ [m].i Skip if bit i of Data Memory is 0

Description If bit i of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction.

Operation Skip if [m].i=0

Affected flag(s) None

TABRD [m] Read table (specific page) to TBLH and Data Memory

Description: The low byte of the program code (specific page) addressed by the table pointer (TBLP and TBHP) is moved to the specified Data Memory and the high byte moved to TBLH.

Operation: [m] ← program code (low byte)
TBLH ← program code (high byte)

Affected flag(s): None

TABRDL [m] Read table (last page) to TBLH and Data Memory

Description: The low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.

Operation: [m] ← program code (low byte)
TBLH ← program code (high byte)

Affected flag(s): None

ITABRD [m] Increment table pointer low byte first and read table (specific page) to TBLH and Data Memory

Description: Increment table pointer low byte, TBLP, first and then the program code (specific page) addressed by the table pointer (TBHP and TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.

Operation: [m] ← program code (low byte)
TBLH ← program code (high byte)

Affected flag(s): None

ITABRDL [m] Increment table pointer low byte first and read table (last page) to TBLH and Data Memory

Description: Increment table pointer low byte, TBLP, first and then the low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.

Operation: [m] ← program code (low byte)
TBLH ← program code (high byte)

Affected flag(s): None

XOR A,[m] Logical XOR Data Memory to ACC

Description: Data in the Accumulator and the specified Data Memory perform a bitwise logical XOR operation. The result is stored in the Accumulator.

Operation: ACC ← ACC ″XOR″ [m]

Affected flag(s): Z

XORM A,[m] Logical XOR ACC to Data Memory

Description: Data in the specified Data Memory and the Accumulator perform a bitwise logical XOR operation. The result is stored in the Data Memory.

Operation: [m] ← ACC ″XOR″ [m]

Affected flag(s): Z

XOR A,x Logical XOR immediate data to ACC

Description: Data in the Accumulator and the specified immediate data perform a bitwise logical XOR operation. The result is stored in the Accumulator.

Operation: ACC ← ACC ″XOR″ x

Affected flag(s): Z

Extended Instruction Definition

The extended instructions are used to directly access the data stored in any data memory sections.

LADC A,[m] Add Data Memory to ACC with Carry

Description The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the Accumulator.

Operation ACC ← ACC + [m] + C

Affected flag(s) OV, Z, AC, C, SC

LADCM A,[m] Add ACC to Data Memory with Carry

Description The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the specified Data Memory.

Operation [m] ← ACC + [m] + C

Affected flag(s) OV, Z, AC, C, SC

LADD A,[m] Add Data Memory to ACC

Description The contents of the specified Data Memory and the Accumulator are added. The result is stored in the Accumulator.

Operation ACC ← ACC + [m]

Affected flag(s) OV, Z, AC, C, SC

LADDM A,[m] Add ACC to Data Memory

Description The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory.

Operation [m] ← ACC + [m]

Affected flag(s) OV, Z, AC, C, SC

LAND A,[m] Logical AND Data Memory to ACC

Description Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator.

Operation ACC ← ACC "AND" [m]

Affected flag(s) Z

LANDM A,[m] Logical AND ACC to Data Memory

Description Data in the specified Data Memory and the Accumulator perform a bitwise logical AND operation. The result is stored in the Data Memory.

Operation [m] ← ACC "AND" [m]

Affected flag(s) Z

LCLR [m] Clear Data Memory

Description Each bit of the specified Data Memory is cleared to 0.

Operation [m] ← 00H

Affected flag(s) None

LCLR [m].i Clear bit of Data Memory

Description Bit i of the specified Data Memory is cleared to 0.

Operation [m].i ← 0

Affected flag(s) None

LCPL [m] Complement Data Memory

Description Each bit of the specified Data Memory is logically complemented (1′s complement). Bits which previously contained a 1 are changed to 0 and vice versa.

Operation $[m] \leftarrow \overline{[m]}$

Affected flag(s) Z

LCPLA [m] Complement Data Memory with result in ACC

Description Each bit of the specified Data Memory is logically complemented (1′s complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged.

Operation $ACC \leftarrow \overline{[m]}$

Affected flag(s) Z

LDAA [m] Decimal-Adjust ACC for addition with result in Data Memory

Description Convert the contents of the Accumulator value to a BCD (Binary Coded Decimal) value resulting from the previous addition of two BCD variables. If the low nibble is greater than 9 or if AC flag is set, then a value of 6 will be added to the low nibble. Otherwise the low nibble remains unchanged. If the high nibble is greater than 9 or if the C flag is set, then a value of 6 will be added to the high nibble. Essentially, the decimal conversion is performed by adding 00H, 06H, 60H or 66H depending on the Accumulator and flag conditions. Only the C flag may be affected by this instruction which indicates that if the original BCD sum is greater than 100, it allows multiple precision decimal addition.

Operation $[m] \leftarrow ACC + 00H$ or
$[m] \leftarrow ACC + 06H$ or
$[m] \leftarrow ACC + 60H$ or
$[m] \leftarrow ACC + 66H$

Affected flag(s) C

LDEC [m] Decrement Data Memory

Description Data in the specified Data Memory is decremented by 1.

Operation $[m] \leftarrow [m] - 1$

Affected flag(s) Z

LDECA [m] Decrement Data Memory with result in ACC

Description Data in the specified Data Memory is decremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.

Operation $ACC \leftarrow [m] - 1$

Affected flag(s) Z

LINC [m] Increment Data Memory

Description Data in the specified Data Memory is incremented by 1.

Operation $[m] \leftarrow [m] + 1$

Affected flag(s) Z

LINCA [m] Increment Data Memory with result in ACC

Description Data in the specified Data Memory is incremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.

Operation $ACC \leftarrow [m] + 1$

Affected flag(s) Z

LMOV A,[m] Move Data Memory to ACC

Description The contents of the specified Data Memory are copied to the Accumulator.

Operation ACC ← [m]

Affected flag(s) None

LMOV [m],A Move ACC to Data Memory

Description The contents of the Accumulator are copied to the specified Data Memory.

Operation [m] ← ACC

Affected flag(s) None

LOR A,[m] Logical OR Data Memory to ACC

Description Data in the Accumulator and the specified Data Memory perform a bitwise logical OR operation. The result is stored in the Accumulator.

Operation ACC ← ACC "OR" [m]

Affected flag(s) Z

LORM A,[m] Logical OR ACC to Data Memory

Description Data in the specified Data Memory and the Accumulator perform a bitwise logical OR operation. The result is stored in the Data Memory.

Operation [m] ← ACC "OR" [m]

Affected flag(s) Z

LRL [m] Rotate Data Memory left

Description The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0.

Operation [m].(i+1) ← [m].i; (i=0~6)
[m].0 ← [m].7

Affected flag(s) None

LRLA [m] Rotate Data Memory left with result in ACC

Description The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.

Operation ACC.(i+1) ← [m].i; (i=0~6)
ACC.0 ← [m].7

Affected flag(s) None

LRLC [m] Rotate Data Memory left through Carry

Description The contents of the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into bit 0.

Operation [m].(i+1) ← [m].i; (i=0~6)
[m].0 ← C
C ← [m].7

Affected flag(s) C

LRLCA [m] Rotate Data Memory left through Carry with result in ACC

Description Data in the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into the bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.

Operation ACC.(i+1) ← [m].i; (i=0~6)
ACC.0 ← C
C ← [m].7

Affected flag(s) C

LRR [m] Rotate Data Memory right

Description: The contents of the specified Data Memory are rotated right by 1 bit with bit 0 rotated into bit 7.

Operation:
[m].i ← [m].(i+1); (i=0~6)
[m].7 ← [m].0

Affected flag(s): None

LRRA [m] Rotate Data Memory right with result in ACC

Description: Data in the specified Data Memory is rotated right by 1 bit with bit 0 rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.

Operation:
ACC.i ← [m].(i+1); (i=0~6)
ACC.7 ← [m].0

Affected flag(s): None

LRRC [m] Rotate Data Memory right through Carry

Description: The contents of the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7.

Operation:
[m].i ← [m].(i+1); (i=0~6)
[m].7 ← C
C ← [m].0

Affected flag(s): C

LRRCA [m] Rotate Data Memory right through Carry with result in ACC

Description: Data in the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.

Operation:
ACC.i ← [m].(i+1); (i=0~6)
ACC.7 ← C
C ← [m].0

Affected flag(s): C

LSBC A,[m] Subtract Data Memory from ACC with Carry

Description: The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.

Operation: $ACC \leftarrow ACC - [m] - \overline{C}$

Affected flag(s): OV, Z, AC, C, SC, CZ

LSBCM A,[m] Subtract Data Memory from ACC with Carry and result in Data Memory

Description: The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.

Operation: $[m] \leftarrow ACC - [m] - \overline{C}$

Affected flag(s): OV, Z, AC, C, SC, CZ

LSDZ [m] Skip if decrement Data Memory is 0

Description The contents of the specified Data Memory are first decremented by 1. If the result is 0 the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0 the program proceeds with the following instruction.

Operation [m] ← [m] − 1
Skip if [m]=0

Affected flag(s) None

LSDZA [m] Skip if decrement Data Memory is zero with result in ACC

Description The contents of the specified Data Memory are first decremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0, the program proceeds with the following instruction.

Operation ACC ← [m] − 1
Skip if ACC=0

Affected flag(s) None

LSET [m] Set Data Memory

Description Each bit of the specified Data Memory is set to 1.

Operation [m] ← FFH

Affected flag(s) None

LSET [m].i Set bit of Data Memory

Description Bit i of the specified Data Memory is set to 1.

Operation [m].i ← 1

Affected flag(s) None

LSIZ [m] Skip if increment Data Memory is 0

Description The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0 the program proceeds with the following instruction.

Operation [m] ← [m] + 1
Skip if [m]=0

Affected flag(s) None

LSIZA [m] Skip if increment Data Memory is zero with result in ACC

Description The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0 the program proceeds with the following instruction.

Operation ACC ← [m] + 1
Skip if ACC=0

Affected flag(s) None

LSNZ [m].i Skip if bit i of Data Memory is not 0

Description If bit i of the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is 0 the program proceeds with the following instruction.

Operation Skip if [m].i ≠ 0

Affected flag(s) None

LSNZ [m] Skip if Data Memory is not 0

Description The contents of the specified Data Memory are read out and then written to the specified Data Memory again. If the content of the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is 0 the program proceeds with the following instruction.

Operation Skip if [m] ≠ 0

Affected flag(s) None

LSUB A,[m] Subtract Data Memory from ACC

Description The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.

Operation ACC ← ACC − [m]

Affected flag(s) OV, Z, AC, C, SC, CZ

LSUBM A,[m] Subtract Data Memory from ACC with result in Data Memory

Description The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.

Operation [m] ← ACC − [m]

Affected flag(s) OV, Z, AC, C, SC, CZ

LSWAP [m] Swap nibbles of Data Memory

Description The low-order and high-order nibbles of the specified Data Memory are interchanged.

Operation [m].3~[m].0 ↔ [m].7~[m].4

Affected flag(s) None

LSWAPA [m] Swap nibbles of Data Memory with result in ACC

Description The low-order and high-order nibbles of the specified Data Memory are interchanged. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.

Operation ACC.3~ACC.0 ← [m].7~[m].4
ACC.7~ACC.4 ← [m].3~[m].0

Affected flag(s) None

LSZ [m] Skip if Data Memory is 0

Description The contents of the specified Data Memory are read out and then written to the specified Data Memory again. If the contents of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0 the program proceeds with the following instruction.

Operation Skip if [m]=0

Affected flag(s) None

LSZA [m] Skip if Data Memory is 0 with data movement to ACC

Description The contents of the specified Data Memory are copied to the Accumulator. If the value is zero, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0 the program proceeds with the following instruction.

Operation ACC ← [m]
Skip if [m]=0

Affected flag(s) None

LSZ [m].i Skip if bit i of Data Memory is 0

Description If bit i of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a three cycle instruction. If the result is not 0, the program proceeds with the following instruction.

Operation Skip if [m].i=0

Affected flag(s) None

LTABRD [m] Read table (specific page) to TBLH and Data Memory

Description The low byte of the program code (specific page) addressed by the table pointer (TBHP and TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.

Operation [m] ← program code (low byte)
TBLH ← program code (high byte)

Affected flag(s) None

LTABRDL [m] Read table (last page) to TBLH and Data Memory

Description The low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.

Operation [m] ← program code (low byte)
TBLH ← program code (high byte)

Affected flag(s) None

LITABRD [m] Increment table pointer low byte first and read table (specific page) to TBLH and Data Memory

Description Increment table pointer low byte, TBLP, first and then the program code (specific page) addressed by the table pointer (TBHP and TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.

Operation [m] ← program code (low byte)
TBLH ← program code (high byte)

Affected flag(s) None

LITABRDL [m] Increment table pointer low byte first and read table (last page) to TBLH and Data Memory

Description Increment table pointer low byte, TBLP, first and then the low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.

Operation [m] ← program code (low byte)
TBLH ← program code (high byte)

Affected flag(s) None

LXOR A,[m] Logical XOR Data Memory to ACC

Description Data in the Accumulator and the specified Data Memory perform a bitwise logical XOR operation. The result is stored in the Accumulator.

Operation ACC ← ACC ″XOR″ [m]

Affected flag(s) Z

LXORM A,[m] Logical XOR ACC to Data Memory

Description Data in the specified Data Memory and the Accumulator perform a bitwise logical XOR operation. The result is stored in the Data Memory.

Operation [m] ← ACC ″XOR″ [m]

Affected flag(s) Z

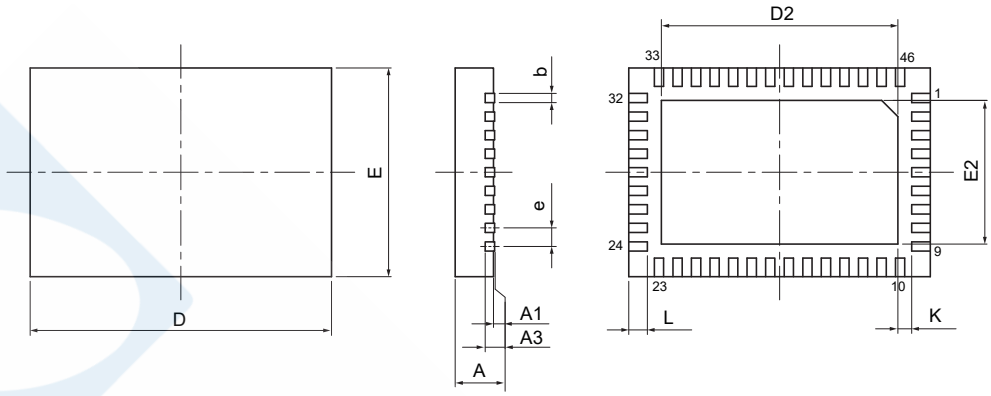
Package Information

Note that the package information provided here is for consultation purposes only. As this information may be updated at regular intervals users are reminded to consult the Holtek website for the latest version of the Package/Carton Information.

Additional supplementary information with regard to packaging is listed below. Click on the relevant section to be transferred to the relevant website page.

- Package Information (include Outline Dimensions, Product Tape and Reel Specifications)
- The Operation Instruction of Packing Materials
- Carton information

SAW Type 46-pin QFN (6.5mm×4.5mm×0.75mm) Outline Dimensions



Symbol	Dimensions in inch		
	Min.	Nom.	Max.
A	0.028	0.030	0.031
A1	0.000	0.001	0.002
A3		0.008 REF	
b	0.006	0.008	0.010
D		0.256 BSC	
E		0.177 BSC	
e		0.016 BSC	
D2	0.197	—	0.205
E2	0.118	—	0.126
L	0.014	0.016	0.018
K	0.008	—	—

Symbol	Dimensions in mm		
	Min.	Nom.	Max.
A	0.70	0.75	0.80
A1	0.00	0.02	0.05
A3		0.203 REF	
b	0.15	0.20	0.25
D		6.50 BSC	
E		4.50 BSC	
e		0.40 BSC	
D2	5.00	—	5.20
E2	3.00	—	3.20
L	0.35	0.40	0.45
K	0.20	—	—